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Design of Distributed Multi-Actuator Systems with Incomplete State Information for Vibration Control of Large Structures [†]

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Abstract: In this paper, we investigate the design and performance of static feedback controllers with partial-state information for the seismic protection of tall buildings equipped with incomplete multi-actuation systems. The proposed approach considers a partially instrumented multi-story building with an incomplete system of interstory force—actuation devices implemented on selected levels of the building, and an associated set of collocated sensors that measure the corresponding interstory drifts and interstory velocities. The main elements of the proposed controller design methodology are presented by means of a twenty-story building equipped with a system of ten interstory actuators arranged in three different layouts: concentrated, semi-distributed and fully-distributed. For these control configurations, partial-state controllers are designed following a static output-feedback H-infinity controller design approach, and the corresponding frequency and time responses are investigated. The obtained results clearly indicate that the proposed partial-state controllers are effective in mitigating the building seismic response. They also show that a suitable distribution of the instrumented stories is a relevant factor in the control system performance.

Keywords: structural vibration control; static output-feedback; large-scale structures; multi-actuator systems; distributed sensors; information constraints

1. Introduction

Vibration control of large buildings and civil structures is an important issue that has attracted increasing research attention over the last years [1–5]. In recent works, a variety of complex and sophisticated control systems with multiple actuation devices has been considered to provide an improved level of vibrational protection. Thus, for example, the usage of multiple Tuned Mass Dampers (TMDs) is proposed in [6–8], an innovative Multiple Tuned Mass Damper system that uses the building floors as TMD devices is investigated in [9,10], the combined action of TMDs and Tuned Liquid Column Dampers for control of wind-induced vibrations in super tall buildings is discussed in [11], a system of Multiple Active Tendons is proposed in [12] for vibration control of irregular buildings, and a system of Multiple Cardan Gyroscopes for vibration control of tower structures is proposed in [13]. Interstory force–actuation devices, as those schematically displayed in

Figures 1b and 2, are particularly suitable to make up multi-actuator systems for large-scale structures. This kind of actuation device facilitates acquiring valuable state information by means of collocated sensors [14], and also allows for considering complex control configurations formed by a large set of actuators and sensors that are distributed throughout the structure. Relevant works in this line include the design and study of complex multi-actuation systems with information constraints using advanced control design methodologies. Thus, the design of optimal passive-damping systems using a state-feedback control approach is conducted in [15,16]. In [17], decentralized time-delayed dynamic output-feedback H_{∞} controllers are obtained using a homotopic transformation. Decentralized and semi-decentralized control strategies based on substructure decomposition are investigated in [18-20]. Decentralized static output-feedback H_{∞} controllers are obtained in [21] by solving a single-step Linear Matrix Inequality (LMI) optimization problem. Static output-feedback H_{∞} controllers with a finite frequency range are proposed in [22] and further developed in [23]. An adaptive control strategy for nonlinear building models with semi-active magnetorheological dampers is investigated in [24]. Dynamic output-feedback controllers with both continuous and sampled feedback information are computed in [25] by means of iterative LMI procedures. In addition, an LMI formulation for robust non-fragile static controllers with sampled-data feedback information is obtained in [26] using a dissipativity approach.

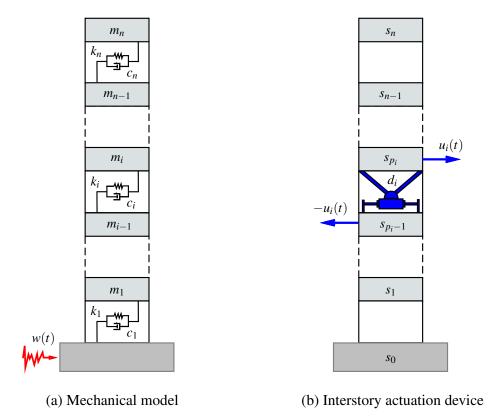


Figure 1. (a) simplified mechanical model for an n-story building with story masses m_i , stiffness coefficients k_i and damping coefficients c_i , $i=1,\ldots,n$; and (b) interstory actuation device. The actuation device d_i is implemented at the building position p_i between the stories s_{p_i-1} and s_{p_i} . The interstory actuator d_i produces a pair of opposite structural forces of magnitude $|u_i(t)|$ on the supporting stories s_{p_i-1} and s_{p_i} .

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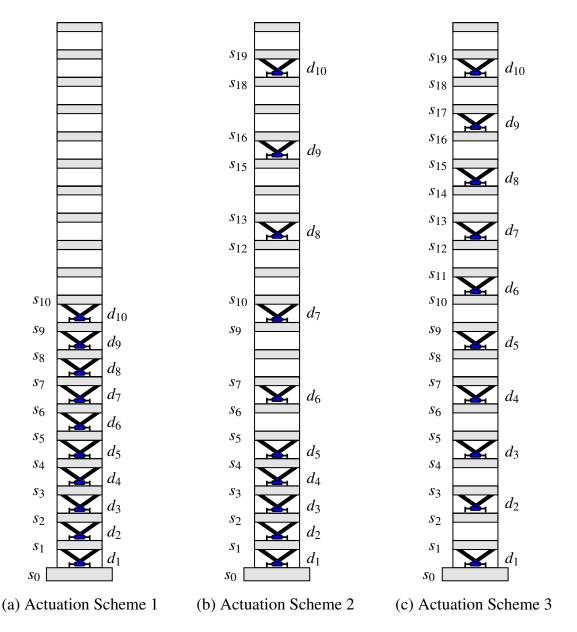


Figure 2. Incomplete actuation schemes in a 20-story building. (a) concentrated actuation scheme AS1 with m=10 actuators located at positions $AP_1=[1,2,\ldots,10]$; (b) semi-distributed actuation scheme AS2 with m=10 actuators located at positions $AP_2=[1,2,3,4,5,7,10,13,16,19]$; (c) distributed actuation scheme AS3 with m=10 actuators located at positions $AP_3=[1,3,5,7,9,11,13,15,17,19]$.

