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Robust H_∞ -Control for Uncertain Stochastic Systems with Impulsive Effects

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Abstract: Robust stabilization and H_∞ controller design for uncertain systems with impulsive and stochastic effects have been deeply discussed. Some sufficient conditions for the considered system to be robustly stable are derived in terms of linear matrix inequalities (LMIs). In addition, an example with simulations is given to better demonstrate the usefulness of the proposed H_∞ controller design method.

Keywords: impulsive stochastic systems; uncertain; H_∞ -control; robustly stochastically stable (RSS)

1. Introduction

Analysis and synthesis of dynamical systems with impulsive effects have attracted recurring interest for the past few decades [1–3]. The sudden change of system states at a certain point is the characteristic of a pulse dynamic system. Take an example from economics. When higher prices cause inflation, the government may raise the interest rate in real time to quickly reduce the circulation of money in the market, which is a typical pulse phenomenon. For a deterministic case, a large number of conclusions about stability and control for systems with impulsive perturbations can be found; see [4–8] and the reference therein.

However, because science and engineering applications offer stochastic models a great role in many areas, stochastic system theory has received widespread attention. Many basic results of systems without stochastic disturbance have been expanded to stochastic systems [9–16]. At the same time, the theory and application of stochastic differential equations have made great progress because it has played a key role in many fields; for example, option investment, population growth forecast, system control and filtering [17–21]. Among them, Ref. [21] gave a survey of impulsive differential equation theory that has been developed in recent years. Parameter uncertainties appear in stochastic impulsive systems, and exponential stability was analyzed in [22], guaranteed cost control was discussed in [23] and H_∞ filtering has been dealt with in [24,25]. It should be noted that for uncertain systems with impulsive and stochastic effects, little research has been carried out on robust H_∞ control, which aroused our interest.

In this note, the studies of robust stabilization and H_∞ controller design are conducted for an uncertain stochastic system with impulsive effects. Its time-varying uncertain parameters, which appear both in state, control and disturbance part, are supposed to be norm-bounded. An LMI-based sufficient condition is derived for an existing memoryless state feedback controller guaranteeing asymptotic stability and meeting H_∞ performance.

The note has the following arrangement: Section 2 begins with the problem formulation and reviews some useful definitions and lemmas; Section 3 discusses the robust stability and robust stabilization; Section 4 develops LMI-based H_∞ controller design method; Section 5 gives an example, which illustrates the applicability of the theoretical results; Section 6 summarizes the full text.

Notations. $X > 0$ ($X \geq 0$) indicates that a symmetric positive (semi-positive) definite matrix; X^T and X^{-1} represents the transpose and the inverse of X ; $\mathbb{R}^n$ is the n -dimensional Euclidean space; $\mathcal{L}_2[0, \infty)$ (respectively, $\mathcal{l}_2[0, \infty)$) is the space of the square-integrable vector functions (respectively, the squares and vector sequences) on $[0, \infty)$; $\|\cdot\|$ denotes the Euclidean vector norm; $\|\cdot\|_{\mathcal{L}_2}$ (respectively, $\|\cdot\|_{\mathcal{l}_2}$) represents the $\mathcal{L}_2[0, \infty)$ (respectively, $\mathcal{l}_2[0, \infty)$) norm on $[0, \infty)$; while $\|\cdot\|_{E_2}$ indicates the norm in $\mathcal{L}_2((\Omega, \mathbb{F}, \mathbb{P}), [0, \infty))$; $(\Omega, \mathbb{F}, \mathbb{P})$ is the complete probability space with Ω the sample space and $\mathbb{F}$ the σ -algebra of subsets of the sample space; $\mathcal{E}(\cdot)$ corresponds to the mathematical expectation; the maximum (minimum) eigenvalues of a matrix are represented by $\lambda_{\max}(\cdot)$ ($\lambda_{\min}(\cdot)$).

2. Problem Description and Preliminaries

We consider the uncertain stochastic system with impulsive effects:

$$\begin{cases} dx(t) = [(A + \Delta A(t))x(t) + (B + \Delta B(t))u(t) + (B_v + \Delta B_v(t))v(t)]dt \\ \quad + [(H + \Delta H(t))x(t) + (G + \Delta G(t))u(t) + (G_v + \Delta G_v(t))v(t)]dw(t), \quad t \neq \iota_\kappa, \\ x(\iota_\kappa) = C_\kappa x(\iota_\kappa^-) + D_\kappa \delta(\iota_\kappa), \quad t = \iota_\kappa, \quad \kappa = 0, 1, \dots, \\ z(t) = C_z x(t) + B_z u(t) + D_z v(t), \\ x(t_0) = x_0, \quad t_0 = 0, \end{cases} \quad (1)$$

where $x(t) \in \mathbb{R}^{n_1}$ is the system state, $z(t) \in \mathbb{R}^{n_3}$ is the controlled output, and $u(t) \in \mathbb{R}^{n_2}$ is the control input, $v(t) \in \mathbb{R}^{m_1}$ is the continuous disturbance of $\mathcal{L}_2[0, \infty)$. $\delta(\iota_\kappa) \in \mathbb{R}^{m_2}$ is the discrete disturbance of $\mathcal{l}_2[0, \infty)$. $w(t)$ is a one-dimensional Brownian motion defined on a complete probability space $(\Omega, \mathbb{F}, \mathbb{P})$. $\{\iota_\kappa, \kappa = 0, 1, \dots\}$ are the impulsive time instants and satisfy $0 = \iota_0 < \iota_1 < \dots < \iota_\kappa < \iota_{\kappa+1} < \dots$. Assume that $A, B, B_v, H, G, G_v, C_\kappa, D_\kappa, C_z, B_z$ and D_z are known matrices with appropriate dimensions, and $\Delta A(t), \Delta B(t), \Delta B_v(t), \Delta H(t), \Delta G(t)$ and $\Delta G_v(t)$ are unknown matrices denoting norm-bounded time-varying parameter uncertainties with the forms:

$$\begin{aligned} & [\Delta A(t) \quad \Delta B(t) \quad \Delta B_v(t) \quad \Delta H(t) \quad \Delta G(t) \quad \Delta G_v(t)] \\ & = MF(t)[N_A \quad N_B \quad N_H \quad N_G \quad N_{B_v} \quad N_{G_v}], \end{aligned} \quad (2)$$

where $M, N_A, N_B, N_H, N_G, N_{B_v}$ and N_{G_v} are known constant matrices, and $F(t) \in \mathbb{R}^{k \times l}$ is an unknown time-varying matrix function satisfying

$$F(t)^T F(t) \leq I, \quad \forall t. \quad (3)$$

For convenience, abbreviating $\Delta A(t), \Delta B(t), \Delta B_v(t), \Delta H(t), \Delta G(t), \Delta G_v(t), x(t), v(t), u(t)$ and $w(t)$ to $\Delta A, \Delta B, \Delta B_v, \Delta H, \Delta G, \Delta G_v, x, v, u$ and w , where $\Delta A, \Delta B, \Delta B_v, \Delta H, \Delta G$ and ΔG_v are considered admissible if both (2) and (3) hold.

