

Article



Modelling and Daisy Chaining Control Allocation of a Multirotor Helicopter with a Single Tilting Rotor

Robert Porter *, Bijan Shirinzadeh and Man Choi

Robotics and Mechatronics Research Laboratory, Department of Mechanical and Aerospace Engineering, Monash University, Clayton, VIC 3800, Australia; bijan.shirinzadeh@monash.edu (B.S.); man.choi@monash.edu (M.C.)

* Correspondence: robert.porter@monash.edu; Tel.: +61-399-053-545

Academic Editor: Sergio Montenegro Received: 7 October 2016; Accepted: 15 November 2016; Published: 23 November 2016

Abstract: This paper presents the development and implementation of a single tilting rotor multirotor helicopter. A single tilting rotor multirotor helicopter is proposed that allows for decoupled lateral acceleration and attitude states. A dynamics model of the proposed multirotor helicopter is established to enable control system development. A control system architecture and daisy chaining-based control allocation scheme is developed and implemented. The control architecture facilitates the control of decoupled lateral accelerations and attitudes. Further, a computational and experimental analysis is undertaken and offers evidence that the proposed multirotor helicopter and control system architecture enables the multirotor helicopter to achieve lateral accelerations without requiring attitude actuation.

Keywords: unmanned aerial vehicles; multirotor helicopter; tilt-rotor; control allocation

1. Introduction

Unmanned aerial vehicles have rapidly gained in popularity in recent years [1–3]. In particular, there has been significant interest in unmanned multirotor helicopters [4–11]. This is primarily due to the wide range of applications and to their mechanical simplicity and durability. However, the mechanical simplicity of multirotor helicopters does come at a cost: traditional multirotor helicopters are underactuated. In particular, only four of the six degrees-of-freedom are controllable. Multirotor helicopters are capable of tracking desired attitudes, headings and accelerations in the body-fixed vertical direction. Multirotor helicopters cannot achieve accelerations in the body-fixed horizontal plane. This results in a coupling between multirotor helicopters' attitude and acceleration. This underactuation places fundamental restrictions on the ability to perform tasks requiring manipulation of the surrounding environment. It also necessitates the use of complex mechanical componentry if a payload is to remain at a given attitude independent of the desired movement of the multirotor helicopter.

In an effort to overcome this limitation, several approaches have been proposed. An aerial vehicle with three ducted fans mounted on gimbals is under investigation by Jayakody [12] and Yuan [13]. The gimbals allow the thrust produced by the ducted fans to be directed with respect to the body of the aircraft, allowing for horizontal accelerations independent of attitude. Another approach has been presented by Cetinsoy et al. [14]. Their platform consists of a multirotor helicopter with sub-rotor control surfaces. The control surfaces allow the down-wash from the rotors to be directed, thus producing a lateral component in the resultant force. Further, tilt-rotor actuation has been investigated by Kendoul et al. [15] and Ryll et al. [16]. Tilt-rotor actuation involves mounting the rotors of traditional multirotor helicopters on gimbals, thus allowing the resultant thrust to be vectored as in the approach by Jayakody and Yuan [12,13]. Finally, helicopters utilise swash-plates to allow cyclic-pitch rotor

blade actuation [17]. The cyclic-pitch of a rotor affects the moment produced by the rotor about axes perpendicular to the rotor's axis of rotation. The cyclic-pitch further causes the thrust produced by the rotor to be vectored with respect to the body of the multirotor helicopter. Therefore, cyclic-pitch allows lateral accelerations to be achieved. However, changes to a rotor's cyclic-pitch induce a secondary change in attitude towards the direction of thrust produced.

Drawing inspiration from research into tilt-rotor-equipped multirotor helicopters, in this paper, a minimal actuation concept that aims to achieve the same result is proposed. Rather than allowing all rotors to tilt, only a single tilting rotor is included on an otherwise traditional multirotor helicopter. The proposed aircraft is mechanically simpler than a tilt-rotor multirotor helicopter, as only a single additional actuator is required. The proposed multirotor helicopter can achieve control of five degrees-of-freedom; allowing for horizontal accelerations that are independent of attitude. To achieve this, a daisy chaining-based control allocation methodology is proposed. This control allocation methodology takes full advantage of the proposed multirotor helicopter by prioritising maintaining the desired attitude when manoeuvring and utilising changes in attitude only when absolutely necessary.

The remainder of this paper is structured as follows. In Section 2, the dynamics model of the single tilt-rotor multirotor helicopter is developed. An analysis is conducted to determine the available control bandwidth in Section 3. The proposed daisy chaining-based control allocation methodology is detailed in Section 4. Computational and experimental analyses of the single tilt-rotor multirotor helicopter and control system architecture are presented in Sections 5 and 6, respectively. Finally, concluding remarks are given in Section 7.

2. Multirotor Helicopter Dynamics Model

The multirotor helicopter proposed in this research consists of two rigid bodies: the multirotor helicopter body, *B*, which includes the non-tilting rotor groups; and the tilting rotor group, *P*. The tilting rotor group includes the gimbal, the motor driving the tilting rotor and the rotor itself. The tilting rotor is constrained to only rotate about the arm to which it is attached. A schematic of the proposed multirotor helicopter is presented in Figure 1. In this section, the dynamics model of this multirotor helicopter is developed. The nomenclature utilised in this section is presented in Table 1.



Figure 1. The proposed single tilt-rotor multirotor helicopter.