In this paper, we present a controller design methodology for the seismic protection of tall buildings equipped with a distributed set of actuators and collocated sensors. More specifically, we assume that some of the building stories are equipped with an interstory force–actuation device and a sensing unit that can measure the corresponding interstory drift and interstory velocity [14]. We also assume that the building is only partially instrumented and, consequently, the available state information is incomplete. This kind of control configuration is poorly studied in the literature and poses serious computational challenges in large-dimensional problems. The solution proposed in this work is based on an advanced static output-feedback H_{∞} approach that allows for obtaining controllers with incomplete state information by solving a two-step Linear Matrix Inequality (LMI) optimization problem [27–29]. This controller design methodology is applied to study the effectiveness of different configurations of incomplete multi-actuator systems with partial state information, paying special attention to the influence of the actuation system distribution on the control performance.

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In particular, we consider three different actuation schemes for a 20-story building equipped with a system of 10 actuation devices: a concentrated actuation scheme, a semi-distributed actuation scheme, and a fully-distributed actuation scheme (see Figure 2). For these three actuation layouts, partial-state controllers that only use the state information corresponding to the instrumented stories are designed and their frequency and time responses are investigated. The obtained results confirm the effectiveness of structural vibration control strategies with incomplete multi-actuator systems and partial state information. They also point out that an enhanced performance can be achieved by a suitable distribution of the instrumented stories.

The rest of the paper is organized as follows: in Section 2, a state-space model for the twenty-story building with incomplete actuation schemes is presented. In Section 3, full-state and partial-state controllers are designed for the different control configurations and the corresponding frequency characteristics are investigated. In Section 4, numerical simulations of the time responses are conducted and illustrative peak-value plots of the interstory drifts, absolute story accelerations and control efforts are presented and compared. Finally, in Section 5, some conclusions and future research directions are briefly discussed.

2. Building Model

In this section, we provide a dynamical model for the lateral displacement of an *n*-story building equipped with different actuation schemes. The particular models corresponding to a twenty-story building with the actuation schemes displayed in Figure 2 are discussed in detail. These particular models will be used in the controllers designs presented in Section 3 and the numerical simulations carried out in Section 4. The building motion can be described by the second-order differential equation

$$M\ddot{q}(t) + C_d \dot{q}(t) + K_s q(t) = T_u u(t) + T_w w(t),$$
 (1)

where $q(t) = [q_1(t), \ldots, q_n(t)]^T$ is the vector of story displacements with respect to the ground, w(t) is the seismic ground acceleration and u(t) is the vector of control actions. M, C_d and K_s are the mass, damping and stiffness matrices, respectively, which model the mechanical characteristics of the building; T_u is the control location matrix, which determines the structural effect of the actuation forces; and T_w is the disturbance input matrix, which models the effect of the ground disturbance. The mass matrix is a diagonal matrix $M = \text{diag}(m_1, m_2, \ldots, m_n)$ and the stiffness matrix has the following tridiagonal structure:

$$K_{s} = \begin{bmatrix} k_{1} + k_{2} & -k_{2} \\ -k_{2} & k_{2} + k_{3} & -k_{3} \\ & \cdots & \cdots & \cdots \\ & & \cdots & \cdots \\ & & -k_{n-1} & k_{n-1} + k_{n} & -k_{n} \\ & & -k_{n} & k_{n} \end{bmatrix},$$
(2)

where m_i and k_i , $i=1,\ldots,n$ denote the mass and stiffness coefficient of the i-th story, respectively (see Figure 1a). When the story damping coefficients c_i , $i=1,\ldots,n$ are known, a damping matrix C_d with the structure shown in Equation (2) can be obtained by substituting the coefficients k_i by the corresponding c_i . However, most frequently, the story damping coefficients are unknown. In this case, the damping matrix C_d can be computed from M and K by setting a proper damping ratio on the building modes [30]. The disturbance input matrix has the form $T_w = -M[1]_{n \times 1}$, where $[1]_{n \times 1}$ denotes a vector of dimension n with all its entries equal to 1. The structure of the control location matrix T_u depends on the particular characteristics of the considered actuation scheme. In this work, we assume that the control forces are exerted by a system of m devices d_1, \ldots, d_m , where d_i is an interstory actuator implemented at the building position p_i . As indicated in Figure 1b, the actuation device d_i produces a force $u_i(t)$ on the story number p_i (upper story), and an opposite force $-u_i(t)$

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on the story number $p_i - 1$ (lower story). An actuation scheme is determined by a list of actuation positions $AP = [p_1, \ldots, p_m]$. If the actuation scheme is complete, that is, if an actuation device is implemented at every building level, then the list of actuation positions is $AP_c = [1, 2, \ldots, n]$ and the corresponding control location matrix is a square matrix of size n with the following upper-diagonal band form:

For an incomplete actuation scheme with a list of positions $AP = [p_1, \ldots, p_m], m < n$, the corresponding control location matrix T_u^{AP} is a rectangular matrix of size $n \times m$ that can be obtained by extracting the corresponding columns from $T_u^{AP_c}$, that is:

$$T_u^{AP} = T_u^{AP_c}(1, 2, \dots, n; AP) = T_u^{AP_c}(1, 2, \dots, n; p_1, \dots, p_m).$$
 (4)

Thus, for the concentrated actuation scheme AS1 in Figure 2a, the list of positions is $AP_1 = [1, 2, ..., 10]$ and we have the control location matrix $T_u^{AP_1} = T_u^{AP_c}(1, 2, ..., 20; AP_1)$, that is:

For the semi-distributed actuation scheme AS2 in Figure 2b, the actuation positions are $AP_2 = [1, 2, 3, 4, 5, 7, 10, 13, 16, 19]$ and the corresponding control location matrix has the following form:

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Finally, for the fully-distributed actuation scheme AS3 in Figure 2c, the list of actuation positions is $AP_3 = [1, 3, 5, 7, 9, 11, 13, 15, 17, 19]$, and we obtain the control location matrix