Now, we recall some basic concepts about robust stability and stabilization.

Definition 1 ([13]). The impulsive stochastic system (1) is said to be mean-square stable with $u = 0, \delta(\iota_\kappa) = 0$ and $v = 0$, if there is a $\alpha(\varepsilon) > 0$ for $\forall \varepsilon > 0$, when $\mathcal{E}|x_0| < \alpha(\varepsilon), t > 0$, such that $\mathcal{E}|x|^2 < \varepsilon$. If $\lim_{t \rightarrow \infty} \mathcal{E}|x|^2 = 0$ satisfies any initial conditions, then (1) with $u = 0, \delta(\iota_\kappa) = 0$ and $v = 0$ is called mean-square asymptotically stable. And, the system (1) is said to be robustly stochastic stability (RSS for short) if (1) with $v = 0, \delta(\iota_\kappa) = 0$ and $u = 0$ is mean-square asymptotically stable for all admissible uncertainties ΔA and ΔH .

Definition 2 ([13]). Given a real number $\gamma > 0$, the system (1) is said to be RSS and the H_∞ performance γ exists, if it is RSS in the sense of Definition 1 and under zero initial conditions, for all $v \in \mathcal{L}_2[0, \infty)$, $\delta \in \mathcal{L}_2[0, \infty)$ and all admissible uncertainties $\Delta A, \Delta H$, the inequality $\|z\|_{\mathcal{E}_2} \leq \gamma(\|v\|_{\mathcal{L}_2}^2 + \|\delta\|_{\mathcal{L}_2}^2)^{1/2}$ holds.

Next, We list two lemmas, which are very important for the discussion in later chapters.

Lemma 1 ([26]). Let matrices $\mathfrak{R}, \mathfrak{Q}, \mathfrak{H}, \mathfrak{N}$ and F with appropriate dimensions satisfying $\mathfrak{N} > 0$ and $F^T F \leq I$, then:

(1) For scalar $\epsilon > 0$ and vectors $a, b \in \mathbb{R}^n$,

$$2a^T \mathfrak{Q} F \mathfrak{H} b \leq \epsilon^{-1} a^T \mathfrak{Q} \mathfrak{Q}^T a + \epsilon b^T \mathfrak{H}^T \mathfrak{H} b;$$

(2) For $\forall \epsilon > 0$ to make $\mathfrak{N} - \epsilon \mathfrak{Q} \mathfrak{Q}^T > 0$,

$$(\mathfrak{N} + \mathfrak{Q} F \mathfrak{H})^T \mathfrak{N}^{-1} (\mathfrak{N} + \mathfrak{Q} F \mathfrak{H}) \leq \mathfrak{N}^T (\mathfrak{N} - \epsilon \mathfrak{Q} \mathfrak{Q}^T)^{-1} \mathfrak{N} + \epsilon^{-1} \mathfrak{H}^T \mathfrak{H}.$$

Lemma 2 ([27]). Let $\Delta_1, \Delta_2, \Delta_3$ be given matrices, where $\Delta_1 = \Delta_1^T$, $\Delta_2 > 0$ and $\Delta_3 = \Delta_3^T$, then the following inequalities are equivalent:

(1) $\Delta_1 + \Delta_2 \Delta_3^{-1} \Delta_2^T < 0$;

(2) $\begin{bmatrix} \Delta_1 & \Delta_2 \\ \Delta_2^T & -\Delta_3 \end{bmatrix} < 0$.

3. Robust Stabilization

In this chapter, we restrict our study to the uncontrolled system (i.e., $v(t) = 0$ and $\delta(\iota_\kappa) = 0$ in (1)):

$$\begin{cases} dx(t) = [(A + \Delta A)x(t) + (B + \Delta B)u(t)]dt \\ \quad + [(H + \Delta H)x(t) + (G + \Delta G)u(t)]dw(t), \quad t \neq \iota_\kappa, \\ x(\iota_\kappa) = C_\kappa x(\iota_\kappa^-), \quad t = \iota_\kappa, \quad \kappa = 0, 1, \dots, \\ z(t) = C_z x(t) + B_z u(t), \\ x(t_0) = x_0, \quad t_0 = 0. \end{cases} \quad (4)$$

First of all, we present some sufficient conditions for RSS of (4) with $u(t) = 0$.

Theorem 1. Assume there exist two positive scalars $\epsilon_1 > 0$, $\epsilon_2 > 0$ and matrix $X > 0$, such that:

$$\begin{bmatrix} XA^T + AX + \epsilon_1 MM^T & XN_A^T & XN_H^T & XH^T \\ N_A X & -\epsilon_1 & 0 & 0 \\ N_H X & 0 & -\epsilon_2 & 0 \\ HX & 0 & 0 & -(X - \epsilon_2 MM^T) \end{bmatrix} < 0, \quad (5)$$

$$\begin{bmatrix} -X & XC_\kappa^T \\ C_\kappa X & -X \end{bmatrix} \leq 0, \quad \kappa = 0, 1, \dots, \quad (6)$$

then (4) with $u(t) = 0$ is mean-square asymptotically stable.

