

Article

Design of a Novel MEMS Gyroscope Array

Wei Wang *, Xiaoyong Lv and Feng Sun

College of Automation, Harbin Engineering University, No. 145 Nantong Street, Harbin 150001, Heilongjiang, China; E-Mails: lengfeng0506@163.com (X.L.); Sunfeng_hrbeu@163.com (F.S.)

* Author to whom correspondence should be addressed; E-Mail: chinaww2006@yahoo.com.cn; Tel.: +86-451-8256-8488; Fax: +86-451-8251-8741.

Received: 23 November 2012; in revised form: 18 January 2013 / Accepted: 21 January 2013 / Published: 28 January 2013

Abstract: This paper reports a novel four degree-of-freedom (DOF) MEMS vibratory gyroscope. A MEMS gyroscope array is then presented using the novel gyroscope unit. In the design of the proposed 4-DOF MEMS vibratory gyroscope, the elements of the drive-mode are set inside the whole gyroscope architecture, and the elements of sense-mode are set around the drive-mode, which thus makes it possible to combine several gyroscope units into a gyroscope array through sense-modes of all the units. The complete 2-DOF vibratory structure is utilized in both the drive-mode and sense-mode of the gyroscope unit, thereby providing the desired bandwidth and inherent robustness. The gyroscope array combines several gyroscope units by using the unique detection mass, which will increase the gain of sense-mode and improve the sensitivity of the system. The simulation results demonstrate that, compared to a single gyroscope unit, the gain of gyroscope array (n=6) is increased by about 8 dB; a 3 dB bandwidth of 100 Hz in sense-mode and 190 Hz in drive-mode are also provided. The bandwidths of both modes are highly matched with each other, providing a bandwidth of 100 Hz for the entire system, thus illustrating that it could satisfy the requirements in practical applications.

Keywords: structural design; MEMS; gyroscope array; multi-DOF; wide bandwidth

Classification: PACS 85.85.+j

1. Introduction

Since the first micromachined tuning-fork gyroscope was reported by Draper laboratory in 1991 [1], many types of MEMS gyroscopes have been designed and manufactured. Due to advantages such as small size, low cost, reduced power consumption, and so on, micromachined vibratory gyroscopes have received a great deal of attention in the past several decades. Currently, some micromachined vibratory gyroscopes have been utilized in a variety of practical applications, such as in the automotive industry, electronics, and INS(inertial navigation system) for tactical weapons and vehicles, *etc*.

Conventional micromachined vibratory gyroscopes mainly employ the architecture with 1-DOF drive-mode and sense-mode, which increases the mechanical gain by matching the natural frequencies of two modes and allows enhancing the sensitivity [2–4]. These designs often utilize the design of flexural beams or tuning electronics to adjust the structural frequency for the purpose of mode-matching, which could increase the sensitivity of the system. However, the high sensitivity of these mode-matched gyroscopes is always at the cost of bandwidth or robustness and is susceptible to structural and environmental parameter variations. Meanwhile, the structural coupling between two modes is also a drawback.

To resolve the problems in the single-DOF MEMS vibratory gyroscopes, some multi-DOF micromachined vibratory gyroscopes have been presented by increasing DOF of the drive-mode or sense-mode. For the purpose of improving the robustness of MEMS vibratory gyroscopes, the 2-DOF vibratory structure has been utilized in drive-mode or sense-mode to increase the bandwidth of the respective mode. A 3-DOF MEMS vibratory gyroscope with 2-DOF drive-mode and 1-DOF sense-mode is described [5], which uses 2-DOF dynamic vibration absorber(DVA) structure in drive-mode. This design illustrated the wide bandwidth of drive-mode, but there is a narrow bandwidth of sense-mode. Another type of 3-DOF micromachined vibratory gyroscope, with 1-DOF drive-mode and 2-DOF DVA structure in sense-mode, is described [6], which showed the wide bandwidth of sense-mode.