The interstory drift $r_i(t)$ is the relative displacement between the consecutive stories located at the building levels i and i-1. The vector of interstory drifts $r(t) = [r_1(t), r_2(t), \dots, r_n(t)]^T$ can be computed as follows:

$$\begin{cases}
r_1(t) = q_1(t), \\
r_i(t) = q_i(t) - q_{i-1}(t), & \text{for } i = 2, \dots, n.
\end{cases}$$
(8)

By considering the state vector

$$x(t) = [r_1(t), \dot{r}_1(t), r_2(t), \dot{r}_2(t), \dots, r_n(t), \dot{r}_n(t)]^T$$
(9)

that arranges the interstory drifts and interstory velocities in increasing order, we obtain a first-order state-space model

$$\dot{x}(t) = A x(t) + B u(t) + E w(t),$$
 (10)

with

$$A = \mathcal{C} \,\widehat{A} \,\mathcal{C}^{-1}, \quad B = \mathcal{C} \,\widehat{B}, \quad E = \mathcal{C} \,\widehat{E},$$
 (11)

$$\widehat{A} = \begin{bmatrix} [0]_{n \times n} & I_n \\ -M^{-1}K_s & -M^{-1}C_d \end{bmatrix}, \quad \widehat{B} = \begin{bmatrix} [0]_{n \times m} \\ M^{-1}T_u \end{bmatrix}, \quad \widehat{E} = \begin{bmatrix} [0]_{n \times 1} \\ -[1]_{n \times 1} \end{bmatrix}, \quad (12)$$

where I_n is the identity matrix of dimension n, $[0]_{n \times m}$ represents a zero-matrix of the indicated dimensions, and C is the change-of-basis matrix corresponding to the state transformation

$$x(t) = \mathcal{C} \begin{bmatrix} q(t) \\ \dot{q}(t) \end{bmatrix}. \tag{13}$$

3. Controllers Design

In this section, we consider a twenty-story building model defined by the mass, stiffness and damping parameters collected in Table 1 [14], and equipped with the incomplete actuation schemes AS1, AS2 and AS3 displayed in Figure 2. For these three actuation layouts, we first compute optimal H_{∞} state-feedback controllers by assuming that the building incorporates a complete system of sensors that provide full-access to the state information. Next, we consider a more realistic scenario of a

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partially instrumented building equipped with a restricted system of sensors, which are collocated with the actuation devices. For this second control setup, we compute suboptimal H_{∞} controllers following a static output-feedback approach.

Table 1. Twent	v-story buildi	ng model: mass	s, stiffness and	damping of	characteristics.

Story	1–5	6–11	12–14	15–17	18–19	20
mass ($\times 10^6$ Kg) stiffness ($\times 10^8$ N/m) relative damping		1.10 5.54		1.10 2.91	1.10 2.56	1.10 1.72

3.1. Controllers with Full-State Information

Assuming that the control objective is to reduce the building vibrational response by means of moderate control actions, we consider the vector of controlled-outputs

$$z(t) = C_z x(t) + D_z u(t)$$
(14)

defined by the matrices

$$C_z = \begin{bmatrix} I_{2n} \\ [0]_{m \times 2n} \end{bmatrix}, \quad D_z = \alpha \begin{bmatrix} [0]_{2n \times m} \\ I_m \end{bmatrix}, \tag{15}$$

where α is a scaling factor that allows adjusting the control action intensity. When the full state information is available, we can consider a state-feedback controller

$$u(t) = Gx(t) \tag{16}$$

defined by the state gain matrix $G \in \mathbb{R}^{m \times 2n}$, which produces the closed-loop system

$$\begin{cases} \dot{x}(t) = A_G x(t) + Ew(t), \\ z(t) = C_G x(t), \end{cases}$$
(17)

with

$$A_G = A + BG, \qquad C_G = C + DG. \tag{18}$$

The H_{∞} controller design methodology considers the system norm

$$\gamma_{\rm G} = \sup_{\|w\|_2 \neq 0} \frac{\|z\|_2}{\|w\|_2},\tag{19}$$

where $\|f\|_2 = \left[\int_0^\infty f^T(t)\,f(t)\,dt\right]^{1/2}$ is the usual continuous 2-norm. The value γ_G represents the worst-case gain from the disturbance-input to the closed-loop controlled-output defined by the control gain matrix G. The controller design objective is to obtain an optimal state-feedback controller $u(t) = \widetilde{G}x(t)$ that produces an asymptotically stable matrix $A_{\widetilde{G}}$ and attains a minimum H_∞ -norm value $\gamma_{\widetilde{G}}$. Using an LMI formulation, the optimal state-feedback H_∞ controller can be computed by solving the following optimization problem [31]:

$$\mathcal{P}: \begin{cases} \text{maximize } \eta, \\ \text{subject to } X > 0, \, \eta > 0, \text{ and the LMI in (21),} \end{cases}$$
 (20)

$$\begin{bmatrix} AX + XA^T + BY + Y^TB^T + \eta EE^T & * \\ C_zX + D_zY & -I_{2n+m} \end{bmatrix} < 0,$$
(21)

where * denotes the transpose of the symmetric entry and $X=X^T\in\mathbb{R}^{2n\times 2n}$, $Y\in\mathbb{R}^{m\times 2n}$ are the optimization variables. If an optimal value $\widetilde{\eta}$ is attained in \mathcal{P} for the pair $(\widetilde{X},\widetilde{Y})$, then the state-feedback gain matrix $\widetilde{G}=\widetilde{Y}\widetilde{X}^{-1}$ is an optimal solution to the H_∞ controller synthesis problem and the corresponding γ -value can be computed as $\gamma_{\widetilde{G}}=\left(\widetilde{\eta}\right)^{-1/2}$.