Symbol	Description		
В	Multirotor helicopter body		
F_B	Coordinate system attached to B		
$F_{B_{O}}$	Origin of F_B		
F_E	Earth-fixed coordinate system		
F_P	Coordinate system attached to P		
F_{P_O}	Origin of F_P		
g	Gravity vector in F_E		
I_B	Moment of inertia matrix of the multirotor helicopter		
I_{R_0}	Moment of inertia matrix of the tilting rotor		
I_{R_i}	Moment of inertia matrix of the <i>i</i> -th rotor		
k_D	Trajectory PID control derivative action gain		
k_I	Trajectory PID control integral action gain		
k_P	Trajectory PID control proportional action gain		
$k_{T,i}$	Rotor thrust constant		
$k_{\tau,i}$	Rotor torque constant		
L_{R}^{B}	Vector from F_B to F_P		
$L_{P_i}^{\scriptscriptstyle B}$	Vector from F_B to the <i>i</i> -th rotor		
1	Distance from F_{B_O} to the rotor hubs		
т	Mass of the multirotor helicopter		
P	Tilting rotor group		
p	Position of F_B in F_E		
q_{R}^{B}	Orientation of F_P with respect to F_B		
q_B^E	Orientation of F_B with respect to F_E		
T_0	Thrust of the tilting rotor in F_P		
T_1	Attitude PD^2 control gain		
T_2	Attitude PD^2 control gain		
T_3	Attitude PD ² control gain		
\boldsymbol{u}_T	Thrust virtual control inputs		
$u_{ au}$	Torque virtual control inputs		
β	Tilt of <i>P</i> with respect to <i>B</i>		
ω_0	Angular velocity of the tilting rotor in F_P		
ω_0	Angular velocity of the tilting rotor blade about F_{P_Z}		
ω_i	Angular velocity of the <i>i</i> -th rotor about F_{B_z}		
ω_i	Angular velocity of the <i>i</i> -th rotor in F_E		
ω_{B}	Angular velocity of F_B with respect to F_E		
ω_B^r	Angular velocity of F_B with respect to F_P		
$ au_0$	Iorque produced by the tilting rotor		
$ au_i$	Iorque produced by the <i>t</i> -th rotor		
$ au_{ext_0}$	External torque applied to the tilting rotor		
$ au_{ext_i}$	External torque applied to the <i>i</i> -th rotor		

Table 1. Nomenclature.

A right-hand coordinate frame F_B , with origin F_{B_O} , is attached to the centre of *B* as in Figure 1. Additionally, a coordinate frame F_P is attached to the tilting rotor group. The origin of F_P , F_{P_O} , is positioned at the intersection of the axis about which the rotor group rotates and the axis about which the rotor itself rotates. Without loss of generality, it is assumed that the tilting rotor is positioned along the *x* axis of F_B . An inertial coordinate frame is defined as F_E . The orientation of F_B with respect to F_E is given by the attitude quaternion, q_B^E . Likewise, the orientation of F_P with respect to F_B is given by q_B^P .

The following assumptions are made:

- The multirotor helicopter body is rigid and symmetrical
- $F_{B_{\Omega}}$ and the centre of mass of the multirotor helicopter coincide
- The rotors are rigid
- The multirotor helicopter is operating in the vicinity of the hover condition
- The aerodynamic drag force on the multirotor helicopter body is negligible

Under these assumptions, the dynamics model of the multirotor helicopter can be derived using the Newton–Euler formalism.

The angular velocity of the tilting rotor in F_P , ω_0 , is given by:

$$\begin{pmatrix} 0 & \boldsymbol{\omega}_0 \end{pmatrix} = \overline{q}_P^B \otimes \begin{pmatrix} 0 & \boldsymbol{\omega}_B \end{pmatrix} \otimes q_P^B + \begin{pmatrix} 0 & \left(\dot{\boldsymbol{\beta}} & 0 & \boldsymbol{\omega}_0 \right) \end{pmatrix}$$
(1)

where \overline{q} indicates the conjugate of quaternion q; ω_B is the angular velocity of F_B with respect to F_E ; β is the angle between F_{B_z} and F_{P_z} (the subscript z indicates the z axis); and ω_0 is the angular velocity of the tilting rotor about F_{P_z} . Differentiating this expression results in the angular acceleration of the tilting rotor in F_P , which is given in the following equation.

$$\begin{pmatrix} 0 & \dot{\omega}_0 \end{pmatrix} = \begin{pmatrix} 0 & \omega_B^P \times \omega_B \end{pmatrix} + \overline{q}_P^B \otimes \begin{pmatrix} 0 & \dot{\omega}_B \end{pmatrix} \otimes q_P^B + \begin{pmatrix} 0 & \left(\ddot{\beta} & 0 & \dot{\omega}_0 \right) \end{pmatrix}$$
(2)

In Equation (2), ω_B^P is the angular velocity of F_B with respect to F_P . The torque produced by the tilting rotor, τ_0 , is given by:

$$\boldsymbol{\tau}_0 = I_{R_0} \dot{\boldsymbol{\omega}}_0 + \boldsymbol{\omega}_0 \times I_{R_0} \boldsymbol{\omega}_0 - \boldsymbol{\tau}_{ext_0} \tag{3}$$

where τ_{ext_0} is the external torque applied to the tilting rotor. τ_{ext_0} is primarily due to a counter-rotating torque about F_{P_z} caused by the air drag [17] and is typically considered to be of the following form in normal operating conditions (see [8,18,19], for example),

$$\boldsymbol{\tau}_{ext_i} = \begin{pmatrix} 0 & 0 & -k_{\tau,i}\omega_i \|\omega_i\| \end{pmatrix}^T$$
(4)

where the subscript *i* indicates the rotor number; and k_{τ} is a constant determined by the rotor size and profile.