However, there is the limitation of structural design space in the DVA architecture. Moreover, the tradeoffs among operational frequency, bandwidth and die size, as well as detection capacitance must be considered [7]. To overcome the limitation in the DVA structure, a complete 2-DOF architecture with two masses and three springs has been employed in the sense-mode of some 3-DOF micromachined vibratory gyroscopes [8–10]. This design demonstrated the advantage of gain and bandwidth in sense-mode. However, these designs merely improve the bandwidth of drive-mode or sense-mode, respectively, and in the other mode, the robustness also requires improvement.

Recently, a MEMS gyroscope with 2-DOF drive-mode and sense-mode has been proposed to improve the robustness of both drive-mode and sense-mode [11], which utilizes the complete 2-DOF vibratory structure to provide a robust gain-bandwidth frequency region and inherently improves the robustness of drive-mode and sense-mode at the same time. The device decouples drive-mode and sense-mode by adding decoupled mass and proof mass and, concurrently, a complete 2-DOF structure is formed in both modes. Compared to the previous 1-DOF structure, one of the advantages of utilizing the complete 2-DOF structure is its robust gain-bandwidth frequency region. Meanwhile, the implementation of the complete 2-DOF system will reduce the complexity of electronic circuits and eliminate the electronic noise due to the complex control circuit in drive-mode [12,13]. However, the gain of sense-mode also

needs to increase in order to improve the sensitivity of the device. In this paper, another novel 4-DOF MEMS vibratory gyroscope structure is proposed. Based on this 4-DOF MEMS vibratory gyroscope structure, a novel gyroscope array structure is presented. The proposed gyroscope array is different from the previous gyroscope structure, which combines several gyroscope units into a single gyroscope structure in a special way. The proposed gyroscope array could improve the gain of sense-mode compared to the previous 4-DOF gyroscope [11].

The rest of this paper is organized as follows. Section 2 introduces the design of 4-DOF gyroscope unit. Section 3 presents the structural design of the proposed gyroscope array based on the proposed gyroscope unit. Section 4 gives a theoretical dynamics analysis and the solutions of the gyroscope array. Section 5 reports the comparative simulation results, and Section 6 concludes the paper with a summary of results.

2. Gyroscope Unit

The conceptual schematic of proposed gyroscope unit is shown in Figure 1. Similar to the multi-DOF gyroscope [11], the complete 2-DOF vibratory structure is implemented in both drive-mode and sense-mode, which provides both modes with a large bandwidth. The drive-combs are set on the driving mass to drive drive-mode to oscillate along with x axis. However, unlike the previous 4-DOF MEMS vibratory gyroscope [11], the elements of drive-mode are inside the schematic and the sense-mode is around the drive-mode. The sensing-capacitors are set on the detection mass used to sense the Coriolis-induced changing capacitance. The detection mass is set at the edge of the schematic, which will be suitable for combining several gyroscope units into a gyroscope array.

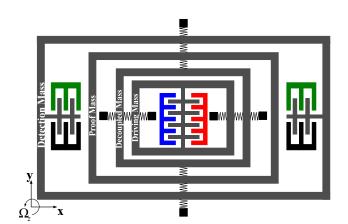
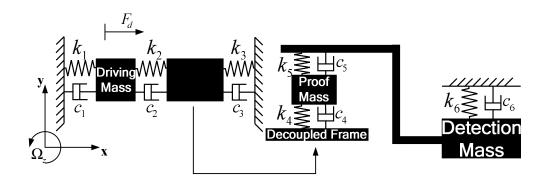


Figure 1. The schematic of proposed gyroscope unit.

The lump model of proposed 4-DOF MEMS gyroscope is simplified, as shown in Figure 2. The proposed gyroscope consists of driving mass, decoupled frame, proof mass and detection mass and springs. The driving mass, decoupled frame and springs, k_1 , k_2 , k_3 , form the complete 2-DOF drive-mode and the proof mass, detection mass and springs, k_4 , k_5 , k_6 , form the complete 2-DOF sense-mode. Drive-mode and sense-mode are linked to each other through decoupled frame and proof mass, which isolate vibration between both modes.

Figure 2. Model of the proposed 4-DOF MEMS vibratory gyroscope.