For a given state-feedback controller u(t) = Gx(t), the corresponding γ -value can also be computed by considering the disturbance to controlled-output closed-loop transfer function

$$T_G(\omega) = C_G(2\pi\omega i I_{2n} - A_G)^{-1} E_{\ell}$$
 (22)

and solving the optimization problem

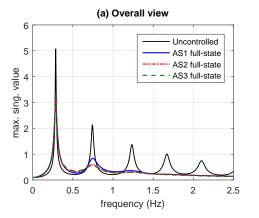
$$\gamma_G = \sup_{\omega} \ \sigma_{\max} [T_G(\omega)], \tag{23}$$

where $j=\sqrt{-1}$, ω is the frequency in hertz, and $\sigma_{\max}[\,\cdot\,]$ denotes the maximum singular value.

To obtain an optimal state-feedback control gain matrix \widetilde{G}_1 for the actuation scheme AS1, we solve the LMI optimization problem \mathcal{P} with the values n=20, m=10; the matrices A, B, E in Equation (11) corresponding to the values in Table 1 and the control location matrix $T_u^{AP_1}$ in Equation (5); and the controlled-output matrices C_z , D_z in Equation (15) with the scaling factor $\alpha=10^{-7.4}$. Proceeding in a similar way, we also compute optimal state-feedback control gain matrices \widetilde{G}_2 and \widetilde{G}_3 for the actuation schemes AS2 and AS3 using, respectively, the control location matrices $T_u^{AP_2}$ in Equation (6) and $T_u^{AP_3}$ in Equation (7). The γ -values obtained for the full-state optimal controllers are the following:

$$\gamma_{\tilde{G}_1} = 2.9153, \qquad \gamma_{\tilde{G}_2} = 3.0850, \qquad \gamma_{\tilde{G}_3} = 2.9852$$
 (24)

and the corresponding frequency responses are displayed in Figure 3. Looking at the overall view presented in Figure 3a, it can be appreciated that the three controllers produce a significant reduction of the building resonant peaks. In accordance with the obtained γ -values, the plots in Figure 3b show that a slightly higher performance is attained by the concentrated actuation scheme AS1 with the controller \widetilde{G}_1 in the main resonant peak. However, looking again at the overall view in Figure 3a, it should be noted that a better behavior is exhibited by the distributed actuation schemes AS2 and AS3 in the second resonant peak. The γ -value corresponding to the uncontrolled system is $\gamma_{\rm unc} = 5.0720$. The relative γ -value reductions with respect to the uncontrolled building response attained by the optimal full-state controllers are collected in Table 2.



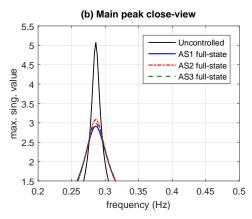


Figure 3. Frequency response of the optimal controllers with full-state information. Maximum singular values of the closed-loop transfer functions $T_{\widetilde{G}_1}(\omega)$ (blue solid line), $T_{\widetilde{G}_2}(\omega)$ (red dash-dotted line) and $T_{\widetilde{G}_3}(\omega)$ (green dashed line), and the open-loop transfer function $T(\omega) = C(2\pi\omega j I_{2n} - A)^{-1}E$ (black solid line). (a) overall view of the first five resonant modes; and (b) close view of the main resonant peak.

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Table 2. Percentages of γ -value reductions attained by the full-state and partial-state controllers for the actuation schemes AS1, AS2 and AS3 with respect to the H_{∞} -norm of the uncontrolled building $\gamma_{\text{unc}} = 5.0720$.

Actuation Scheme	AS1	AS2	AS3	
Full-state controller \widetilde{G}_i	42.52	39.18	41.14	
Partial-state controller \widetilde{K}_i	38.74	39.18	41.14	

3.2. Controllers with Partial State Information

In this section, we assume that the available feedback information is supplied by a system of sensors that are collocated with the actuation devices, and provide a vector of 2*m* measured outputs

$$y(t) = [y_1(t), y_2(t), \dots, y_{2m-1}(t), y_{2m}(t)]^T,$$
(25)

where $y_{2i-1}(t) = r_{p_i}(t)$ and $y_{2i}(t) = \dot{r}_{p_i}(t)$ are, respectively, the interstory drift and interstory velocity associated to the actuation device d_i implemented at the building position p_i (between the stories s_{p_i-1} and s_{p_i}). Using a measured-output matrix $C_y \in \mathbb{R}^{2m \times 2n}$, the vector y(t) can be written in the form $y(t) = C_y x(t)$. For a complete actuation scheme, C_y is an identity matrix of dimension 2n. For an incomplete actuation scheme defined by a list of positions $AP = [p_1, \dots, p_m]$, the corresponding measured-output matrix C_y^{AP} can be obtained by selecting the rows in positions $2p_i - 1, 2p_i, i = 1, \dots, m$ from the identity matrix I_{2n} , that is:

$$C_y^{AP} = I_{2n}(2p_1 - 1, 2p_1, \dots, 2p_m - 1, 2p_m; 1, 2, \dots, 2n).$$
 (26)

Thus, for the concentrated actuation scheme AS1, the list of actuation positions is $AP_1 = [1, 2, ..., 10]$ and the corresponding measured-output matrix has the form $C_y^{AP_1} = [I_{20}, [0]_{20 \times 20}]$. For the distributed actuation scheme AS2, the list of actuation positions is $AP_2 = [1, 2, 3, 4, 5, 7, 10, 13, 16, 19]$ and the associated measured-output matrix $C_y^{AP_2}$ is the submatrix of the identity matrix I_{40} formed by the rows in positions 1, 2, 3, 4, 5, 6, 7, 8, 9, 10, 13, 14, 19, 20, 25, 26, 31, 32, 37 and 38. For the distributed actuation scheme AS3, the list of actuation positions is $AP_3 = [1, 3, 5, 7, 9, 11, 13, 15, 17, 19]$ and the associated measured-output matrix $C_y^{AP_3}$ is the submatrix of the identity matrix I_{40} formed by the rows in positions 1, 2, 5, 6, 9, 10, 13, 14, 17, 18, 21, 22, 25, 26, 29, 30, 33, 34, 37 and 38.

