



Article

Research on Temperature Dependence of Single-Event Burnout in Power MOSFETs

Chen Wang ¹, Yi Liu ^{1,*}, Changqing Xu ^{2,*}, Xinfang Liao ¹, Dongdong Chen ¹ and Zhenyu Wu ¹

- Laboratory of Digital IC and Space Application, School of Microelectronics, Xidian University, Xi'an 710071, China; wdh8537@163.com (C.W.)
- ² Guangzhou Institute of Technology, Xidian University, Guangzhou 510555, China
- * Correspondence: yiliu@mail.xidian.edu.cn (Y.L.); cqxu@xidian.edu.cn (C.X.)

Abstract: Power MOSFETs are found to be very vulnerable to single-event burnout (SEB) in space irradiation environments, and the military components generally require that devices could operate reliably as the temperature varies from 218 K to 423 K ($-55\,^{\circ}$ C to $150\,^{\circ}$ C); thus, the temperature dependence of single-event burnout (SEB) in power MOSFETs should be investigated. Our simulation results showed that the Si power MOSFETs are more tolerant to SEB at a higher temperature at the lower LET ($10\,\text{MeV}\cdot\text{cm}^2/\text{mg}$) due to the decrease of the impact ionization rate, which is in good agreement with the previous research. However, the state of the parasitic BJT plays a primary role in the SEB failure mechanism when the LET value is greater than $40\,\text{MeV}\cdot\text{cm}^2/\text{mg}$, which exhibits a completely different temperature dependence from that of $10\,\text{MeV}\cdot\text{cm}^2/\text{mg}$. Results indicate that with the temperature increasing, the lower difficulty to turn on the parasitic BJT and the increasing current gain all make it easier to build up the regenerative feedback process responsible for SEB failure. As a result, the SEB susceptibility of power MOSFETs increases as ambient temperature increases when the LET value is greater than $40\,\text{MeV}\cdot\text{cm}^2/\text{mg}$.

Keywords: power MOSFETs; temperature dependence; LET; single-event burnout (SEB)



Citation: Wang, C.; Liu, Y.; Xu, C.; Liao, X.; Chen, D.; Wu, Z. Research on Temperature Dependence of Single-Event Burnout in Power MOSFETs. *Micromachines* **2023**, *14*, 1028. https://doi.org/10.3390/ mi14051028

Academic Editor: Sadia Ameen

Received: 22 March 2023 Revised: 6 May 2023 Accepted: 9 May 2023 Published: 11 May 2023



Copyright: © 2023 by the authors. Licensee MDPI, Basel, Switzerland. This article is an open access article distributed under the terms and conditions of the Creative Commons Attribution (CC BY) license (https://creativecommons.org/licenses/by/4.0/).

1. Introduction

Power MOSFETs are widely used in space electronic systems, attributed to their strong driving ability, low power consumption, and high block voltages [1,2]. However, power MOSFETs are also found to be very vulnerable to single-event burnout (SEB) in a space irradiation environment, which is a catastrophic failure and can be triggered by the energetic particle penetrating the device in the off-state [3–5].

It has been extensively proven that the SEB failure of power MOSFETs is related to the establishment of the regenerative feedback process, where the two relevant mechanisms are the current amplification of the parasitic BJT and the carrier multiplication at the Nepi/N+sub junction [3,6]. That is, the turn-on and amplification of the parasitic BJT can be maintained with the hole current provided by the impact ionization process, meanwhile, the carrier impact ionization at the N-epi/N+sub junction can also be fed by the electron current coming from the parasitic BJT. This continuous reaction can ultimately lead to the catastrophic and permanent thermal failure of power MOSFETs. Nowadays, the Technology Computer-Aided Design (TCAD) simulator can help us know more details about the SEB failure in power MOSFETs and study other radiation-hardened techniques easily [7–9].