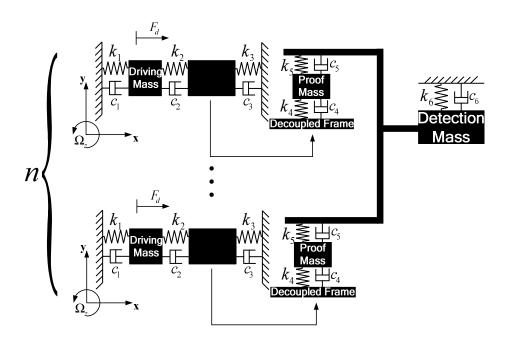


When the gyroscope unit is operated, the driving force is applied to driving mass and drives driving mass and decoupled frame to vibrate in drive direction (x). Moreover, proof mass could oscillate with decoupled frame in drive direction (x). If an angular rate Ω_z is input along the direction (z) perpendicular to x-y plane, the Coriolis-induced force will excite proof mass and detection mass to move in sense direction (y). Then the capacitors on detection mass will change, which helps pick up the motion and calculate the input angular rate Ω_z .

3. Gyroscope Array

In this section, a gyroscope array will be designed based on the gyroscope unit presented in Section 2. The lump model of gyroscope array is shown in Figure 3. There are a group of gyroscope units forming a gyroscope array. All the units utilize the identical elements of drive-mode and proof mass respectively, but there is just a detection mass, which is utilized by all the gyroscope units. The proof masses of every gyroscope unit are connected to the unique detection mass through a beam.

Figure 3. Lump model of the 4-DOF MEMS vibratory gyroscope array.



In the design of gyroscope array, similar to gyroscope unit, the complete 2-DOF vibratory structure is implemented in drive-mode of every gyroscope unit. In addition, the complete 2-DOF drive-mode could ensure the desired bandwidth the same as a gyroscope unit. In sense-mode, however, the complete 2-DOF sense-mode of a gyroscope unit is changed into another model as shown in Figures 3 and 4. The proof masses also belong to the respective gyroscope unit, but the detection masses of all the units in gyroscope array are combined to a common detection mass, which is used by all the units. Then the sense-mode of every gyroscope unit is also a complete 2-DOF vibratory structure providing a wide bandwidth to sense-mode of gyroscope array.

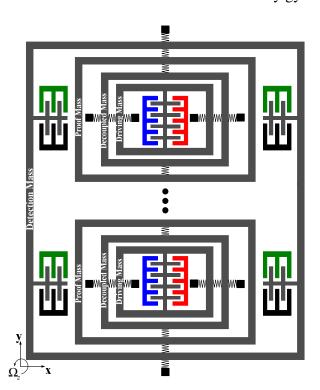


Figure 4. Combination of 4-DOF MEMS vibratory gyroscopes.

When the gyroscope array is operated, the driving masses of all the gyroscope units are driven along x axis by the sinusoidal forces with the same frequency, amplitude and phase, which ensure that the entire drive-mode will work synchronously. When there is an angular rate Ω_z perpendicular to the x-y plane, the Coriolis-induced force is applied to the respective proof mass. Then all the proof masses are driven to oscillate along y axis and detection mass is excited by the force transmitted from all the proof masses. The Coriolis-induced motion is picked up by the detection mass, which can help calculate the input angular rate Ω_z .

Compared with a single gyroscope unit, the gyroscope array combines several (*i.e.*, n) gyroscope units and possesses a unique detection mass, which will enlarge the Coriolis-induced force applied to the detection mass and increase the amplitude of the detection mass. The enlarged amplitude of the detection mass will increase the changing capacitance and sensitivity of the device. As shown in Figures 3 and 4, there are several independent drive-mode elements and proof masses, which form a special drive-mode and could improve insensitivity to structural parameters variations. The detailed analysis will be given later.