Similarly, the torque produced by each of the non-tilting rotors can be expressed as:

$$\boldsymbol{\tau}_i = I_{R_i} \dot{\boldsymbol{\omega}}_i + \boldsymbol{\omega}_i \times I_{R_i} \boldsymbol{\omega}_i - \boldsymbol{\tau}_{ext_i} \quad i \in 1, 2, 3$$
(5)

where:

$$\boldsymbol{\omega}_i = \boldsymbol{\omega}_B + \begin{pmatrix} 0 & 0 & \omega_i \end{pmatrix}^T . \tag{6}$$

The torque on the multirotor helicopter in F_B , τ_B , can be described as:

$$\begin{pmatrix} 0 & \boldsymbol{\tau}_B \end{pmatrix} = \begin{pmatrix} 0 & I_B \dot{\boldsymbol{\omega}}_B + \boldsymbol{\omega}_B \times I_B \boldsymbol{\omega}_B \end{pmatrix} + q_P^B \otimes \begin{pmatrix} 0 & \boldsymbol{\tau}_0 \end{pmatrix} \otimes \overline{q}_P^B + \sum_{i=1}^3 \begin{pmatrix} 0 & \boldsymbol{\tau}_i \end{pmatrix} .$$
(7)

The torques produced by the multirotor helicopter are due to the thrusts produced by each of the rotors and are given by:

$$\begin{pmatrix} 0 & \boldsymbol{\tau}_B \end{pmatrix} = \begin{pmatrix} 0 & \boldsymbol{L}_P^B \end{pmatrix} \times \boldsymbol{q}_P^B \otimes \begin{pmatrix} 0 & \boldsymbol{T}_0 \end{pmatrix} \otimes \overline{\boldsymbol{q}}_P^B + \sum_{i=1}^3 \begin{pmatrix} 0 & \boldsymbol{L}_{P_i}^B \times \boldsymbol{T}_i \end{pmatrix}$$
(8)

where L_P^B is the vector from F_B to F_P ; T_0 is the thrust of the tilting rotor in F_P ; $L_{P_i}^B$ is the vector from F_B along each of the arms of the multirotor helicopter to the axis of rotation of the *i*-th rotor; and T_i is the thrust vector of the *i*-th rotor in F_B . It should be noted that, as for Equation (4), it is typical to assume T takes the following form for a fixed-pitch rotor in normal operating conditions,

$$\boldsymbol{T}_{i} = \begin{pmatrix} 0 & 0 & k_{T,i}\omega_{i}^{2} \end{pmatrix}^{T}$$
(9)

where $k_{T,i}$ is a constant determined by the rotor size and profile.

The translational dynamics of the multirotor helicopter are derived using a force balance and are presented in the following equation.

$$m\begin{pmatrix} 0 & \ddot{p} \end{pmatrix} = m\begin{pmatrix} 0 & g \end{pmatrix} + \bar{q}_B^E \otimes \left(\bar{q}_P^B \otimes \begin{pmatrix} 0 & T_0 \end{pmatrix} \otimes q_P^B + \sum_{i=1}^3 \begin{pmatrix} 0 & T_i \end{pmatrix} \right) \otimes q_B^E$$
(10)

In Equation (10), *m* is the mass of the multirotor helicopter; *p* is the position vector of F_B relative to F_E ; and *g* is the gravity vector in F_E .

The complete dynamics model of the multirotor helicopter with single tilting rotor is given by Equations (1) to (10).

3. Control Bandwidth Analysis

One of the primary constraints when implementing a single tilting rotor is that the lateral thrust produced by the tilting rotor cannot be balanced by vectoring thrust produced by another rotor. Instead, the resulting moment about F_{B_z} must be balanced by torque produced by the non-tilting rotors. As the torque produced by the rotors is an order of magnitude less than the thrust produced, this constraint requires careful consideration.

To characterise the available control bandwidth, the multirotor helicopter dynamics developed in the previous section can be simplified into Equations (11) to (16).

$$\left(I_B + \sum_{i=0}^3 I_{R_i}\right)\dot{\omega}_B = -\omega_B \times I_B \omega_B - G_\tau + u_\tau$$
(11)

$$m\begin{pmatrix} 0 & \mathbf{\ddot{p}} \end{pmatrix} = m\begin{pmatrix} 0 & \mathbf{g} \end{pmatrix} + \overline{q}_{B}^{E} \otimes \begin{pmatrix} 0 & \mathbf{u}_{T} \end{pmatrix} \otimes q_{B}^{E}$$
(12)

$$\begin{pmatrix} 0 & \boldsymbol{G}_{\tau} \end{pmatrix} = q_{P}^{B} \otimes \begin{pmatrix} 0 & \boldsymbol{G}_{\tau}' \end{pmatrix} \otimes \overline{q}_{P}^{B} + \begin{pmatrix} 0 & \boldsymbol{G}_{\tau}'' \end{pmatrix}$$
(13)

$$\boldsymbol{G}_{\tau}^{\prime} = \boldsymbol{I}_{R_{0}} \begin{pmatrix} \boldsymbol{\omega}_{B}^{P} \times \boldsymbol{\omega}_{B} + \begin{pmatrix} \ddot{\boldsymbol{\beta}} & 0 & \dot{\boldsymbol{\omega}}_{0} \end{pmatrix}^{T} \end{pmatrix} + \boldsymbol{\omega}_{0} \times \boldsymbol{I}_{R_{0}} \boldsymbol{\omega}_{0}$$
(14)