As we know, the ambient temperature varies dramatically in space flight missions; therefore, we have to consider the influence of the ambient temperature on the radiation tolerance of power MOSFETs. Some research has revealed that the SEB tolerance of power MOSFETs increases with the increase of ambient temperature, that is, the worst case for SEB occurs at a lower temperature [10–13]. This is due to that the impact ionization rate decreases with increases in temperature. However, on the other hand, the increase in temperature will increase the forward current of the emitter junction of parasitic BJT, which

Micromachines 2023, 14, 1028 2 of 10

makes it easier to form a local hot spot and induce the SEB failure. Therefore, the two factors influenced by ambient temperature exhibit a completely opposite effect on the SEB performance of power MOSFETs. Thus, to get a better understanding of the temperature dependence of the SEB failure in power MOSFETs, we have to find out which factor plays a leading role in the different conditions.

In this paper, we find that for power MOSFETs, the SEB failure has an opposite temperature dependence at the lower and higher LET. At the lower LET ($10~\text{MeV}\cdot\text{cm}^2/\text{mg}$), the SEB susceptibility of power MOSFETs decreases with the increasing temperature since the decrease of the impact ionization rate plays a dominant role in this case, which agrees well with the previous research [10–12]. While, at the higher LET ($100~\text{MeV}\cdot\text{cm}^2/\text{mg}$), the condition is just the opposite. With the increase in temperature, the transient response of power MOSFETs to the high-energy particle strike is stronger. This is because the parasitic BJT starts to play a primary role in the regenerative feedback mechanism now. Due to the lower difficulty to turn on the parasitic BJT and the increasing current gain at a higher temperature, the worst case for SEB occurs at a higher temperature at the higher LET.

2. TCAD Simulation Model

The schematic cross-sectional view of the simulated power MOSFET device is shown in Figure 1. Table 1 gives the corresponding device structure parameters in detail. As shown in Figure 2, the breakdown voltage (BV) and threshold voltage (V_{TH}) of the device are 424 V and 5.5 V, respectively, which are in good agreement with [7,14]. The single event burnout (SEB) threshold voltage of the device in [14] is 330 V (50% BV) under a LET value of 17 MeV·cm²/mg, and the simulated device in this paper has a SEB threshold voltage of 307 V (72% BV) under the same LET value, exhibiting an increase of about 20% since the adoption of the SEB hardening techniques of the Pplus extension and the insertion of the buffer layer in our device structure [15–17].

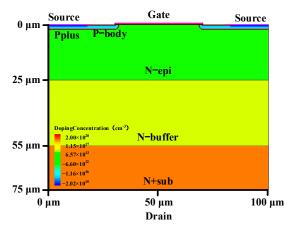


Figure 1. Schematic cross-sectional view of the simulated power MOSFET.

Table 1. Device structural parameters for simulations.

Device Parameters	Values
Gate Oxide Thickness (nm)	100
N+ Source Doping (cm $^{-3}$)	2×10^{20}
Pplus Doping (cm ^{−3})	2×10^{19}
P-Body Doping (cm ⁻³)	$2 imes 10^{17}$
N-Drift Thickness (μm)	25
N-Drift Doping (cm $^{-3}$)	$5 imes 10^{14}$
N Buffer Thickness (μm)	30
N Buffer Doping (cm ⁻³)	5×10^{15}

Micromachines 2023, 14, 1028 3 of 10

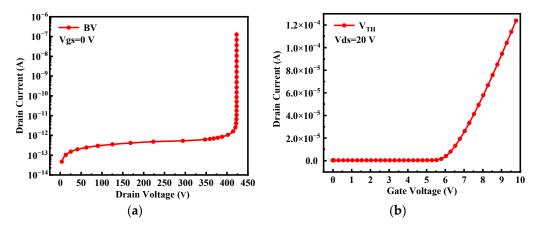


Figure 2. Static characteristics of the power MOSFET device (a) breakdown characteristic and (b) transfer characteristic.

The numerical simulations are performed using the Sentaurus TCAD simulator in this paper. Several physical models have been used to simulate the SEB failure process, including (1) the mobility degradation model, which contains ionized impurity scattering, carrier–carrier scattering, interface scattering, and high field velocity saturation. (2) recombination model, which includes Shockley Read Hall (SRH) recombination considering doping concentration and temperature dependence, Auger recombination regarding the high carrier densities caused by the heavy ion impact. (3) University of Bologna impact ionization model, which is developed to model the carrier multiplication at the N-epi/N+sub junction. (4) electro-thermal model, which considers the lattice temperature variation induced by the large current density after a heavy ion's strike by adding the thermodynamic model to simulate the lattice self-heating effect. In this paper, we define the substrate contact as an ideal heat sink, and its temperature is fixed at the ambient temperature.