4. Theoretical Analysis

4.1. Drive-Mode

As the structural schematic of the micromachined vibratory gyroscope is shown in Figures 2 and 3, all the gyroscope units use the identical drive-mode elements and the dynamic equations of drive-mode can be represented by simplified equations,

$$m_1\ddot{x}_1 + (c_1 + c_2)\dot{x}_1 - c_2\dot{x}_2 + (k_1 + k_2)x_1 - k_2x_2 = F_d$$
(1)

$$(m_f + m_2)\ddot{x}_2 - c_2\dot{x}_1 + (c_2 + c_3)\dot{x}_2 - k_2x_1 + (k_2 + k_3)x_2 = 0$$
(2)

where x_1 and x_2 represent the motion of driving mass m_1 and decoupled frame m_f (with proof mass m_2) along with x axis. The driving force F_d is sinusoidal applied to the driving mass and drives driving mass m_1 and decoupled frame m_f (with proof mass m_2) to oscillate in the drive direction(x). The proof mass m_2 provides the link between two modes and without input angular rate, it only oscillates in the drive direction (x) with the decoupled frame m_f . When there is rotation about the x axis perpendicular to the x-y plane, the Coriolis-induced force is applied to the proof mass m_2 and drives proof mass m_2 and detection mass m_3 to move in the sense direction(y).

Because the decoupled frame m_f and proof mass m_2 provide the link between two vibratory modes and their vibration along x axis will be transferred to y axis due to Coriolis effect, the dynamic response (x_2) of decoupled frame m_f (with proof mass m_2) along x axis will occur. Assuming the solution x_2 are sinusoidal, according to Laplace transform the steady solution for x_2 is given by,

$$x_2 = \frac{(k_2 + j\omega c_2)F_d}{\Delta_d(\omega)} \tag{3}$$

where $\Delta_d(\omega) = (k_1 + k_2 - m_1\omega^2 + j\omega(c_1 + c_2))(k_3 + k_2 - (m_f + m_2)\omega^2 + j\omega(c_2 + c_3)) - (k_2 + j\omega c_2)^2$. The Equation (2) provides theoretical argument for performance analysis. The structural frequency of the complete 2-DOF dynamic system in drive-mode can be defined by,

$$\omega_{d1}^2 = \frac{k_1 + k_2}{m_1}, \omega_{d2}^2 = \frac{k_2 + k_3}{m_f + m_2} \tag{4}$$

according to vibration dynamics, the anti-resonant frequency ω_{d0} of the driving mass m_1 is always between the two resonant frequencies of drive-mode. For obtaining the precise position of the robust operational region in drive-mode, the structural frequencies ω_{d1} and ω_{d2} are set equal to ω_{d0} .

Assuming zero-damping system and solving the eigenvalue equation $\Delta_d(\omega) = 0$, the resonant frequencies of drive-mode can be obtained,

$$\omega_{dH,L} = \sqrt{\omega_{d0}^2 \pm \sqrt{\frac{k_2^2}{m_1(m_2 + m_f)}}}$$
 (5)

where $\omega_{dH,L}$ are the high and low resonant frequencies of drive-mode. Define the resonant peak spacing Δ_d as,

$$\Delta_d = \omega_{dH} - \omega_{dL} \tag{6}$$

And by using Equations (4), (5) and (6), the design equations of the stiffnesses in drive-mode are given by,

$$\begin{cases} k_2 = \Delta_d \sqrt{m_1(m_2 + m_f)} \sqrt{\omega_{d0}^2 - 0.25\Delta_d^2} \\ k_1 = m_1 \omega_{d0}^2 - k_2 \\ k_3 = (m_f + m_2)\omega_{d0}^2 - k_2 \end{cases}$$
(7)

It is obvious that the resonant peak spacing Δ_d of drive-mode can be set independently of the operational frequency ω_{d0} , which eliminates the limitation of DVA structure [7]. After the layout of driving mass m_1 and decoupled frame m_f (with proof mass m_2), the operational frequency can be set by adjusting the stiffnesses. In the proposed gyroscope array, its drive-mode includes several independent same driving elements belonging to respective gyroscope units and their dynamics equations and design equations are the same as Equations (1), (2) and (7), which is a different point in contrast to others.