$$\boldsymbol{G}_{\tau}^{\prime\prime} = \sum_{i=1}^{3} (-1)^{i} \left(I_{R_{i}} \begin{pmatrix} 0 & 0 & \dot{\omega}_{i} \end{pmatrix}^{T} + \boldsymbol{\omega}_{i} \times I_{R_{i}} \boldsymbol{\omega}_{i} \right)$$
(15)

$$\begin{pmatrix} \boldsymbol{u}_{T} \\ \boldsymbol{u}_{\tau} \end{pmatrix} = \begin{bmatrix} 0 & 0 & 0 & 0 \\ k_{T}\sin(\beta) & 0 & 0 & 0 \\ k_{T}\cos(\beta) & k_{T} & k_{T} & k_{T} \\ 0 & -lk_{T} & 0 & lk_{T} \\ lk_{T}\cos(\beta) - k_{\tau}\sin(\beta) & 0 & -lk_{T} & 0 \\ lk_{T}\sin(\beta) + k_{\tau}\cos(\beta) & -k_{\tau} & k_{\tau} & -k_{\tau} \end{bmatrix} \begin{pmatrix} \omega_{0}^{2} \\ \omega_{1}^{2} \\ \omega_{2}^{2} \\ \omega_{3}^{2} \end{pmatrix}$$
(16)

In (16), l is the distance from F_{B_0} to the axis each rotor rotates about; β and w_i^2 $i \in [0,3]$ are the control inputs, and u_T and u_τ are the virtual control inputs to the dynamics system. The process for setting the control inputs given a set of virtual control inputs is referred to as control allocation. Adding the virtual control input abstraction allows for simpler outer loop controller design, as the details of control allocation do not have to be considered.

The control bandwidth can be characterised by an analysis of Equation (16). In particular, given control input constraints, the maximum and minimum achievable lateral forces, u_{T_y} , that can be generated can be determined as a function of required vertical force, u_{T_z} , and torques, u_{τ} . The results presented in Figures 2 to 4 were calculated by setting u_{T_z} and u_{τ} and increasing or decreasing u_{T_y} until the system could no longer be solved to satisfy the control input constraints. The data points represent the maximum and minimum values of u_{T_y} for a given set of u_{T_z} and u_{τ} . The parameters k_T , k_{τ} and l are presented in Table 2, along with the control input constraints and additional multirotor helicopter

parameters. It should further be noted that these results are valid for operating conditions sufficiently close to the hover condition where assumptions made in Equations (4) and (9) are valid.

I_{R_0} (Nm·s)	$I_{B_{xx}}$ (Nm·s)	$I_{B_{yy}}$ (Nm·s)	$I_{B_{zz}}$ (Nm·s)
$5.824 imes 10^{-5}$	$1.773 imes 10^{-2}$	1.880×10^{-2}	3.461 × 10 ⁻²
<i>m</i> (kg)	<i>l</i> (m)	$k_{ au}$ (Nm·s ²)	k_T (N·s ²)
2.168	0.225	$1.954 imes 10^{-5}$	$4.389 imes 10^{-7}$
ω_{\min} (radian/s) 52.4	ω _{max} (radian/s)	$eta_{ extsf{min}}$ (radian)	β_{max} (radian)
	890.1	-1.57	1.57

Table 2. Physical parameters for the single tilt-rotor multirotor helicopter.



Figure 2. The maximum and minimum values of u_{T_y} that can be achieved given a set of u_{T_z} and u_{τ} . Note that $u_{\tau_y} = u_{\tau_z} = 0$ (Nm).



Figure 3. The maximum and minimum values of u_{T_y} that can be achieved given a set of u_{T_z} and u_{τ} . Note that $u_{\tau_x} = u_{\tau_z} = 0$ (Nm).



Figure 4. The maximum and minimum values of u_{T_y} that can be achieved given a set of u_{T_z} and u_{τ} . Note that $u_{\tau_x} = u_{\tau_y} = 0$ (Nm).

To demonstrate how the addition of a single tilting rotor affects the control bandwidth of the multirotor helicopter, consider Figure 2. In this case, $u_{\tau_y} = u_{\tau_z} = 0$ Nm. Consider a case where u_{T_z} and u_{T_y} are initially 40 N and 0 N, respectively. Referring to the figure, it is observed that under these conditions, the multirotor helicopter is able to generate ± 2 Nm of torque about the *x* axis. If the desired u_{T_y} is increased to 0.5 N and the other desired control inputs remain the same, the torque the multirotor helicopter can generate is reduced to ± 1 Nm. If the desired u_{T_y} is increased further to 1 N, the magnitude of torque the multirotor helicopter can generate is reduced to <0.5 Nm. If u_{T_y} continues to be increased, the system cannot be solved for any value of u_{τ_x} once u_{T_y} reaches 1.3 N. In this case, to solve for the desired values u_{T_y} and u_{τ} , u_{T_z} has to be decreased.

4. Control System Architecture

The objective of the control system in this research is to maintain the multirotor helicopter at a level attitude while allowing horizontal accelerations. However, the control system should also be capable of changing the attitude of the multirotor helicopter in response to external disturbances or excessive commanded horizontal accelerations. To this end, a daisy chain control methodology is implemented. This method assumes a hierarchy of control effectors and does not utilise an available control effector until all higher priority control effectors become saturated [20]. For this research, the tilting rotor will be assigned the highest priority when tracking commanded horizontal accelerations. The attitude of the multirotor helicopter will only be affected when the tilting rotor cannot supply the required control bandwidth.