Moreover, the heavy ion model is activated to simulate the transient response of the power MOSFET to the ion. The carrier generation rate induced by the heavy ion is computed by:

$$G(l,\omega,t) = G_{LET}(l)R(\omega,l)T(t)$$
(1)

where l is the length of the particle strike track, ω is the radius defined as the perpendicular distance to the track and t is the transient time. In addition, $R(\omega,l)$ and T(t) are Gaussian functions describing the spatial and temporal variations of the carrier generation rate with a characteristic radius of w_0 and a characteristic time of S_{hi} , respectively. The LET is assumed to be a constant value when the energetic ion passes through the device. In this work, the particle strike track is perpendicular to the surface of the device, and the track length is set to 80 μ m, which is able to penetrate the entire device. Table 2 shows the default heavy ion parameters used for our simulations.

Table 2. Parameters used for the heavy ion simulations.

Track Length l_0	Track Radius w_0	Characteristic Time of Gaussian Function S _{hi}	Peak Charge Generation Time t_0	Ion Incident Position x_0
80 μm	0.02 μm	4 ps	2 ns	40 μm

3. Results and Discussion

3.1. Single Event Burnout Sensitive Region

Previous investigations have shown that the neck region of power MOSFETs is the sensitive area to SEB [8,12,18,19]. This is because the neck region, which is far from P-body, is more resistive when the lateral hole current flows to the source electrode, developing a larger ohmic voltage drop across the emitter-base junction, making the parasitic BJT easier

Micromachines 2023, 14, 1028 4 of 10

to be forward biased. Thus, the corresponding neck region from $x_0 = 31 \, \mu m$ to $x_0 = 50 \, \mu m$ of the simulated power MOSFET is chosen to obtain the most sensitive ion strike position in our simulations. Figure 3 exhibits the relationship between the ion's incident position x_0 and SEB threshold voltage ($V_{th,SEB}$) at LET = 75 MeV·cm²/mg, where $V_{th,SEB}$ is defined as the minimum drain voltage required to trigger SEB. It can be concluded that $x_0 = 40 \, \mu m$ is the most sensitive incident position with the lowest $V_{th,SEB}$ of 164 V. Therefore, $x_0 = 40 \, \mu m$ is selected as the default incident position in our following simulations so as to obtain the worst case for SEB. In addition, it should also be pointed out that we regard the SEB failure criterion as the maximum lattice temperature of the device after the ion strike reaches the melting point of silicon (1688 K) [9,20].

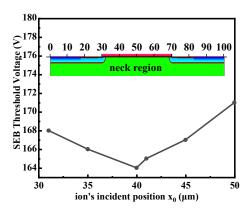


Figure 3. SEB threshold voltage as a function of the ion strike position (LET = $75 \text{ MeV} \cdot \text{cm}^2/\text{mg}$).

3.2. The Temperature Dependence of SEB at Different LETs

Figure 4 shows $V_{th, SEB}$ as a function of ambient temperature (Ta) at different LETs. Our selected ambient temperature varies from 218 K to 423 K (-55 °C to 150 °C), which corresponds to the operating temperature range of military components. It is obvious that the SEB response exhibits different temperature dependencies at different LET values. At the lower LET value of 10 MeV·cm²/mg, $V_{th,SEB}$ increases slightly with the increase of ambient temperature. On the contrary, with the LET value greater than 40 MeV·cm²/mg, $V_{th,SEB}$ decreases as temperature increases. To get a better understanding of the discrepancy of the temperature dependence of SEB at different LETs, the following discussions are mainly focused on the two conditions: condition A is at the lower LET (10 MeV·cm²/mg) and Vds = 300 V; condition B is at the higher LET (100 MeV·cm²/mg) and Vds = 100 V.