4.2. Sense-Mode

As is shown in Figures 2 and 3, sense-mode elements of the gyroscope unit and gyroscope array are different from each other. Because there is proof mass m_2 in sense-mode of all gyroscope units and they are identical, the motion of proof mass m_2 in sense-mode of every gyroscope unit can be described by the same equation,

$$m_2\ddot{y}_1 + (c_4 + c_5)\dot{y}_1 - c_5\dot{y}_2 + (k_4 + k_5)y_1 - k_5y_2 = F_C$$
(8)

where y_1 represents the coordinate of proof mass m_2 in all gyroscope units and y_2 represents the coordinate of detection mass m_3 in gyroscope array. F_C is the Coriolis-induced force applied to proof mass m_2 . In gyroscope array, the unique detection mass m_3 is utilized by all the gyroscope units of gyroscope array and its motion can be given by,

$$m_3\ddot{y}_2 - nc_5\dot{y}_1 + (nc_5 + c_6)\dot{y}_2 - nk_5y_1 + (nk_5 + k_6)y_2 = 0$$
(9)

where n is the number of gyroscope units in proposed gyroscope array.

Because the detection mass m_3 is used to pick up the Coriolis-induced motion through the sensing-capacitor in it, the motion y_2 of detection mass m_3 is mainly focused on. Using Equations (8) and (9), the solution of the detection mass can be obtained,

$$y_2 = \frac{n(k_5 + j\omega c_5)F_C}{\Delta_s(\omega)} \tag{10}$$

where
$$\Delta_s(\omega) = (k_4 + k_5 - m_2\omega^2 + j\omega(c_4 + c_5))(nk_5 + k_6 - m_3\omega^2 + j\omega(nc_5 + c_6)) - n(k_5 + j\omega c_5)^2$$
.

After the design and layout of the masses, the stiffnesses of all the springs must be calculated. According to vibration dynamics, the anti-resonant frequency ω_{s0} of the proof mass m_2 is always between the two resonant frequencies of sense-mode. Define the structural frequencies of sense-mode by,

$$\omega_{s1}^2 = \frac{k_4 + k_5}{m_2}, \omega_{s2}^2 = \frac{nk_5 + k_6}{m_3} \tag{11}$$

For obtaining the precise position of the robust operational region in sense-mode, we set $\omega_{s1} = \omega_{s2} = \omega_{s0}$ (ω_{s0} is the anti-resonant frequency of proof mass). Then assuming zero-damping

system and solving the eigenvalue equation ($\Delta_s(\omega) = 0$) of sense-mode, the resonant frequencies of sense-mode can be obtained,

$$\omega_{sH,L} = \sqrt{\omega_{s0}^2 \pm \sqrt{\frac{nk_5^2}{m_2 m_3}}} \tag{12}$$

where $\omega_{sH,L}$ are the high and low resonant frequencies of sense-mode. The resonant peak spacing Δ_s of sense-mode are defined by,

$$\Delta_s = \omega_{sH} - \omega_{sL} \tag{13}$$

Using Equations (11), (12) and (13), the desired stiffnesses k_4 , k_5 , k_6 , can be given by,

$$\begin{cases} k_5 = \frac{1}{\sqrt{n}} \Delta_s \sqrt{m_2 m_3} \sqrt{\omega_{s0}^2 - 0.25 \Delta_s^2} \\ k_4 = m_2 \omega_{s0}^2 - k_5 \\ k_6 = m_3 \omega_{s0}^2 - n k_5 \end{cases}$$
(14)

which are the design equations of sense-mode in gyroscope array. The design equations of sense-mode are different from those of drive-mode due to the different structures of both modes. Compared with other gyroscope structures, another different point is that it utilizes several proof masses to enlarge the Coriolis force and their dynamics equations are the same as Equation (8).