4.1. Position Control

To calculate the translational virtual control inputs, u_T , a proportional-integral-derivative (PID) controller is utilised as follows,

$$\boldsymbol{u}_{\text{PID}} = k_P(\boldsymbol{p} - \boldsymbol{p}_d) + k_I \int_0^t (\boldsymbol{p} - \boldsymbol{p}_d) \, \mathrm{d}t - k_D \dot{\boldsymbol{p}}_d + \boldsymbol{u}_{\text{FF}}$$
(17)

where u_{PID} is the input in F_E from the PID controller; k_P , k_I and k_D are the proportional, integral and derivative gains, respectively; and u_{FF} is the feed-forward control input and is given by:

$$u_{\rm FF} = g \ . \tag{18}$$

 u_T can then be calculated by transforming u_{PID} into F_B using the current multirotor helicopter attitude and heading as follows:

$$\begin{pmatrix} 0 & \boldsymbol{u}_T \end{pmatrix} = q_B^E \otimes \begin{pmatrix} 0 & \boldsymbol{u}_{\text{PID}} \end{pmatrix} \otimes \overline{q}_B^E \,. \tag{19}$$

4.2. Control Allocation and Attitude Control

Having calculated u_T , the tilting rotor tilt and the rotor angular velocities must be calculated. From Equation (16), it can be seen that the first element of u_T cannot be non-zero. Therefore, it is omitted from the following control allocation, and u_T is replaced with $u'_T = \begin{pmatrix} u_{T_y} & u_{T_z} \end{pmatrix}^T$. To map the virtual control inputs to the physical control inputs, Equation (16) is linearised about β_0 , which results in:

$$\begin{pmatrix} \boldsymbol{u}_{T}' \\ \boldsymbol{u}_{\tau} \end{pmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{pmatrix} \omega_{0}^{2} \delta \beta \\ \omega_{0}^{2} \\ \omega_{1}^{2} \\ \omega_{2}^{2} \\ \omega_{3}^{2} \end{pmatrix}$$
(20)

where:

$$\begin{split} A &= \begin{bmatrix} k_T \cos(\beta_0) & k_T \sin(\beta_0) \\ -k_T \sin(\beta_0) & k_T \cos(\beta_0) \end{bmatrix}; \\ B &= \begin{bmatrix} 0 & 0 & 0 \\ k_T & k_T & k_T \end{bmatrix}; \\ C &= \begin{bmatrix} 0 & 0 & 0 \\ -lk_T \sin(\beta_0) - k_\tau \cos(\beta_0) & lk_T \cos(\beta_0) - k_\tau \sin(\beta_0) \\ lk_T \cos(\beta_0) - k_\tau \sin(\beta_0) & lk_T \sin(\beta_0) + k_\tau \cos(\beta_0) \end{bmatrix}; \\ D &= \begin{bmatrix} -lk_T & 0 & lk_T \\ 0 & -lk_T & 0 \\ -k_\tau & k_\tau & -k_\tau \end{bmatrix}; \end{split}$$

and $\delta\beta$ is some perturbation in β away from β_0 .

As the multirotor helicopter is operating at about a zero attitude, it is initially assumed that u_{τ} is negligible. This assumption allows Equation (20) to be reduced as follows,

$$\boldsymbol{u}_{T}^{\prime} = \begin{bmatrix} A - BD^{-1}C \end{bmatrix} \begin{pmatrix} \omega_{0}^{2}\delta\beta \\ \omega_{0}^{2} \end{pmatrix} = \begin{bmatrix} A_{11}^{\prime} & A_{12}^{\prime} \\ A_{21}^{\prime} & A_{22}^{\prime} \end{bmatrix} \begin{pmatrix} \omega_{0}^{2}\delta\beta \\ \omega_{0}^{2} \end{pmatrix}$$
(21)

where:

$$\begin{aligned} A'_{11} &= k_T \cos(\beta_0); \\ A'_{12} &= k_T \sin(\beta_0); \\ A'_{21} &= \left(\frac{k_T^2 l}{k_\tau} - \frac{2k_\tau}{l}\right) \cos(\beta_0) - 4k_T \sin(\beta_0); \\ A'_{22} &= \left(\frac{k_T^2 l}{k_\tau} - \frac{2k_\tau}{l}\right) \sin(\beta_0) + 4k_T \cos(\beta_0) \,. \end{aligned}$$

 $\delta\beta$ and a preliminary estimate of ω_0^2 are therefore given by Equations (22) and (23), respectively.

$$\delta\beta = \frac{A'_{22}u_{T_y} - A'_{12}u_{T_z}}{A'_{11}u_{T_z} - A'_{21}u_{T_y}}$$
(22)

$$\omega_{0,\,\text{estimate}}^{2} = \frac{1}{A_{11}^{\prime}A_{22}^{\prime} - A_{12}^{\prime}A_{21}^{\prime}} (A_{11}^{\prime}\boldsymbol{u}_{T_{z}} - A_{21}^{\prime}\boldsymbol{u}_{T_{y}})$$
(23)

It should be noted that the denominator in Equation (23) is non-zero, and the denominator in Equation (22) is only zero if $\omega_{0, \text{ estimate}}^2$ is zero.