Considering the imposed constraints on the feedback information, we are interested in designing a static output-feedback controller

$$u(t) = Ky(t) \tag{27}$$

defined by a constant output-feedback gain matrix $K \in \mathbb{R}^{m \times 2m}$, which allows computing the vector of control actions from the vector of measured-outputs y(t) by means of a simple matrix multiplication. Now, the closed-loop system has the following form:

$$\begin{cases} \dot{x}(t) = A_K x(t) + Ew(t), \\ z(t) = C_K x(t), \end{cases}$$
 (28)

with

$$A_K = A + BKC_y, \qquad C_K = C_z + D_z KC_y, \tag{29}$$

and the associated H_{∞} -norm takes the value

$$\gamma_K = \sup_{\omega} \sigma_{\max} [T_K(\omega)], \tag{30}$$

where the closed-loop transfer function $T_K(\omega)$ has now the form

$$T_K(\omega) = C_K (2\pi\omega j I_{2n} - A_K)^{-1} E. \tag{31}$$

In this new context, the H_∞ controller design aims at obtaining an optimal static output-feedback controller $u(t) = \widetilde{K}y(t)$ that produces an asymptotically stable closed-loop matrix $A_{\widetilde{K}}$ and, at the same time, attains a minimum H_∞ -norm value $\gamma_{\widetilde{K}}$. The effective computation of this kind of optimal controllers for large-scale systems is a difficult problem that poses major theoretical and computational challenges. However, according to the results presented in [27,28], a suboptimal static output-feedback H_∞ controller can be designed by solving the following LMI optimization problem:

$$\mathcal{P}_{0}: \begin{cases} \text{maximize } \eta, \\ \text{subject to } X_{Q} > 0, X_{R} > 0, \eta > 0 \text{ and the LMI in Equation (33),} \end{cases}$$
 (32)

$$\begin{bmatrix} AQX_{Q}Q^{T} + QX_{Q}Q^{T}A^{T} + ARX_{R}R^{T} + RX_{R}R^{T}A^{T} + BY_{R}R^{T} + RY_{R}^{T}B^{T} + \eta EE^{T} & * \\ C_{z}QX_{Q}Q^{T} + C_{z}RX_{R}R^{T} + D_{z}Y_{R}R^{T} & -I_{2n+m} \end{bmatrix} < 0, \quad (33)$$

where X_Q , X_R and Y_R are the optimization variables, Q is a matrix whose columns contain a basis of $Ker(C_V)$, and the matrix R has the form

$$R = C_y^{\dagger} + QQ^{\dagger} \widetilde{X} C_y^T (C_y \widetilde{X} C_y^T)^{-1}, \tag{34}$$

where

$$C_y^{\dagger} = C_y^T (C_y C_y^T)^{-1}, \qquad Q^{\dagger} = (Q^T Q)^{-1} Q^T$$
 (35)

are the *Moore–Penrose* pseudoinverses of C_y and Q, respectively, and \widetilde{X} is the optimal X-matrix obtained in the LMI optimization problem \mathcal{P} corresponding to the full-state controller design. If an optimal value $\widetilde{\eta}_o$ is attained in \mathcal{P}_o for the triplet $(\widetilde{X}_Q, \widetilde{X}_R, \widetilde{Y}_R)$, then the matrix $\widetilde{K} = \widetilde{Y}_R(\widetilde{X}_R)^{-1}$ defines a static output-feedback controller $u(t) = \widetilde{K}y(t)$ with asymptotically stable closed-loop matrix $A_{\widetilde{K}}$, and the value $\widetilde{\gamma}_{\widetilde{V}} = (\widetilde{\eta}_o)^{-1/2}$ provides an *upper bound* of the associated H_∞ -norm value $\gamma_{\widetilde{V}}$.

To design a partial-state controller for the proposed twenty-story building model with the concentrated actuation scheme AS1, we consider the measured-output vector

$$y^{(1)}(t) = [r_1(t), \dot{r}_1(t), r_2(t), \dot{r}_2(t), \dots, r_{10}(t), \dot{r}_{10}(t)]^T,$$
(36)

and solve the LMI optimization problem \mathcal{P}_0 in Equation (32) with the same matrices A, B, E, C_z , D_z used in the design of the state-feedback gain matrix \widetilde{G}_1 and the matrices Q, R defined by the measured-output matrix $C_y^{AP_1} = [I_{20}, [0]_{20 \times 20}]$ and the optimal X-matrix computed in the associated full-state controller design. As a result, we obtain an output-feedback gain matrix \widetilde{K}_1 with a γ -value upper bound $\widetilde{\gamma}_{\widetilde{K}_1} = 3.3846$. The frequency responses of the partial-state controller $u(t) = \widetilde{K}_1 y^{(1)}(t)$ and the associated full-state controller $u(t) = \widetilde{G}_1 x(t)$ are displayed in Figure 4b. The plots in the figure show a small (but clearly appreciable) loss of performance of the partial-state controller (blue solid line) with respect to the full-state controller (black dashed line). Using the closed-loop transfer function $T_{\widetilde{K}_1}(\omega)$ and solving the optimization problem in Equation (30), we obtain that the actual γ -value of the partial-state controller is $\gamma_{\widetilde{K}_1} = 3.1071$, which represents an increment of 6.58% with respect to the optimal value $\gamma_{\widetilde{G}_1} = 2.9153$.

For the semi-distributed actuation scheme AS2, the vector of measured outputs $y^{(2)}(t)$ contains the interstory drifts and interstory velocities indicated in the list of positions $AP_2 = [1, 2, 3, 4, 5, 7, 10, 13, 16, 19]$. In this case, we obtain an output-feedback gain matrix \widetilde{K}_2 with a γ -value upper bound $\widetilde{\gamma}_{\widetilde{K}_2} = 3.0857$ by solving the LMI optimization problem \mathcal{P}_0 with the same matrices A, B, E, C_z, D_z used in the design of the state-feedback gain matrix \widetilde{G}_2 and the matrices Q, R defined by

the measured-output matrix $C_y^{AP_2}$ and the corresponding auxiliary X-matrix. Comparing the value of the upper bound $\widetilde{\gamma}_{\widetilde{K}_2}$ with the optimal value $\gamma_{\widetilde{G}_2}=3.0850$, it can be seen that the static output-feedback controller $u(t)=\widetilde{K}_2\,y^{(2)}(t)$ is practically optimal. In fact, the plots presented in Figure 4c show that the frequency response of the partial-state controller defined by the output-feedback gain matrix \widetilde{K}_2 (red solid line) is practically equal to the frequency response of the optimal full-state controller defined by the state-feedback gain matrix \widetilde{G}_2 (black dashed line).