Proof. Consider (4) with $u(t) = 0$, that is,

$$\begin{cases} dx(t) = [(A + \Delta A)x(t)]dt + [(H + \Delta H)x(t)]dw(t), \quad t \neq \iota_\kappa, \\ x(\iota_\kappa) = C_\kappa x(\iota_\kappa^-), \quad t = \iota_\kappa, \quad \kappa = 0, 1, \dots, \\ x(t_0) = x_0, \quad t_0 = 0. \end{cases} \quad (7)$$

Let $\forall \varepsilon_1 > 0, \varepsilon_2 > 0$, and matrix $X > 0$ be a solution of (5) and (6). Let

$$P = X^{-1}. \quad (8)$$

For $t > 0$ and $t \in [\iota_\kappa, \iota_{\kappa+1})$, Define

$$V(x) = x^T P x. \quad (9)$$

Then, along the trajectory of (7) and apply the Itô's formula [12], we can get

$$dV(x) = \mathbb{L}V(x)dt + 2x^T P[H + \Delta H]x dw, \quad (10)$$

where

$$\mathbb{L}V(x) = 2x^T P(A + \Delta A)x + x^T [H + \Delta H]^T P[H + \Delta H]x. \quad (11)$$

Applying (2), (3) and Lemma 1, for $\forall \varepsilon > 0$,

$$\begin{aligned} 2x^T P(A + \Delta A)x &= 2x^T P(A + MF(t)N_A)x \\ &\leq x^T (A^T P + PA + \varepsilon_1 P M M^T P + \varepsilon_1^{-1} N_A^T N_A)x, \end{aligned} \quad (12)$$

and

$$\begin{aligned} &x^T [(H + \Delta H)^T P(H + \Delta H)]x \\ &= x^T [(H + MF N_H)^T P(H + MF N_H)]x \\ &\leq x^T [H^T (P^{-1} - \varepsilon_2 M M^T)^{-1} H + \varepsilon_2^{-1} N_H^T N_H]x. \end{aligned} \quad (13)$$

Hence, from (11)–(13), we have

$$\begin{aligned} \mathbb{L}V(x) &\leq x^T [A^T P + PA + \varepsilon_1 P M M^T P + \varepsilon_1^{-1} N_A^T N_A \\ &\quad + H^T (P^{-1} - \varepsilon_2 M M^T)^{-1} H + \varepsilon_2^{-1} N_H^T N_H]x \\ &= x^T \Xi x, \end{aligned} \quad (14)$$

where $\Xi = A^T P + PA + \varepsilon_1 P M M^T P + \varepsilon_1^{-1} N_A^T N_A + H^T (P^{-1} - \varepsilon_2 M M^T)^{-1} H + \varepsilon_2^{-1} N_H^T N_H$.

Pre- and post-multiplying (5) by $\text{diag}\{P, I, I, I\}$, and by means of Lemma 2, we can get that

$$\begin{bmatrix} A^T P + PA + \varepsilon_1 P M M^T P & N_A^T & N_H^T & H^T \\ N_A & -\varepsilon_1 & 0 & 0 \\ N_H & 0 & -\varepsilon_2 & 0 \\ H & 0 & 0 & -(P^{-1} - \varepsilon_2 M M^T) \end{bmatrix} < 0. \quad (15)$$

By Lemma 2 again, Ξ is equivalent to (15), which shows Ξ is negative-definite. Considering (14), for $t \in [\iota_\kappa, \iota_{\kappa+1})$ and $\forall x \neq 0$, we can obtain

$$\mathbb{L}V(x) \leq -\lambda |x|^2, \quad (16)$$

where $\lambda = \lambda_{\min}(-\Xi) > 0$. Therefore

$$dV(x) \leq -\lambda |x|^2 dt + 2x^T P[H + \Delta H]x dw. \quad (17)$$

Setting $\xi = \lambda / \lambda_{\max}(P) > 0$, by using the integration-by-parts Formula [28] for (17), one get

$$d[e^{\xi t} V(x)] \leq 2e^{\xi t} x^T P[H + \Delta H]x dw.$$

The inequality integral from ι_κ to t , we yields

$$\mathcal{E}[V(x(t))] \leq e^{\xi(\iota_\kappa - t)} \mathcal{E}[V(x(\iota_\kappa))]. \quad (18)$$

In view of (6), pre- and post-multiplying by P , it gives

$$C_\kappa^T P C_\kappa - P \leq 0, \quad \kappa = 0, 1, \dots. \quad (19)$$

Then we have

$$\mathcal{E}V(x(\iota_\kappa)) - \mathcal{E}V(x(\iota_\kappa^-)) = \mathcal{E}[x^T(\iota_\kappa^-)(C_\kappa^T P C_\kappa - P)x(\iota_\kappa^-)] \leq 0.$$

That is,

$$\mathcal{E}[V(x(\iota_\kappa))] \leq \mathcal{E}[V(x(\iota_\kappa^-))], \quad \kappa = 0, 1, \dots. \quad (20)$$

So, for $t \in [\iota_\kappa, \iota_{\kappa+1})$, by (18) and (20), we prove that

$$\mathcal{E}[V(x(t))] \leq e^{\xi(\iota_\kappa - t)} \mathcal{E}[V(x(\iota_\kappa))] \leq e^{\xi(\iota_\kappa^- - t)} \mathcal{E}[V(x(\iota_\kappa^-))]. \quad (21)$$

Similarly, we have

$$\begin{aligned} e^{\xi(\iota_\kappa^- - t)} \mathcal{E}[V(x(\iota_\kappa^-))] &\leq e^{\xi(\iota_{\kappa-1}^- - t)} \mathcal{E}[V(x(\iota_{\kappa-1}^-))], \\ e^{\xi(\iota_{\kappa-1}^- - t)} \mathcal{E}[V(x(\iota_{\kappa-1}^-))] &\leq e^{\xi(\iota_{\kappa-2}^- - t)} \mathcal{E}[V(x(\iota_{\kappa-2}^-))], \\ &\vdots \\ e^{\xi(\iota_1^- - t)} \mathcal{E}[V(x(\iota_1^-))] &\leq e^{-\xi t} \mathcal{E}[V(x(t_0))]. \end{aligned}$$

These implies for $\forall t \geq 0$,

$$\mathcal{E}[V(x)] \leq e^{-\xi t} \mathcal{E}[V(x_0)], \quad (22)$$

Thus, let $\eta = \mathcal{E}[V(x_0)] / \lambda_{\min}(P)$, we can deduce

$$\mathcal{E}|x|^2 \leq e^{-\xi t} \eta$$

which means that (7) is mean-square asymptotically stable. $\square$

We will design a memoryless state feedback controller with the form:

$$u(t) = Kx(t) \quad (23)$$

Making the resulting closed-loop system is RSS, $K \in \mathcal{R}^{m \times n}$ is a constant gain.