An estimate of the achievable horizontal control input can then be calculated as:

$$\boldsymbol{u}_{T_y, \text{ estimate}} = k_T \sin(\beta_1) \omega_{0, \text{ estimate}}^2$$
(24)

where:

$$\beta_1 = \beta_0 + \delta\beta;$$

and it is enforced that $\beta_{\min} \leq \beta_1 \leq \beta_{\max}$ and $\omega_{0,\min} \leq \omega_{0,\text{ estimate}} \leq \omega_{0,\max}$.

Applying the daisy chain principle, the residual of the required control input in F_E , $u_{residual}$ is calculated as:

$$\begin{pmatrix} 0 & \boldsymbol{u}_{\text{residual}} \end{pmatrix} = \begin{pmatrix} 0 & \boldsymbol{u}_{\text{PID}} \end{pmatrix} - \overline{q}_B^E \otimes \begin{pmatrix} 0 & \begin{pmatrix} 0 & \boldsymbol{u}_{T_y, \text{ estimate}} & 0 \end{pmatrix} \end{pmatrix} \otimes q_B^E.$$
(25)

To achieve the residual of the desired control input, the attitude of the multirotor helicopter must be utilised. To this end, the minimum angle rotation to achieve u_{residual} , $q_{B,\text{desired}}^{E}$ is calculated as follows [21],

$$q_{B,\text{ desired}}^{E} = \frac{\left(1 + \hat{\boldsymbol{u}}_{\text{residual}}^{T} \quad \hat{\boldsymbol{u}}_{\text{residual}} \times T\right)}{\left\|\left(1 + \hat{\boldsymbol{u}}_{\text{residual}}^{T} \quad \hat{\boldsymbol{u}}_{\text{residual}} \times T\right)\right\|_{2}};$$
(26)

where \hat{u} indicates the unit vector of vector u; and:

$$T = \begin{pmatrix} 0 & 0 & 1 \end{pmatrix}^T$$
.

The following equation can then be used to find the error between the current attitude and the desired attitude, $q_{B, \text{ error}}^E$:

$$q_{B,\,\text{error}}^E = q_{B,\,\text{desired}}^E \otimes \overline{q}_B^E \,. \tag{27}$$

Further, the time-derivative of $q_{B, \text{ error}}^E$ is calculated as:

$$\dot{q}_{B,\,\mathrm{error}}^{E} = \frac{1}{2} q_{B,\,\mathrm{error}}^{E} \otimes \begin{pmatrix} 0 & \boldsymbol{\omega}_{B} \end{pmatrix} \,. \tag{28}$$

The attitude is regulated utilising a PD² feedback control proposed by Tayebi et al. [8]. The two derivative actions act on the body angular velocity and the derivative of the attitude error. The controller also includes a feed-forward term to negate the Coriolis torque. The PD² control is:

$$\boldsymbol{u}_{\tau} = \boldsymbol{\omega}_{B} \times I_{B} \boldsymbol{\omega}_{B} - (T_{3} + T_{2}T_{1}) \boldsymbol{q}_{B,\,\mathrm{error}}^{E} - T_{2} \boldsymbol{\omega}_{B} - I_{B} T_{1} \dot{\boldsymbol{q}}_{B,\,\mathrm{error}}^{E}$$
(29)

where T_1 is a 3 × 3 symmetric positive definite matrix; and T_2 and T_3 are 3 × 3 diagonal positive definite matrices.

Finally, as u_{T_z} , u_{τ} and $\delta\beta$ have been calculated, the rotor angular velocities can be calculated as follows:

$$\begin{pmatrix} \omega_0^2 \\ \omega_1^2 \\ \omega_2^2 \\ \omega_3^2 \end{pmatrix} = (E^T W E)^{-1} E^T W \begin{pmatrix} \boldsymbol{u}_T' \\ \boldsymbol{u}_T \end{pmatrix}$$
(30)

where W is a diagonal, positive-definite, pseudo-inversion weight matrix; and:

$$E = \begin{bmatrix} k_T \sin(\beta_1) & 0 & 0 & 0 \\ k_T \cos(\beta_1) & k_T & k_T & k_T \\ 0 & -lk_T & 0 & lk_T \\ lk_T \cos(\beta_1) - k_\tau \sin(\beta_1) & 0 & -lk_T & 0 \\ lk_T \sin(\beta_1) + k_\tau \cos(\beta_1) & -k_\tau & k_\tau & -k_\tau \end{bmatrix},$$

thus completing the feedback control for the single tilt-rotor multirotor helicopter.

5. Computational Analysis

To characterise the performance of the single tilt-rotor multirotor helicopter, a computational model was developed. The parameters of the multirotor helicopter are presented in Table 2. In this computational analysis the dynamics model is updated at a rate of 10 kHz, and the control is performed at 200 Hz.

Initially, the multirotor helicopter is commanded to move in the *y*-direction at a rate of 1 m/s. After a period of 10 s, the multirotor helicopter is commanded to increase its velocity to 5 m/s. The complete motion is presented in Figure 5. The attitude of the multirotor helicopter during the manoeuvre is presented in Figure 6. Further, Figure 7 shows the achieved lateral force and the corresponding rotor tilt.

From Figure 7, it is observed that the entire desired lateral force for the initial stage of the manoeuvre can be achieved through actuating the tilting rotor only. This is reflected in Figure 6, as the roll remains unchanged despite the multirotor helicopter moving in the desired direction. During the second stage of the manoeuvre, it can be seen that actuating the only tilting rotor no longer produces the desired lateral force. Instead, the attitude of the multirotor helicopter must be utilised to achieve the desired motion. This is represented in Figure 6 as a change in the roll, ϕ , of the multirotor helicopter.