For the fully-distributed actuation scheme AS3, the vector of measured outputs $y^{(3)}(t)$ contains the interstory drifts and interstory velocities indicated in the list of positions $AP_3=[1,3,5,7,9,11,13,15,17,19]$. Following the same computational procedure used in the previous cases, we obtain an output-feedback gain matrix \widetilde{K}_3 with a γ -value upper bound $\widetilde{\gamma}_{\widetilde{K}_3}=2.9921$. By considering the closed-loop transfer function $T_{\widetilde{K}_3}(\omega)$ and solving the optimization problem in Equation (30), we find that the actual γ -value of the partial-state controller $u(t)=\widetilde{K}_3\,y^{(3)}(t)$ is $\gamma_{\widetilde{K}_3}=2.9852$, which matches the optimal value $\gamma_{\widetilde{G}_3}=2.9852$ attained by the full-state controller. As it happened with the semi-distributed scheme AS2, looking at the plots presented in Figure 4d, it can be appreciated that the frequency response of the partial-state controller defined by the output-feedback gain matrix \widetilde{K}_3 (green solid line) is practically equal to the frequency response of the optimal full-state controller defined by the state-feedback gain matrix \widetilde{G}_3 (black dashed line).

To provide a better comparison of the partial-state controllers behavior, the corresponding frequency responses are displayed together in Figure 4a, where the blue solid line represents the partial-state controller defined by the output-feedback gain matrix \widetilde{K}_1 , the red dash-dotted line pertains to the partial-state controller defined by \widetilde{K}_2 , and the green dashed line represents the partial-state controller defined by \widetilde{K}_3 . The frequency response of the uncontrolled building (black solid line) is also included as a reference. Looking at the plots in this figure, it can be clearly seen that the three partial-state controllers produce a significant reduction of the building resonant peaks. Moreover, considering the percentages of γ -value reduction collected in Table 2 and the behavior in the secondary resonant peaks, it becomes apparent that the best performance under partial-state information corresponds to the fully-distributed configuration AS3.

Remark 1. The total number of elements in the control matrix is related to the number of optimization variables in the LMI design problem. This factor can produce a critical increase of the computation time required by the controller design procedure and is a major issue in large-scale problems. For an n-story building with a system of m actuation devices, the full-state control matrices have 2 nm elements and the proposed partial-state control matrices have 2 m² elements. Specifically, for the 20-story building with m=10 actuation devices considered in this paper, the full-state control matrices have 400 elements and the partial-state control matrices have 200 elements. Obtaining the full-state control matrix \widetilde{G}_1 by solving the LMI optimization problem \mathcal{P} in Equation (20) has required a computation time of 42.71 s. Computing the partial-state control matrix \widetilde{K}_1 by solving the LMI optimization problem \mathcal{P}_0 in Equation (32) has required an additional computation time of 6.60 s. For the full-state control matrices \widetilde{G}_2 and \widetilde{G}_3 , the computation times have been 49.60 s and 43.34 s, respectively, and the additional computation times required to obtain the partial-state controllers \widetilde{K}_2 and \widetilde{K}_3 have been 5.28 s and 4.17 s, respectively.

Remark 2. It should be noted that the R-matrix proposed in Equation (34) uses the optimal X-matrix obtained in the design of the full-state controller. For a given incomplete actuation scheme, the proposed two-step controller design procedure involves computing both the full-state and the partial-state controllers. This approach has the advantage of providing a natural reference for the performance of the suboptimal partial-state controller and can be a computationally effective controller design strategy for large-scale structures with incomplete actuation schemes of moderate dimension.

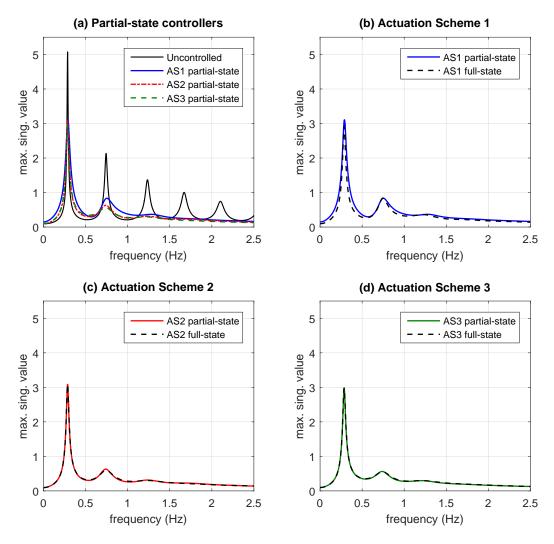


Figure 4. Frequency response of the full-state and partial-state controllers for the different actuation schemes. (a) overall view of the frequency response corresponding to uncontrolled building (black solid line) and the partial-state controllers defined by the output-feedback control matrices \widetilde{K}_1 (blue solid line), \widetilde{K}_2 (red dash-dotted line) and \widetilde{K}_3 (green dashed line); (b) concentrated actuation scheme AS1. Frequency response of the partial-state controller defined by the output-feedback control matrix \widetilde{K}_1 (blue solid line) and the full-state controller defined by the state-feedback control matrix \widetilde{G}_1 (black dashed line); (c) semi-distributed actuation scheme AS2. Frequency response of the partial-state controller defined by the output-feedback control matrix \widetilde{K}_2 (red solid line) and the full-state controller defined by the state-feedback control matrix \widetilde{G}_2 (black dashed line); and (d) fully-distributed actuation scheme AS3. Frequency response of the partial-state controller defined by the output-feedback control matrix \widetilde{K}_3 (green solid line) and the full-state controller defined by the state-feedback control matrix \widetilde{G}_3 (black dashed line).