Applying (23) to (4), generate the following closed-loop systems:

$$\begin{cases} dx(t) = [(\tilde{A} + \Delta \tilde{A})x(t)]dt + (\tilde{H} + \Delta \tilde{H})x(t)]dw(t), & t \neq \iota_\kappa, \\ x(\iota_\kappa) = C_\kappa x(\iota_\kappa^-), & t = \iota_\kappa, \quad \kappa = 0, 1, \dots, \\ x(t_0) = x_0, & t_0 = 0, \end{cases} \quad (24)$$

where $\tilde{A} = A + BK$, $\Delta \tilde{A} = \Delta A + \Delta BK$, $\tilde{H} = H + GK$ and $\Delta \tilde{H} = \Delta H + \Delta GK$, in which $N_{\tilde{A}} = N_A + N_B K$ and $N_{\tilde{H}} = N_H + N_G K$.

Invoking by Theorem 1, it gives

Theorem 2. For the uncertain impulsive stochastic system (24), assume there are $\varepsilon_1 > 0$, $\varepsilon_2 > 0$, and a matrices $X > 0$, such that:

$$\begin{bmatrix} X\tilde{A}^T + \tilde{A}X + \varepsilon_1 MM^T & XN_{\tilde{A}}^T & XN_{\tilde{H}}^T & X\tilde{H}^T \\ N_{\tilde{A}}X & -\varepsilon_1 & 0 & 0 \\ N_{\tilde{H}}X & 0 & -\varepsilon_2 & 0 \\ \tilde{H}X & 0 & 0 & -(X - \varepsilon_2 MM^T) \end{bmatrix} < 0, \quad (25)$$

$$\begin{bmatrix} -X & XC_{\kappa}^T \\ C_{\kappa}X & -X \end{bmatrix} \leq 0, \quad \kappa = 0, 1, \dots, \quad (26)$$

then (4) is robustly stable with controller (23) and $K = YX^{-1}$.

In order to synthesize the gain of the controller, we transform (25) into an easy to calculate form. Note

$$\begin{aligned} \tilde{A}X &= (A + BK)X, \quad N_{\tilde{A}}X = (N_A + N_BK)X, \\ \tilde{H}X &= (H + GK)X, \quad N_{\tilde{H}}X = (N_H + N_GK)X. \end{aligned}$$

Letting $K = YX^{-1}$, (25) is equivalent to the following LMI:

$$\begin{bmatrix} \Psi_1 & XN_A^T + Y^TN_B^T & XN_H^T + Y^TN_G^T & XH^T + Y^TG^T \\ N_AX + N_BY & -\varepsilon_1 & 0 & 0 \\ N_HX + N_GY & 0 & -\varepsilon_2 & 0 \\ HX + GY & 0 & 0 & -(X - \varepsilon_2 MM^T) \end{bmatrix} < 0, \quad (27)$$

where $\Psi_1 = XA^T + AX + BY + Y^TB^T + \varepsilon_1 MM^T$.

Remark 1. Theorem 2 gives a sufficient condition for robust stability of (4), which can be validated effectively by LMIs method. We can also stabilize the feedback gain (4) by solving LMIs.

4. Robust H_{∞} Control

This part is mainly used to study the robust H_{∞} -control problem for (1).

Theorem 3. For the uncertain impulsive stochastic system (1). Given $\gamma > 0$, if there are $\hat{\varepsilon}_1 > 0$, $\hat{\varepsilon}_2 > 0$, matrices $X > 0$ and Y , such that:

$$\begin{bmatrix} \Psi_2 & B_v & XN_A^T + Y^TN_B^T & XN_H^T + Y^TN_G^T & XC_z^T + Y^TB_z^T & XH^T + Y^TG^T \\ B_v^T & -\gamma^2 I & N_{B_v}^T & N_{G_v}^T & D_z^T & G_v^T \\ N_AX + N_BY & N_{B_v} & -\hat{\varepsilon}_1 I & 0 & 0 & 0 \\ N_HX + N_GY & N_{G_v} & 0 & -\hat{\varepsilon}_2 I & 0 & 0 \\ C_zX + B_zY & D_z & 0 & 0 & -I & 0 \\ HX + GY & G_v & 0 & 0 & 0 & \hat{\varepsilon}_2 MM^T - X \end{bmatrix} < 0, \quad (28)$$

$$\begin{bmatrix} -X & 0 & XC_{\kappa}^T \\ 0 & -\gamma^2 I & D_{\kappa}^T \\ C_{\kappa}X & D_{\kappa} & -X \end{bmatrix} \leq 0, \quad \kappa = 0, 1, \dots, \quad (29)$$

where $\Psi_2 = XA^T + AX + BY + Y^T B^T + \varepsilon_1 MM^T$, then (1) is called have H_∞ performance level γ under zero initial condition. Under this circumstance, an H_∞ state feedback controller can be selected by

$$u(t) = Kx(t), \quad (30)$$

in which $K = YX^{-1}$.

Proof. By substituting (30), (1) becomes

$$\begin{cases} dx(t) = [(\tilde{A} + \Delta\tilde{A})x(t) + (B_v + \Delta B_v)v(t)]dt \\ \quad + [(\tilde{H} + \Delta\tilde{H})x(t) + (G_v + \Delta G_v)v(t)]dw(t), \quad t \neq \iota_\kappa, \\ x(\iota_\kappa) = C_\kappa x(\iota_\kappa^-) + D_\kappa \delta(\iota_\kappa), \quad t = \iota_\kappa, \quad \kappa = 0, 1, \dots, \\ z(t) = (C_z + B_z K)x(t) + D_z v(t) \\ x(t_0) = x_0, \quad t_0 = 0. \end{cases} \quad (31)$$

By (28), it is easy to infer that the LMI in (25) holds. Therefore, from the theorem 2, it can be concluded that the closed-loop system (31) is robustly stable. The next, we will prove (31) satisfies

$$\|z\|_{\mathbf{E}_2} \leq \gamma (\|v\|_{\mathbf{L}_2}^2 + \|\delta\|_{\mathbf{L}_2}^2)^{1/2} \quad (32)$$

for all nonzero $v(t) \in \mathbb{R}^{m_1}$, $\delta(\iota_\kappa) \in \mathbb{R}^{m_2}$ under zero initial condition.