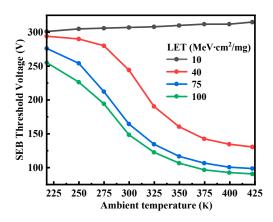


Figure 4. Relationship between $V_{th, SEB,}$ and ambient temperature as a function of LET.

3.2.1. The Temperature Dependence of SEB at the Lower LET

In order to analyze the influence of ambient temperature on SEB sensitivity at the low LET, Figure 5 demonstrates the impact ionization rate distributions at Ta = 250 K and

Micromachines 2023, 14, 1028 5 of 10

Ta = 423 K under condition A (LET = $10 \text{ MeV} \cdot \text{cm}^2/\text{mg}$, Vds = 300 V). From Figure 5, we can find that the peak impact ionization rate is located at the P-body/N-epi junction after the ion strike 100 ps, while it shifts to the N-epi/N-buffer interface after the ion strike 10 ns. This is due to the Kirk effect caused by the injection of electrons from the emitter to the collector and the resulting less positive total charge density in the collector region [18,21]. At Ta = 250 K, the peak impact ionization rates are 9.71×10^{25} /cm³·s and 6.08×10^{25} /cm³·s after the ion strike 100 ps and 10 ns, respectively. However, when Ta = 423 K, the peak impact ionization rates are $6.68 \times 10^{25} / \text{cm}^3 \cdot \text{s}$ and $4.06 \times 10^{25} / \text{cm}^3 \cdot \text{s}$, respectively, which decreases about 40% with the increase of Ta from 250 K to 423 K. This is because the increase of temperature will aggravate the vibration of the lattice, which will lower the mean free path of the carriers and decrease the impact ionization rate in the device. Therefore, the transient current induced by the carrier multiplication process will be smaller at a higher ambient temperature. Consequently, the maximum lattice temperature after the ion strike at Ta = 423 K (826 K) decreases compared with the case at Ta = 250 K (928 K) as shown in Figure 6. Our simulations show that the SEB tolerance of power MOSFETs has a positive correlation with the ambient temperature at the lower LET, which is in good agreement with the literature report [10-12].

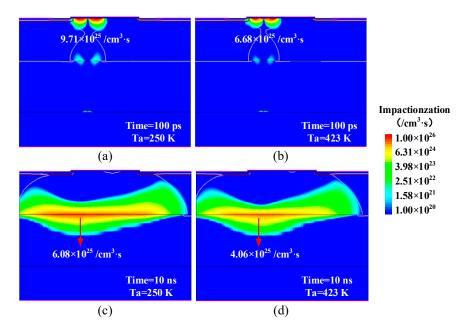


Figure 5. The impact generation rate distributions of the device at (a) Ta = 250 K, (b) Ta = 423 K, (c) Ta = 250 K, (d) Ta = 423 K after the ion strike under condition A (LET = $10 \text{ MeV} \cdot \text{cm}^2/\text{mg}$, Vds = 300 V).

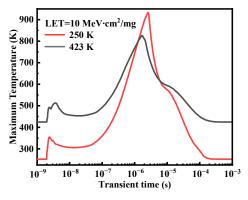
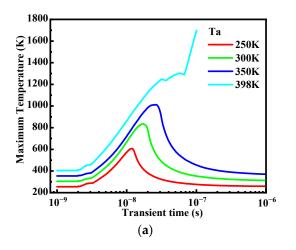


Figure 6. The temperature response after the ion strike under condition A (LET = $10 \text{ MeV} \cdot \text{cm}^2/\text{mg}$ and Vds = 300 V).