4. Numerical Results

To illustrate the time-response characteristics of the proposed partial-state controllers, a suitable set of numerical simulations has been carried out using the scaled Kobe 1995 seismic record as ground acceleration input (see Figure 5). For the concentrated actuation scheme AS1, the absolute peak-values of the interstory-drifts, story absolute accelerations and control efforts corresponding to the different control configurations are presented in Figure 6, where the black solid line with squares presents the response of the uncontrolled building, the blue solid line with triangles corresponds to the full-state controller defined by the state-feedback control matrix \tilde{G}_1 and the red dashed line with asterisks

represents the partial-state controller defined by the output-feedback control matrix \widetilde{K}_1 . The time responses corresponding to the semi-distributed actuation scheme AS2 and the fully-distributed actuation scheme AS3 are presented in Figures 7 and 8, respectively, using the same colors, line styles and symbols. To complement this graphical information, the percentages of reduction in the interstory drift peak-values with respect to the uncontrolled response at selected levels of the building (see Remark 3) are collected in Table 3, and the maximum control-effort peak-values are presented in Table 4.

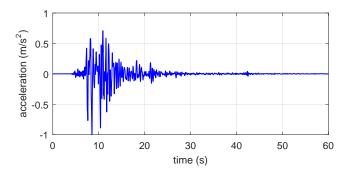


Figure 5. Ground acceleration disturbance. North-South Kobe 1995 seismic record scaled to an acceleration peak-value of 1 m/s^2 .

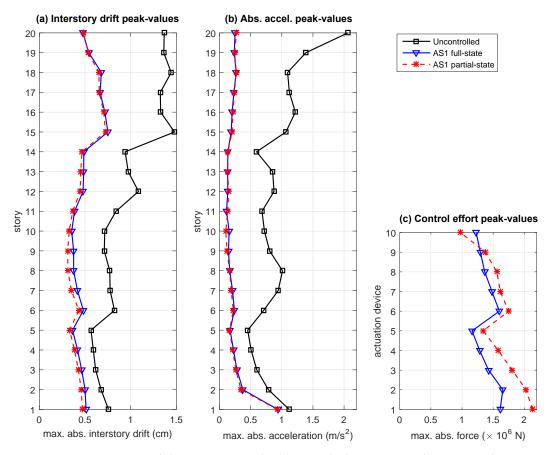


Figure 6. Time response of the twenty-story building with the *concentrated actuation scheme AS1*. Seismic response to the scaled North-South Kobe 1995 ground acceleration record for the uncontrolled configuration (black solid line with squares), the full-state controller defined by the state-feedback gain matrix \widetilde{G}_1 (blue solid line with triangles) and the partial-state controller defined by the output-feedback gain matrix \widetilde{K}_1 (red dashed line with asterisks). (a) interstory drift peak-values; (b) absolute acceleration peak-values; and (c) control effort peak-values.

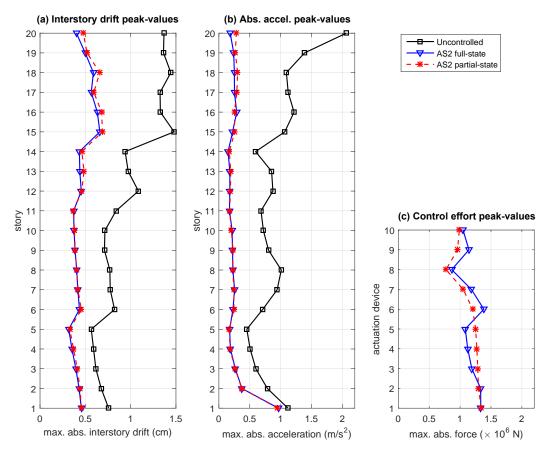


Figure 7. Time response of the twenty-story building with the *semi-distributed actuation scheme AS2*. Seismic response to the scaled North-South Kobe 1995 ground acceleration record for the uncontrolled configuration (black solid line with squares), the full-state controller defined by the state-feedback gain matrix \widetilde{G}_2 (blue solid line with triangles) and the partial-state controller defined by the output-feedback gain matrix \widetilde{K}_2 (red dashed line with asterisks). (a) interstory drift peak-values; (b) absolute acceleration peak-values; and (c) control effort peak-values.

Table 3. Percentage of reduction in the interstory drift peak-values with respect to the uncontrolled response for the proposed full-state and partial-state controllers. The North-South Kobe 1995 seismic record, scaled to an acceleration peak of 1 m/s^2 , has been taken as disturbance input.

Story Level	1	6	12	15	18	20
full-state controller \widetilde{G}_1	32.41	41.37	55.94	49.31	52.97	65.24
full-state controller \widetilde{G}_2	39.76	47.12	58.11	55.38	59.19	70.63
full-state controller \widetilde{G}_3	40.89	43.57	60.03	56.20	59.83	71.94
partial-state controller \widetilde{K}_1	37.77	47.43	58.88	51.15	54.89	64.76
partial-state controller \widetilde{K}_2	39.60	44.67	57.66	53.55	54.37	65.06
partial-state controller \widetilde{K}_3	40.07	42.15	58.96	54.89	58.08	69.78

Table 4. Maximum control-effort peak values (in MN) corresponding to the computed full-state and partial-state controllers. The North-South Kobe 1995 seismic record, scaled to an acceleration peak of 1 m/s^2 , has been taken as disturbance input.