Let $X = P^{-1}$. Pre-and post-multiplying (28) by $\text{diag}[P, I, I, I, I, I]$, it gives

$$\begin{bmatrix} P\tilde{A} + \tilde{A}^T P + \hat{\varepsilon}_1 PMM^T P & PB_v & N_{\tilde{A}}^T & N_{\tilde{H}}^T & C_z^T + K^T B_z^T & \tilde{H}^T \\ B_v^T P & -\gamma^2 I & N_{B_v}^T & N_{G_v}^T & D_z^T & G_v^T \\ N_{\tilde{A}} & N_{B_v} & -\hat{\varepsilon}_1 I & 0 & 0 & 0 \\ N_{\tilde{H}} & N_{G_v} & 0 & -\hat{\varepsilon}_2 I & 0 & 0 \\ C_z + B_z K & D_z & 0 & 0 & -I & 0 \\ \tilde{H} & G_v & 0 & 0 & 0 & \hat{\varepsilon}_2 MM^T - P^{-1} \end{bmatrix} < 0. \quad (33)$$

For (31), applying the Itô's formula to $V(x) = x^T P x$, for $t \in [\iota_\kappa, \iota_{\kappa+1})$, we have that

$$\begin{aligned} dV(x) = & \mathbb{L}V(x)dt + [x^T(\tilde{H} + \Delta\tilde{H})^T P x + v^T(G_v + \Delta G_v)^T P x]dw \\ & + [x^T P(\tilde{H} + \Delta\tilde{H})x + x^T P(G_v + \Delta G_v)v]dw, \end{aligned} \quad (34)$$

where

$$\begin{aligned} \mathbb{L}V(x) = & 2x^T P[(\tilde{A} + \Delta\tilde{A})x + (B_v + \Delta B_v)v] \\ & + [(\tilde{H} + \Delta\tilde{H})x + (G_v + \Delta G_v)v]^T P \\ & \times [(\tilde{H} + \Delta\tilde{H})x + (G_v + \Delta G_v)v]. \end{aligned} \quad (35)$$

Noting $P^{-1} - \hat{\varepsilon}_2 MM^T > 0$ and in view of Lemma 1, it can be shown that for $\hat{\varepsilon}_1 > 0, \hat{\varepsilon}_2 > 0$

$$\begin{aligned} & 2x^T P[(\tilde{A} + \Delta\tilde{A})x + (B_v + \Delta B_v)v] \\ & \leq 2x^T P\tilde{A}x + 2x^T PB_v v + \hat{\varepsilon}_1 x^T PMM^T P x \\ & \quad + \hat{\varepsilon}_1^{-1} (N_{\tilde{A}}x + N_{B_v}v)^T (N_{\tilde{A}}x + N_{B_v}v), \end{aligned} \quad (36)$$

and

$$\begin{aligned} & [(\tilde{H} + \Delta\tilde{H})x + (G_v + \Delta G_v)v]^T P [(\tilde{H} + \Delta\tilde{H})x + (G_v + \Delta G_v)v] \\ & \leq (\tilde{H}x + G_v v)^T (P^{-1} - \hat{\varepsilon}_2 MM^T)^{-1} (\tilde{H}x + G_v v) \\ & \quad + \hat{\varepsilon}_2^{-1} (N_{\tilde{H}}x + N_{G_v}v)^T (N_{\tilde{H}}x + N_{G_v}v). \end{aligned} \quad (37)$$

From (35)–(37), we get

$$\mathbb{L}V(x) \leq \begin{bmatrix} x^T & v^T \end{bmatrix} Y_1 \begin{bmatrix} x \\ v \end{bmatrix}. \quad (38)$$

where

$$Y_1 = \begin{bmatrix} P\tilde{A} + \tilde{A}^T P + \hat{\varepsilon}_1 P M M^T P & P B_v \\ B_v^T P & 0 \end{bmatrix} + \hat{\varepsilon}_1^{-1} \begin{bmatrix} N_{\tilde{A}}^T \\ N_{B_v}^T \end{bmatrix} \begin{bmatrix} N_{\tilde{A}} & N_{B_v} \end{bmatrix} \\ + \begin{bmatrix} \tilde{H}^T \\ G_v^T \end{bmatrix} (P^{-1} - \hat{\varepsilon}_2 M M^T)^{-1} \begin{bmatrix} \tilde{H} & G_v \end{bmatrix} + \hat{\varepsilon}_2^{-1} \begin{bmatrix} N_{\tilde{H}}^T \\ N_{G_v}^T \end{bmatrix} \begin{bmatrix} N_{\tilde{H}} & N_{G_v} \end{bmatrix}.$$

It can be inferred that $Y_1 < 0$ from (33). Thus, combined with (34) and (38), we obtain

$$dV(x) \leq \begin{bmatrix} x^T & v^T \end{bmatrix} Y_1 \begin{bmatrix} x \\ v \end{bmatrix} dt \\ + 2x^T P[(\tilde{H} + \Delta \tilde{H})x + (G_v + \Delta G_v)v]dw. \quad (39)$$

Then, the sides of (34) are integrated from t_κ to t , we have

$$\mathcal{E}[V(x(t))] - \mathcal{E}[V(x(t_\kappa))] = \mathcal{E}\left[\int_{t_\kappa}^t dV(x(\mu))\right] = \mathcal{E}\left[\int_{t_\kappa}^t \mathbb{L}V(x(\mu))d\mu\right], \quad (40)$$