Figure 5. Multirotor helicopter position.



Figure 7. Lateral thrust produced by the multirotor helicopter and tilt-rotor tilt angle.

6. Experimental Analysis

To validate the results obtained from the computational analysis in the previous section, an experimental analysis has been undertaken. To this end, an experimental research facility has been established. The research facility incorporates a single tilt-rotor multirotor helicopter. The multirotor helicopter is presented in Figure 8. The remainder of the research facility has been presented in previous work by the author [22] and includes a laser interferometry-based sensing and measurement unit to precisely record the position of the multirotor helicopter at a rate of 1000 Hz. Measurements from accelerometers, gyroscopes and magnetometers mounted onboard the multirotor helicopter are recorded at a rate of 200 Hz. The algorithm proposed in [23] is utilised to combine these sensor measurements to produce attitude estimates. The desired rotor angular velocities and rotor tilt angle are also updated a rate of 200 Hz. The PID and PD² controller gains in Equations (17) and (29), respectively, were calculated utilising the Ziegler–Nichols step-response tuning methodology [24]. This methodology was chosen as the same set of tuning rules could be applied to multirotor helicopters with different rotor configurations, removing the effect of controller tunings from any comparisons drawn.



Figure 8. Single tilt-rotor multirotor helicopter in flight.

In the first experiment, the multirotor helicopter was commanded to move in the *y*-direction at 0.8 m/s. The measured position and estimated attitude of the multirotor helicopter during this manoeuvre are presented in Figures 9 and 10, respectively. The experiment was repeated with the same multirotor helicopter with all four rotors fixed in the traditional positions, the results of which are also presented in the figures.

0.2

0.1

0

4

0

1

-0.1

X (m)





Figure 9. Multirotor helicopter position throughout the first manoeuvre.



Figure 10. Multirotor helicopter attitude throughout the first manoeuvre.

Referring to Figure 9, it is observed that the differences between the trajectories for the multirotor helicopter with a single tilt-rotor configuration and the fixed-rotor configuration are minimal. The most notable difference is the drift in the *x*-direction of the single tilt-rotor multirotor helicopter, which differs from the desired trajectory by 0.08 m. The corresponding drift of the fixed-rotor multirotor helicopter is only 0.015 m. The differences between the performance of the multirotor helicopter configurations is observed in Figure 10. Here, the fixed-rotor multirotor helicopter is required to roll up to 4° to achieve the desired trajectory. The multirotor helicopter in the single tilt-rotor configuration exhibits a maximum roll of only 0.5°.

In the second experiment, the multirotor helicopter was commanded to track a step input of 1 m in the *y*-direction. The position and attitude of the multirotor helicopter during the second manoeuvre are presented in Figures 11 and 12, respectively. As in the previous case, the differences in trajectories for the fixed and tilt-rotor are minimal. Further, it is observed that both multirotor helicopters were required to perform roll manoeuvres to track the desired position trajectory. However, the maximum roll for the tilt-rotor configuration was approximately 3° , compared to 7° in the case of the fixed-rotor configuration.



Figure 11. Multirotor helicopter position throughout the second manoeuvre.



Figure 12. Multirotor helicopter attitude throughout the second manoeuvre.

These results are in line with those predicted by the computational analysis. Further, these results suggest that a single tilting-rotor multirotor helicopter can be utilised to track trajectories involving horizontal accelerations that are independent of attitude.

7. Concluding Remarks

This paper presents the development and implementation of a single tilting-rotor multirotor helicopter. The dynamics model of such a multirotor helicopter is developed and presented in detail.

The identified dynamics model allows for the establishment of a feedback control system for the proposed multirotor helicopter. The control system provisions the desired control effort to the actuators of the system with the aim of minimising the attitude of the aerial vehicle while maintaining the desired trajectory.

A characterisation of the available control bandwidth of the single tilting-rotor multirotor helicopter is presented and demonstrates the validity of the proposed concept. Further, a computational analysis of the control system and complete dynamics model of the multirotor helicopter demonstrates that the control system allows for effective trajectory tracking while simultaneously minimising the attitude of the multirotor helicopter. The computational analysis is supported by an experimental analysis that compares the performance of the proposed multirotor helicopter to a multirotor helicopter with no tilting rotors. The experimental analysis demonstrates that the single tilting rotor multirotor helicopter is able to achieve comparable trajectory tracking performance for low-speed manoeuvres without requiring a change in attitude. Thus, the attitude and horizontal accelerations have been decoupled for low-speed manoeuvres by the addition of a single tilting rotor.

Further research will be undertaken to explore the full capabilities of the proposed single tilting rotor multirotor helicopter.

Author Contributions: Robert Porter and Bijan Shirinzadeh developed the single tilt-rotor multirotor helicopter and control system architecture, and conceived and designed the experimental analysis; Robert Porter performed the experimental analysis; Man Choi aided in the development of the experimental research platform and control system architecture.

Conflicts of Interest: The authors declare no conflict of interest.