Both Equations (7) and (14) are the design equations of proposed gyroscope array and the stiffnesses k_1 , k_2 , k_3 , k_4 , k_5 , k_6 , can be gained in terms of the design equations. Because the resonant peak spacing Δ_{d0} and Δ_{s0} determine the bandwidths of drive-mode and sense-mode, and the robust region is up to the central operational frequency, Δ_{d0} and Δ_{s0} should be set appropriately. Meanwhile, the anti-resonant frequencies ω_{d0} equal to ω_{s0} is the central operational frequency. Then the bandwidth of drive-mode will be highly matched with that of sense-mode, which provides the desired bandwidth of the system. At the same time, the number n of gyroscope units in the gyroscope array should be selected appropriately to ensure that the combination of gyroscope units increases the sensitivity of the system without decreasing the desired bandwidth of the system.

5. Simulation Results

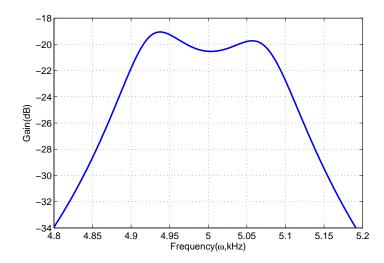
According to the theoretical analysis above, the solutions of both modes and design equations of the system are given by Equations (3), (7), (10) and (14). The values of the main parameters in the gyroscope unit and proposed gyroscope array are given in Table 1. The stiffness could be calculated by the design equations. Then the simulation will be shown in terms of these analysis.

Since the drive-modes of both gyroscope unit and gyroscope array are identical, their amplitude-frequency responses are the same. Figure 5 shows the frequency response of drive-mode. It clear illustrates a 3 dB bandwidth of about 190 Hz around the central operational frequency 5 kHz. Thus, the gyroscope unit and gyroscope array have the same bandwidth and gain in the drive-mode.

Table 1.	Parameters of th	ne gyroscope	unit and prop	osed gyroscope array.

Parameters	Values of gyroscope unit	Values of gyroscope array
m_1	$2.36e^{-7}\mathrm{kg}$	$2.36e^{-7}~\mathrm{kg}$
m_f	$5.23e^{-8} \text{ kg}$	$5.23e^{-8} { m kg}$
m_2	$2.547e^{-7} \text{ kg}$	$2.547e^{-7} \text{ kg}$
m_3	$1.35e^{-7}~\mathrm{kg}$	$8.1e^{-7} \text{ kg}$
$\omega_{s0} = \omega_{d0}$	5.0 kHz	5.0 kHz
Δ_d	150 Hz	150 Hz
Δ_s	250 Hz	100 Hz
c_1	$1e^{-4}$ Ns/m	$1e^{-4}$ Ns/m
c_2	$5e^{-6}$ Ns/m	$5e^{-6}$ Ns/m
c_3	$2e^{-4}$ Ns/m	$2e^{-4}$ Ns/m
c_4	$1e^{-4}$ Ns/m	$1e^{-4}$ Ns/m
c_5	$5e^{-6}$ Ns/m	$5e^{-6}$ Ns/m
c_6	$2e^{-4}$ Ns/m	$2e^{-4}$ Ns/m

Figure 5. Frequency response of drive-mode.



In sense-mode of the gyroscope array, the number n of gyroscope units needs to be set. Because the number n has an effect on the frequency response of sense-mode, it must be set appropriately to satisfy the desired bandwidth and gain. Figure 6 shows frequency response of sense-mode with different n. When the number is 6, the frequency response has a high gain and also changes softly, which is better than others. Figure 7 demonstrates the frequency response of gyroscope unit and gyroscope array in sense-mode. In gyroscope unit, the 3 dB bandwidth is about 300 Hz around the central operational frequency 5 kHz, and the gain is from -23 dB to -20 dB; in gyroscope array, the 3 dB bandwidth is about 100 Hz around the central operational frequency 5 kHz, and the gain is increased by 8 dB, from -15 dB to -12 dB.

Figure 6. Frequency response of gyroscope array with different number of gyroscope units in sense-mode.