By means of (39) and (40), we get

$$\mathcal{E}[V(x(t))] = \mathcal{E}[V(x(t_\kappa))] + \mathcal{E}\left[\int_{t_\kappa}^t \mathbb{L}V(x(\mu))d\mu\right] \\ \leq \mathcal{E}[V(x(t_\kappa))] + \mathcal{E}\left\{\int_{t_\kappa}^t \begin{bmatrix} x^T(\mu) & v^T(\mu) \end{bmatrix} Y_1 \begin{bmatrix} x(\mu) \\ v(\mu) \end{bmatrix} d\mu\right\}. \quad (41)$$

Therefore, for $\forall t \in [t_\kappa, t_{\kappa+1})$,

$$\mathcal{E}\left\{\int_{t_\kappa}^t [z^T(\mu)z(\mu) - \gamma^2 v^T(\mu)v(\mu) + \mathbb{L}V(x(\mu))]d\mu\right\} \\ = \mathcal{E}\left\{\int_{t_\kappa}^t [(C_z + B_z K)x(\mu) + D_z v(\mu)]^T [(C_z + B_z K)x(\mu) + D_z v(\mu)] - \gamma^2 v^T(\mu)v(\mu) + \mathbb{L}V(x(\mu))]d\mu\right\} \\ \leq \mathcal{E}\left\{\int_{t_\kappa}^t \begin{bmatrix} x^T(\mu) & v^T(\mu) \end{bmatrix} Y \begin{bmatrix} x(\mu) \\ v(\mu) \end{bmatrix} d\mu\right\}, \quad (42)$$

where

$$Y = \begin{bmatrix} P\tilde{A} + \tilde{A}^T P + \hat{\varepsilon}_1 P M M^T P & P B_v \\ B_v^T P & -\gamma^2 \end{bmatrix} + \hat{\varepsilon}_1^{-1} \begin{bmatrix} N_{\tilde{A}}^T \\ N_{B_v}^T \end{bmatrix} \begin{bmatrix} N_{\tilde{A}} & N_{B_v} \end{bmatrix} \\ + \begin{bmatrix} \tilde{H}^T \\ G_v^T \end{bmatrix} (P^{-1} - \hat{\varepsilon}_2 M M^T)^{-1} \begin{bmatrix} \tilde{H} & G_v \end{bmatrix} + \hat{\varepsilon}_2^{-1} \begin{bmatrix} N_{\tilde{H}}^T \\ N_{G_v}^T \end{bmatrix} \begin{bmatrix} N_{\tilde{H}} & N_{G_v} \end{bmatrix} \\ + \begin{bmatrix} (C_z + B_z K)^T \\ D_z^T \end{bmatrix} \begin{bmatrix} C_z + B_z K & D_z \end{bmatrix}.$$

Because of $Y < 0$, it can be deduced from (38)–(42) that

$$\mathcal{E}\left\{\int_{t_\kappa}^t [z^T(\mu)z(\mu) - \gamma^2 v^T(\mu)v(\mu)]d\mu\right\} < \mathcal{E}[V(x(t_\kappa))]. \quad (43)$$

Let $X = P^{-1}$. Pre- and post-multiplying (29) by $\text{diag}\{P, I, I\}$, we have

$$C_{\kappa}^T P C_{\kappa} - P + C_{\kappa}^T P D_{\kappa} (\gamma^2 I - D_{\kappa}^T P D_{\kappa})^{-1} D_{\kappa}^T P C_{\kappa} \leq 0, \quad (44)$$

where $\gamma^2 I - D_{\kappa}^T P D_{\kappa} > 0$. By (31), we can confirm

$$\begin{aligned} & -\gamma^2 \delta^T(\iota_{\kappa}) \delta(\iota_{\kappa}) + \mathcal{E}[V(x(\iota_{\kappa}))] - \mathcal{E}[V(x(\iota_{\kappa}^-))] \\ & = \mathcal{E}\left\{ \begin{bmatrix} x^T(\iota_{\kappa}^-) & \delta^T(\iota_{\kappa}) \end{bmatrix} \begin{bmatrix} C_{\kappa}^T P C_{\kappa} - P & C_{\kappa}^T P D_{\kappa} \\ D_{\kappa}^T P C_{\kappa} & D_{\kappa}^T P D_{\kappa} - \gamma^2 I \end{bmatrix} \begin{bmatrix} x(\iota_{\kappa}^-) \\ \delta(\iota_{\kappa}) \end{bmatrix} \right\} \\ & \leq \mathcal{E}\{x^T(\iota_{\kappa}^-) [C_{\kappa}^T P C_{\kappa} - P + C_{\kappa}^T P D_{\kappa} (\gamma^2 I - D_{\kappa}^T P D_{\kappa})^{-1} D_{\kappa}^T P C_{\kappa}] x(\iota_{\kappa}^-)\} \leq 0. \end{aligned} \quad (45)$$

That is, relying on (19), we can know

$$\mathcal{E}[V(x(\iota_{\kappa}))] - \mathcal{E}[V(x(\iota_{\kappa}^-))] = \mathcal{E}[x^T(\iota_{\kappa}^-) (C_{\kappa}^T P C_{\kappa} - P) x(\iota_{\kappa}^-)] \leq 0.$$

Namely, $\mathcal{E}[V(x(\iota_{\kappa}))] \leq \mathcal{E}[V(x(\iota_{\kappa}^-))]$. So, combined with (45), we infer that

$$-\gamma^2 \delta^T(\iota_{\kappa}) \delta(\iota_{\kappa}) \leq \mathcal{E}[V(x(\iota_{\kappa}^-))] - \mathcal{E}[V(x(\iota_{\kappa}))], \quad (46)$$

Also, we have

$$\begin{aligned} & -\gamma^2 \delta^T(\iota_{\kappa}^-) \delta(\iota_{\kappa}^-) \leq \mathcal{E}[V(x(\iota_{\kappa-1}^-))] - \mathcal{E}[V(x(\iota_{\kappa}^-))], \\ & -\gamma^2 \delta^T(\iota_{\kappa-1}^-) \delta(\iota_{\kappa-1}^-) \leq \mathcal{E}[V(x(\iota_{\kappa-2}^-))] - \mathcal{E}[V(x(\iota_{\kappa-1}^-))], \\ & \quad \vdots \\ & -\gamma^2 \delta^T(\iota_1^-) \delta(\iota_1^-) \leq \mathcal{E}[V(x(\iota_0))] - \mathcal{E}[V(x(\iota_1^-))]. \end{aligned} \quad (47)$$