Controller	\widetilde{G}_1	\widetilde{G}_2	\widetilde{G}_3	\widetilde{K}_1	\widetilde{K}_2	\widetilde{K}_3
maximum control-effort peak-value (×10 ⁶ N)	1.656	1.385	1.342	2.132	1.343	1.273

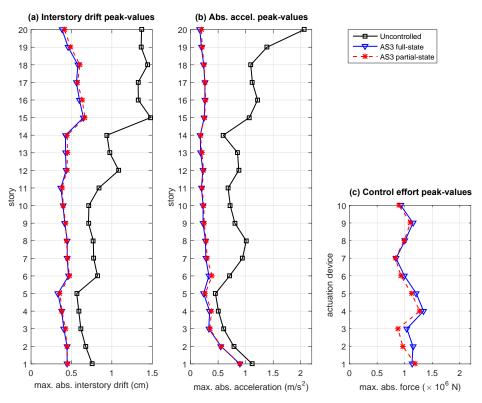


Figure 8. Time response of the twenty-story building with the *fully-distributed actuation scheme AS3*. Seismic response to the scaled North-South Kobe 1995 ground acceleration record for the uncontrolled configuration (black solid line with squares), the full-state controller defined by the state-feedback gain matrix \widetilde{G}_3 (blue solid line with triangles) and the partial-state controller defined by the output-feedback gain matrix \widetilde{K}_3 (red dashed line with asterisks). (a) interstory drift peak-values; (b) absolute acceleration peak-values; and (c) control effort peak-values.

Looking at the plots in Figures 6–8, the following facts can be appreciated: (i) all the proposed controllers provide a good level of reduction in the interstory drift and absolute acceleration peak-values with respect to the uncontrolled response; (ii) for the distributed actuation schemes AS2 and AS3, the behavior of the full-state controller and the partial-state controller are very similar; and (iii) for the concentrated actuation scheme AS1, the interstory drift peak-values produced by the partial-state controller are slightly smaller than those attained by the full-state controller. However, significantly larger control-effort peak-values are also required by the partial-state controller. These time-response results are entirely consistent with the frequency-response characteristics observed in Section 3, and indicate that the performance of the partial-state controllers is not affected by the reduced feedback information in the distributed actuation schemes AS2 and AS3. In contrast, a noticeable loss of performance can be observed in the partial-state controller corresponding to the concentrated actuation scheme AS1. The same facts can be appreciated by looking at the data collected in Table 3 and Table 4, where it becomes apparent that larger percentages of reduction in the interstory drift peak-values are attained by the distributed actuations schemes AS2 and AS3 with smaller control-effort peaks. Particularly remarkable is the performance of the fully-distributed partial-state controller \tilde{K}_3 , which produces larger reductions in the interstory drift peak-values than the full-state controller G_1 and, at the same time, requires a maximum control-effort peak-value that is a 23.1% smaller than the one required by the concentrated full-state controller G_1 .

Remark 3. In order to provide a meaningful summary of the building response, the percentages of reduction in the interstory drift peak-values collected in Table 3 have been constrained to a selected set of building levels, which include the top and bottom level and the intermediate levels where the uncontrolled response presents a

locally maximum value (see the black line with squares in Figure 6a). Specifically, the selected building levels are: 1, 6, 12, 15, 18 and 20. Looking at the data in Table 1, it can be seen that these levels are coincident with the stiffness discontinuities.

Remark 4. All the computations in this paper have been carried out using Matlab® R2015b (MathWorks, Natick, MA, USA) on a regular laptop with an Intel® CoreTM i7-2640M processor (Intel Corporation, Santa Clara, CA, USA) at 2.80 GHz. The LMI optimization problems have been solved with the function mincx() included in the Robust Control ToolboxTM.

5. Conclusions

In this paper, the design of partial-state controllers with incomplete multi-actuation systems for the seismic protection of large buildings has been investigated. The proposed approach considers a partially instrumented building with a system of interstory force—actuation devices implemented at selected levels of the building and an associated set of collocated sensors that provide the interstory drifts and interstory velocities corresponding to the building's instrumented levels.

In order to study the influence of the actuation system distribution on the control performance, three different actuation schemes have been considered for a 20-story building equipped with a system of 10 actuation devices: (i) a concentrated actuation scheme formed by a block of 10 actuation devices implemented at the 10 lowest levels of the building, (ii) a semi-distributed actuation scheme consisting of a block of five actuators implemented at the five lowest levels of the building and a set of five actuators uniformly distributed in the remaining 15 upper levels and (iii) a fully-distributed actuation scheme made up of 10 actuation devices implemented at alternate positions from the first to the nineteenth building levels. For these three different actuation layouts, ideal H_{∞} state-feedback controllers have been designed assuming that the interstory drifts and interstory velocities corresponding to all the building levels are available. Next, partial-state controllers that only use the interstory drifts and interstory velocities of the instrumented building levels as feedback information have been designed using an advanced static output-feedback H_{∞} approach. This controller design methodology provides a novel line of solution to the considered problem, allowing to obtain effective partial-state controllers with a reduced computational cost.

After studying the frequency-response and time-response of the obtained full-state and partial-state controllers, two relevant facts have been observed: (i) for the semi-distributed and fully-distributed actuation schemes, the partial-state controllers and the full-state controllers have practically the same level of performance; and (ii) the fully-distributed actuation scheme is particularly effective, producing similar or even superior levels of response mitigation with lower levels of control effort. For the first fact, a satisfactory explanation can be easily obtained by observing that, despite the reduced system of sensors, the semi-distributed and fully-distributed schemes are still able to capture a "high-quality picture" of the building dynamical state. In contrast, the partial-state controller in the concentrated actuation scheme loses all the state information of the upper half building. Finding a good explanation for the second fact is a more tricky issue. After careful consideration, we have noted that the concentrated and semi-distributed configurations contain actuation devices implemented in adjacent building levels, which can exert conflicting control actions on the common shared floor. We think that this can be a possible explanation for the higher control-effort level observed in the controllers of these actuation schemes.

In summary, the obtained results indicate that properly distributed actuation systems with collocated sensors can be very effective for partial-state control systems of large structures. Additionally, special attention should be paid to avoid possible conflicting interactions between adjacent actuation devices. After these positive outcomes, further research effort should be invested in finding a suitable methodology to determine optimal configurations of distributed multi-actuation systems with partial state information. This is certainly an important and challenging problem, especially for large-scale structures [32,33]. Other related issues of special interest are the study of the potential loss of

performance in adjacent actuation devices and the design of decentralized control strategies with partial-state information.

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