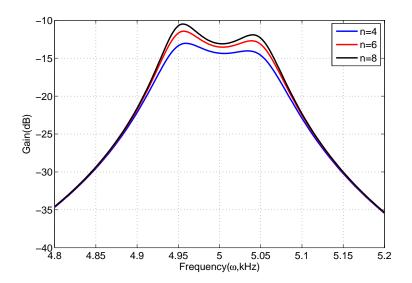
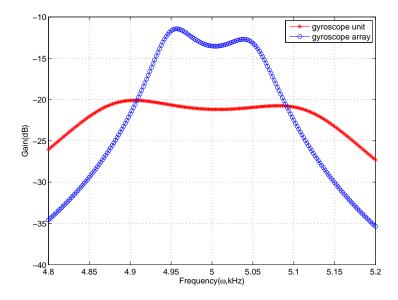


Figure 7. Frequency response of gyroscope unit and gyroscope array(n = 6) in sense-mode.



Through combining gyroscope units into a gyroscope array, the gain of sense-mode is increased and the sensitivity is improved inherently. On the other hand, the value of detection mass is also increased by 6 times, as compared to the gyroscope unit shown in Table 1. Therefore, the changing capacitance is 6 times that of the gyroscope unit, which increases the output signal and also improves the sensitivity. Although the bandwidth of gyroscope array is decreased to 100 Hz, it satisfies the need in practical applications. Meanwhile, the simulation results show that the bandwidth and central frequency of drive-mode are highly matched with those of sense-mode, which provides the system with a bandwidth of 100 Hz and improves the inherent robustness of proposed gyroscope array.

There is some tolerance in practical fabrication and the springs are susceptible to structural parameter variations. The key spring k_5 will be considered because the springs k_5 provide the link between detection mass, and all the proof masses of gyroscope units in the proposed gyroscope array as shown in

Figure 3. Therefore, we mainly discuss the effect of k_5 on the frequency response. The desired value of k_5 can be calculated by design Equation (14).

First, if the spring k_5 in one of the gyroscope units changes with $\pm 10\%$, the frequency response is shown in Figure 8. It is clear that the curves have nearly no fluctuation, which demonstrates that the gyroscope array is insensitive to one changing spring k_5 . Subsequently, all the springs k_5 are considered. Assuming that the sum of all the springs k_5 changes with $\pm 10\%$, the frequency response is shown in Figure 9. The frequency response curves change within 0.5 dB compared to the curve without error.

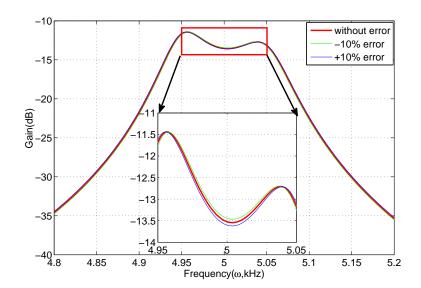
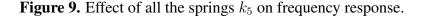
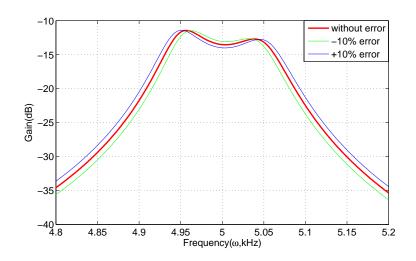


Figure 8. Effect of one of springs k_5 on frequency response.





6. Conclusion

A novel 4-DOF MEMS vibratory gyroscope unit is proposed first. The complete 2-DOF vibratory structure is utilized in both drive-mode and sense-mode, which provides the desired bandwidth of the system and improves the inherent robustness. Unlike the previous 4-DOF MEMS vibratory

gyroscope, the elements of drive-mode lay inside the schematic of the gyroscope unit and the elements of sense-mode are around drive-mode, which is for the convenience of combining some gyroscope units.

Based on the proposed gyroscope unit, a gyroscope array is designed to improve the sensitivity of the system. In gyroscope array, all the gyroscope units use the identical 2-DOF vibratory structure and there is also a proof mass in all gyroscope units. The unique detection mass is shared by all the gyroscope units so that the sensitivity of gyroscopes will be improved and the bandwidth is also adequate.

Compared to the gyroscope unit, the gain of frequency response in sense-mode of the gyroscope array (n=6) is increased by 8 dB and the changing capacitance is increased by six times, which improve the sensitivity inherently. Furthermore, the bandwidth and central frequency are highly matched with each other in both modes, which provides the system with inherent robustness. Although the bandwidth is decreased to 100 Hz, it could satisfy the need in practical applications.