From the above inequalities, we deduce

$$-\sum_{\iota_{\kappa} \in (0, t)} \gamma^2 \delta^T(\iota_{\kappa}) \delta(\iota_{\kappa}) \leq \mathcal{E}[V(x(\iota_0))] - \mathcal{E}[V(x(\iota_{\kappa}))]. \quad (48)$$

Note the zero initial conditions and (43) over all possible ι_{κ} in $[0, t]$, it results in

$$\mathcal{E}\left\{ \int_{\iota_{\kappa}}^t [z^T(\mu) z(\mu) - \gamma^2 v^T(\mu) v(\mu)] d\mu \right\} - \sum_{\iota_{\kappa} \in (0, t)} \gamma^2 \delta^T(\iota_{\kappa}) \delta(\iota_{\kappa}) < 0, \quad (49)$$

which means that (32) is satisfied. This proof is complete. $\square$

5. An Example

In this chapter, we will provide an example to better illustrate the usefulness of the proposed method.

Example 1. Consider a two-dimensional uncertain impulsive stochastic system (1) with the following parameters:

$$\begin{aligned} A &= \begin{bmatrix} -2 & 1.2 \\ 0.8 & -2 \end{bmatrix}, \quad B = \begin{bmatrix} -1 & 2 \\ 0.4 & 3 \end{bmatrix}, \quad B_v = \begin{bmatrix} 2 & 0 \\ 0.2 & -1 \end{bmatrix}, \quad H = \begin{bmatrix} 1.2 & 0 \\ -2 & 0.5 \end{bmatrix}, \\ G &= \begin{bmatrix} 0.5 & 0 \\ -1 & 2 \end{bmatrix}, \quad G_v = \begin{bmatrix} 2 & 0.4 \\ -1 & -4 \end{bmatrix}, \quad C_\kappa = \begin{bmatrix} 0.9 & 0 \\ 0 & 0.9 \end{bmatrix}, \quad D_\kappa = \begin{bmatrix} -0.2 & 1 \\ 0.3 & -0.2 \end{bmatrix}, \\ C_z &= \begin{bmatrix} 1 & 0 \\ 1.2 & 0.1 \end{bmatrix}, \quad B_z = \begin{bmatrix} 0.1 & 1 \\ 0.2 & -0.1 \end{bmatrix}, \quad D_z = \begin{bmatrix} 0.5 & 0 \\ 0.2 & -0.1 \end{bmatrix}, \quad M = \begin{bmatrix} 1 & 0 \\ 0 & -1 \end{bmatrix}, \\ N_A &= \begin{bmatrix} 0.1 & -0.2 \end{bmatrix}, \quad N_B = \begin{bmatrix} 0.2 & 0.4 \end{bmatrix}, \quad N_H = \begin{bmatrix} 0.1 & 0.3 \end{bmatrix}, \quad N_G = \begin{bmatrix} 0.2 & -0.1 \end{bmatrix}, \\ N_{B_v} &= \begin{bmatrix} 0.1 & -0.1 \end{bmatrix}, \quad N_{G_v} = \begin{bmatrix} 0.2 & 0.3 \end{bmatrix}. \end{aligned}$$

Set $\gamma = 1.7$, using the Matlab LMI Control toolbox, we can get the solutions to LMIs (28) and (29) are as follows:

$$X = \begin{bmatrix} 4.2597 & -5.4180 \\ -5.4180 & 23.3299 \end{bmatrix}, \quad Y = \begin{bmatrix} -7.5133 & 9.2519 \\ -3.8391 & -0.3219 \end{bmatrix}, \quad \hat{\varepsilon}_1 = 2.9631, \quad \hat{\varepsilon}_2 = 1.5302.$$

Therefore, from Theorem 3, the H_∞ control law can be chosen as:

$$u(t) = \begin{bmatrix} -1.7874 & -0.0185 \\ -1.3040 & -0.3166 \end{bmatrix} x(t).$$

We select initial value $x_0 = [1, -0.8]^T$ and the impulsive interval $\iota_{\kappa+1} = \iota_\kappa + 0.5$. Figure 1 depicts the state of the uncertain impulsive stochastic system. It is obvious from Figure 2 that the closed-loop system is mean-square asymptotically stable, where $\mathcal{E}|x(t)|^2 = \frac{1}{1000} \sum_{j=1}^{1000} [(x_1^j(t))^2 + (x_2^j(t))^2]$ which $x_s^j, s = 1, 2$ is the j th sample path.

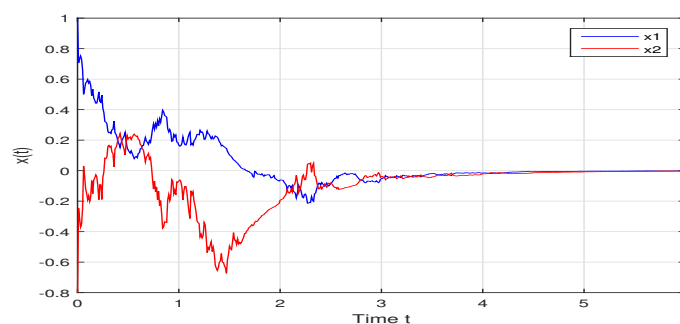


Figure 1. State path of Example 1.

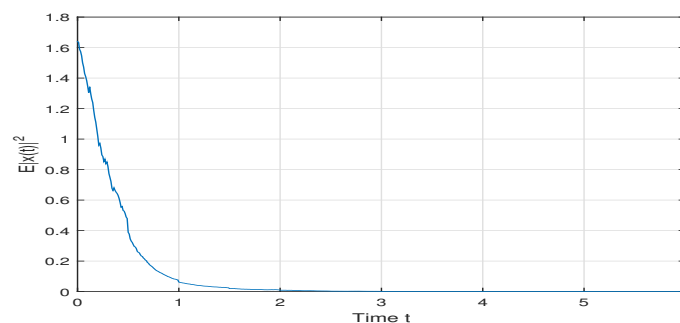


Figure 2. Trajectory of the average value of 1000 sample paths.