References

- 1. Valavanis, K.P. Advances in Unmanned Aerial Vehicles: State of the Art and the Road to Autonomy; Springer Science & Business Media: Berlin/Heidelberg, Germany, 2008; Volume 33.
- Choi, M.H.; Shirinzadeh, B.; Porter, R. System identification-based sliding mode control for small scaled autonomous aerial vehicles with unknown aerodynamics derivatives. *IEEE/ASME Trans. Mechatron.* 2016, 21, 2944–2952.
- Choi, M.; Porter, R.; Shirinzadeh, B. Comparison of attitude determination methodologies with low cost inertial measurement unit for autonomous aerial vehicle. In Proceedings of the 2013 IEEE/ASME International Conference on Advanced Intelligent Mechatronics, Wollongong, Australia, 9–12 July 2013; pp. 1349–1354.
- Cutler, M.; How, J.P. Actuator constrained trajectory generation and control for variable-pitch quadrotors. In Proceedings of the AIAA Guidance, Navigation, and Control Conference (GNC), Minneapolis, MN, USA, 13–16 August 2012.
- Doyle, C.; Bird, J.; Isom, T.; Kallman, J.; Bareiss, D.; Dunlop, D.; King, R.; Abbott, J.; Minor, M. An avian-inspired passive mechanism for quadrotor perching. *IEEE/ASME Trans. Mechatron.* 2013, 18, 506–517.
- 6. Grzonka, S.; Grisetti, G.; Burgard, W. A fully autonomous indoor quadrotor. *IEEE Trans. Robot.* 2012, 28, 90–100.
- 7. Lee, D.J.; Kaminer, I.; Dobrokhodov, V.; Jones, K. Autonomous feature following for visual surveillance using a small unmanned aerial vehicle with gimbaled camera system. *Int. J. Control Autom. Syst.* **2010**, *8*, 957–966.
- Tayebi, A.; McGilvray, S. Attitude stabilization of a VTOL quadrotor aircraft. *IEEE Trans. Control Syst. Technol.* 2006, 14, 562–571.
- 9. Wang, X.; Shirinzadeh, B. Nonlinear multiple integrator and application to aircraft navigation. *IEEE Trans. Aerosp. Electron. Syst.* **2014**, *50*, 607–622.
- 10. Wang, X.; Shirinzadeh, B.; Ang, M.H. Nonlinear double-integral observer and application to quadrotor aircraft. *IEEE Trans. Ind. Electron.* **2015**, *62*, 1189–1200.
- 11. Porter, R.; Shirinzadeh, B.; Choi, M.H. Experimental analysis of variable collective-pitch rotor systems for multirotor helicopter applications. *J. Intell. Robot. Syst.* **2015**, *83*, 271–288.
- Jayakody, H.; Katupitiya, J. An Adaptive Variable Structure Control methodology for attitude and position control of a Vectored Thrust Aerial Vehicle. In Proceedings of the 2015 IEEE International Conference on Advanced Intelligent Mechatronics (AIM), Busan, Korea, 7–11 July 2015; pp. 1014–1019.
- Yuan, W.; Katupitiya, J. Dynamic modelling and control of a vectored thrust aerial vehicle. In Proceedings of the 2013 IEEE/ASME International Conference on Advanced Intelligent Mechatronics (AIM), Wollongong, Australia, 9–12 July 2013; pp. 1361–1366.
- 14. Cetinsoy, E. Design and simulation of a holonomic quadrotor UAV with sub-rotor control surfaces. In Proceedings of the 2012 IEEE International Conference on Robotics and Biomimetics (ROBIO), Guangzhou, China, 11–14 December 2012; pp. 1164–1169.
- 15. Kendoul, F.; Fantoni, I.; Lozano, R. Modelling and control of a small autonomous aircraft having two tilting rotors. *IEEE Trans. Robot.* **2006**, *22*, 1297–1302.
- Ryll, M.; Bülthoff, H.H.; Giordano, P.R. Modelling and control of a quadrotor UAV with tilting propellers. In Proceedings of the 2012 IEEE International Conference on Robotics and Automation (ICRA), St. Paul, MN, USA, 14–18 May 2012; pp. 4606–4613.
- 17. Leishman, J.G. *Principles of Helicopter Aerodynamics;* Cambridge Aerospace Series; Cambridge University Press: New York, NY, USA, 2006.

- Bouabdallah, S.; Noth, A.; Siegwart, R. PID vs LQ control techniques applied to an indoor micro quadrotor. In Proceedings of the 2004 IEEE/RSJ International Conference on Intelligent Robots and Systems, 28 September–2 October 2004; Volume 3, pp. 2451–2456.
- 19. Bresciani, T. Modelling, Identification and Control of a Quadrotor Helicopter. Master's Thesis, Lund University, Lund, Sweden, 2008.
- Oppenheimer, M.W.; Doman, D.B.; Bolender, M.A. Control allocation for over-actuated systems. In Proceedings of the 14th Mediterranean Conference on Control and Automation, Ancona, Italy, 28–30 June 2006; pp. 1–6.
- 21. Markley, F.L. Fast quaternion attitude estimation from two vector measurements. *J. Guid. Control Dyn.* **2002**, 25, 411–414.
- Porter, R.; Shirinzadeh, B.; Choi, M.H.; Bhagat, U. Laser interferometry-based tracking of multirotor helicopters. In Proceedings of the 2015 IEEE International Conference on Advanced Intelligent Mechatronics (AIM), Busan, Korea, 7–11 July 2015; pp. 1559–1564.
- 23. Martin, P.; Salaün, E. Design and implementation of a low-cost observer-based attitude and heading reference system. *Control Eng. Pract.* **2010**, *18*, 712–722.
- 24. Cominos, P.; Munro, N. PID controllers: Recent tuning methods and design to specification. *IEE Proc. Control Theory Appl.* **2002**, *149*, 46–53.



© 2016 by the authors; licensee MDPI, Basel, Switzerland. This article is an open access article distributed under the terms and conditions of the Creative Commons Attribution (CC-BY) license (http://creativecommons.org/licenses/by/4.0/).