Acknowledgments

This work was supported in part by the New Century Excellent Talents Support Program (NCET-11-0827) and Fundamental Research Funds for the Central Universities (HEUCFZ1110).

References

- 1. Bernstein, J.; Cho, S.; King, A.T.; Kourepenis, A.; Maciel, P.; Weinberg, M. A Micromachined Comb-Drive Tuning Fork Rate Gyroscope. In *Proceedings of the IEEE Micro Electro Mechanical Systems—An Investigation of Micro Structures, Sensors, Actuators, Machines and Systems*, Fort Lauderdale, FL, USA, 7–10 Febuary 1993; pp. 143–148.
- 2. Alper, S.E.; Azgin, K.; Akin, T. A high-performance sillicon-on-insulator MEMS gyroscope operating at atmospheric pressure. *Sens. Actuators A Phys.* **2007**, *135*, 34–42.
- 3. Ding, H.T.; Liu, X.S.; Lin, L.T.; Chi, X.Z.; Cui, J.; Kraft, M.; Yang, Z.C.; Yan, G.Z. A high-resolution silicon-on-glass z axis gyroscope operating at atmospheric pressure. *IEEE Sens. J.* **2010**, *10*, 1066–1074.
- 4. Wang, R.; Cheng, P.; Xie, F.; Young, D.; Hao, Z.L. A multiple-beams tuning-fork gyroscope with high quality factors. *Sens. Actuators A Phys.* **2011**, *166*, 22–33.
- 5. Riaz, K.; Bazaz, S.A.; Saleem, M.M.; Shakoor, R.I. Design, damping estimation and experimental characterization of decoupled 3-DoF robust MEMS gyroscope. *Sens. Actuators A Phys.* **2011**, *172*, 523–532.
- 6. Acar, C.; Shkel, A.M. Inherently robust micromachined gyroscopes with 2-DOF sense-mode oscillator. *J. Microelectromech. Syst.* **2006**, *15*, 380–387.
- 7. Schofield, A.R.; Trusov, A.A.; Shkel, A.M. Effects of operational frequency scaling in multi-degree of freedom MEMS gyroscopes. *IEEE Sens. J.* **2008**, *8*, 1672–1680.
- 8. Sahin, K.; Sahin, E.; Alper, S.; Akin, T. A wide-bandwidth and high-sensitivity robust microgyroscope. *J. Micromech. Microeng.* **2009**, *19*, 1722–1726.
- 9. Trusov, A.A.; Schofield, A.R.; Shkel, A.M. Performance characterization of a new temperature-robust gain-bandwidth improved MEMS gyroscope operated in air. *Sens. Actuators A Phys.* **2009**, *155*, 16–22.

10. Schofield, A.R.; Trusov, A.A.; Shkel, A.M. Micromachined gyroscope concept allowing interchangeable operation in both robust and precision modes. *Sens. Actuators A Phys.* **2011**, *165*, 35–42.

- 11. Wang, W.; Lv, X.Y.; Sun, F. Design and simulation of micromachined vibratory gyroscope with two degree-of-freedom drive-mode and sense-mode. *IEEE Sens. J.* **2012**, *12*, 2460–2464.
- 12. Liu, D.C.; Lu, N.N.; Cui, J. Digital Closed-Loop Control Based on Adaptive Filter for Drive Mode of a MEMS Gyroscope. In *Proceedings of the IEEE Sensors*, Kona, HI, USA, 1–4 November 2010; pp. 1722–1726.
- 13. Luo, B.; Wang, A.C.; Wu, M.P. A drive control scheme based on phase-control for silicon micromechanical gyroscopes. *Acta Automatica Sinica* **2012**, *38*, 206–212.
- © 2013 by the authors; licensee MDPI, Basel, Switzerland. This article is an open access article distributed under the terms and conditions of the Creative Commons Attribution license (http://creativecommons.org/licenses/by/3.0/).