6. Conclusions

Robust stabilization and H_∞ control are considered in this paper for stochastic systems with uncertainties and impulsive effects. As for robust stability and robust stabilization, LMIs-based sufficient conditions have been established. Moreover, we proposed a reasonable H_∞ controller design method and its effectiveness has been demonstrated by a numerical example.

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References

1. Yang, T. *Impulsive Control Theory*; Springer: Berlin, Germany, 2001.
2. Yang, T. *Impulsive Systems and Control: Theory and Applications*; Nova Science Publishers: New York, NY, USA, 2001.
3. Li, X.; Cao, J. An impulsive delay inequality involving unbounded time-varying delay and applications. *IEEE Trans. Autom. Control* **2017**, *62*, 3618–3625. [[CrossRef](#)]
4. Li, Z.; Soh, Y.; Wen, C.; Zhai, G. Switched and impulsive systems: Analysis, design, and applications. *IEEE Trans. Autom. Control* **2006**, *51*, 2010–2011.
5. Haddad, W.M.; Chellaboina, V.; Kablar, N.A. Non-linear impulsive dynamical systems. Part II: Stability of feedback interconnections and optimality. *Int. J. Control* **2001**, *74*, 1659–1677. [[CrossRef](#)]
6. Guan, Z.; Chen, G.; Ueta, T. On impulsive control of a periodically forced chaotic pendulum system. *IEEE Trans. Autom. Control* **2000**, *45*, 1724–1727. [[CrossRef](#)]
7. Cheng, H.; Wang, F.; Zhang, T. Multi-state dependent impulsive control for Holling I predator-prey model. *Discret. Dyn. Nat. Soc.* **2012**, *2012*, 181752. [[CrossRef](#)]
8. Li, Y.; Cheng, H.; Wang, Y. A Lycaon pictus impulsive state feedback control model with Allee effect and continuous time delay. *Adv. Differ. Equ.* **2018**, *2018*, 367. [[CrossRef](#)]
9. Li, X.; Bohner, M.; Wang, C. Impulsive differential equations: Periodic solutions and applications. *Automatica* **2015**, *52*, 173–178. [[CrossRef](#)]
10. Chen, W.; Zheng, W. Input-to-state stability and integral input-to-state stability of nonlinear impulsive systems with delays. *Automatica* **2009**, *45*, 1481–1488. [[CrossRef](#)]
11. Karthikeyan, S.; Balachandran, K. Constrained controllability of nonlinear stochastic impulsive systems. *Int. J. Appl. Math. Comput. Sci.* **2011**, *21*, 307–316. [[CrossRef](#)]
12. Spiliotis, J.; Tsinias, J. Notions of exponential robust stochastic stability, ISS and their Lyapunov characterizations. *Int. J. Robust Nonlinear Control* **2003**, *13*, 173–187. [[CrossRef](#)]
13. Rodkina, A.; Schurz, H.; Shaikhet, L. Almost sure stability of some stochastic dynamical systems with memory. *Discret. Contin. Dyn. Syst.* **2008**, *21*, 571–593.
14. Zhou, S.; Liu, X.; Chen, B.; Liu, H. Stability analysis for a Class of discrete-time nonhomogeneous Markov jump systems with multiplicative noises. *Complexity* **2018**, *2018*, 1586846. [[CrossRef](#)]
15. Zhang, W.; Ma, L.; Zhang, T. Discrete-time mean-field stochastic H_2/H_∞ control. *J. Syst. Sci. Complex.* **2017**, *30*, 765–781. [[CrossRef](#)]
16. Tan, C.; Zhang, W. On observability and detectability of continuous-time stochastic Markov jump systems. *J. Syst. Sci. Complex.* **2015**, *28*, 830–847. [[CrossRef](#)]
17. Oksendal, B. *Stochastic Differential Equations: An Introduction with Application*, 5th ed.; Springer: New York, NY, USA, 1998.
18. Li, C.; Sun, J.; Sun, R. Stability analysis of a class of stochastic differential delay equations with nonlinear impulsive effects. *J. Frankl. Inst.* **2010**, *347*, 1186–1198. [[CrossRef](#)]
19. Yang, Z.; Xu, D. Mean square exponential stability of impulsive stochastic difference equations. *Appl. Math. Lett.* **2007**, *20*, 938–945. [[CrossRef](#)]
20. Perestyuk, M.O.; Chernikova, O.S. Some modern aspects of the theory of impulsive differential equations. *Ukr. Math. J.* **2008**, *60*, 91–107. [[CrossRef](#)]

21. Lakshmikantham, V.; Bainov, D.D.; Simeonov, P.S. *Theory of Impulsive Differential Equations*; World Scientific: Singapore, 1989.
22. Zhang, Y.; Wang, C. Robust stochastic stability of uncertain discrete-time impulsive Markovian jump delay systems with multiplicative noises. *Int. J. Syst. Sci.* **2015**, *46*, 2210–2220. [[CrossRef](#)]
23. Dong, Y.; Wang, Q.; Sun, J. Guaranteed cost control for a class of uncertain stochastic impulsive systems with Markovian switching. *Stoch. Anal. Appl.* **2009**, *27*, 1174–1190. [[CrossRef](#)]
24. Xu, S.; Chen, T. Robust H_∞ filtering for uncertain impulsive stochastic systems under sampled measurements. *Automatica* **2003**, *39*, 509–516. [[CrossRef](#)]
25. Dong, Y.; Sun, J.; Wu, Q. H_∞ filtering for a class of stochastic Markovian jump systems with impulsive effects. *Int. J. Robust Nonlinear Control* **2008**, *18*, 1–13. [[CrossRef](#)]
26. Chen, W.; Wang, J.; Tang, Y.; Lu, X. Robust H_∞ control of uncertain linear impulsive stochastic system. *Int. J. Robust Nonlinear Control* **2008**, *18*, 1348–1371. [[CrossRef](#)]
27. Boyd, S.; El Ghaoui, L.; Feron, E.; Balakrishnan, V. *Linear Matrix Inequalities in System and Control Theory*; SIAM: Philadelphia, PA, USA, 1994.
28. Mao, X. Stability of stochastic differential equations with Markovian switching. *Stoch. Process. Appl.* **1999**, *79*, 45–67. [[CrossRef](#)]



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