Micromachines 2023, 14, 1028 6 of 10

3.2.2. The Temperature Dependence of SEB at the Higher LET

Then, the condition at the higher LET (condition B) is discussed. Figure 7 shows the lattice temperature and drain current variations of the device versus time at different Ta under condition B (LET = $100 \text{ MeV} \cdot \text{cm}^2/\text{mg}$ and Vds = 100 V). It can be seen that the ambient temperature has an important influence on the SEB performance of power MOSFETs. With Ta increasing, the maximum lattice temperature after the ion strike varies from 604 K (Ta = 250 K) to 1010 K (Ta = 350 K). These values do not exceed the melting point of silicon and the lattice temperature can eventually recover to the ambient temperature, indicating no burnout. However, at Ta = 398 K, the maximum lattice temperature rises rapidly and can even exceed 1688 K, leading to catastrophic failure, and the drain current can maintain a high level. Once the SEB failure criterion is triggered, the simulation is terminated, and thus the simulation stops at about 100 ns at Ta = 398 K. From the transient temperature response, it can be observed that the device is more sensitive to SEB at a higher temperature under condition B, completely different from which we observed under condition A.



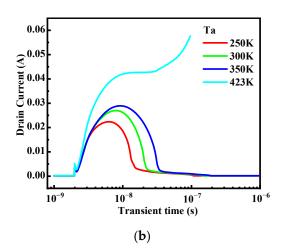


Figure 7. The SEB response of the power MOSFET at different Ta under condition B (LET = $100 \text{ MeV} \cdot \text{cm}^2/\text{mg}$ and Vds = 100 V) (a) temperature responses and (b) drain current responses.

To find out the reason why the temperature dependence of SEB differs at the different LETs, the electric field distributions at the lower LET and higher LET are comparatively studied in Figure 8. As we can see, because of the Kirk effect after the ion strike, the maximum electric field will move from the P-body/N-epi junction to the N-epi/N+buffer interface (N-buffer/N+sub interface) as time increases. However, the fewer generated charges at the lower LET cannot maintain the peak electric field at the N-epi/N-buffer interface and the maximum electric field returns to the P-body/N-epi junction after the ion strike 5 μ s. In contrast, at the higher LET, it will be easier to maintain the peak electric field at the N-buffer/N+sub interface due to the more deposited charges in the epitaxy layer [16]. Therefore, at the higher LET, whether the SEB will occur or not mainly depends on the state of the parasitic BJT, that is to say, whether the electron current provided by the amplification of parasitic BJT can sustain the impact ionization process to build up the regenerative feedback. As a result, the influence of temperature on parasitic BJT plays a dominant role in SEB failure under condition B.

Figure 9 shows the electron current density distributions near the parasitic BJT at (a) Ta = 250 K and (b) Ta = 423 K under condition B. It can be seen that the electron current density near the collector region increases greatly as time increases. This is because the heavy ion-induced hole current moves laterally through the P-body to be collected by the source electrode, resulting in a forward voltage drop across the emitter junction to turn on the parasitic BJT and amplify the current. Compared with the case at Ta = 250 K, the electron current density at Ta = 423 K is larger, especially after 150 ps. We assume that

Micromachines 2023, 14, 1028 7 of 10

the parasitic BJT is turned on when the ratio of the source electron current to source total current reaches 0.97 in our simulations, while it indicates that the emitter junction is turned on and the electron current has become the dominating part of the source total current. For Ta = 423 K, it only takes 99.2 ps to turn on the parasitic BJT while it takes 225.8 ps for Ta = 250 K. Thus, it can be concluded that the parasitic BJT at Ta = 423 K is turned on at about 100 ps earlier than that at Ta = 250 K. This is because the ambient temperature has a significant influence on the built-in voltage of the emitter-base junction and the base resistance (R_B). With the increasing temperature, the intrinsic carrier concentration will increase exponentially, reducing the built-in potential of the emitter-base junction, meanwhile, R_B increases caused by the decreased mobility, both making the parasitic BJT easier and earlier to be turned on [22,23].

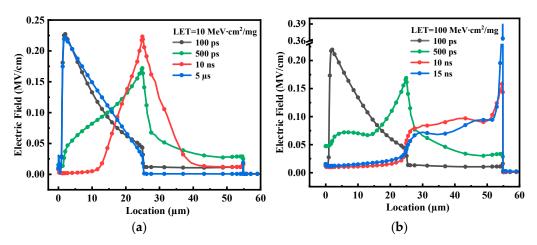


Figure 8. The electric field distributions of power MOSFET after an ion's strike at (a) LET = $10 \text{ MeV} \cdot \text{cm}^2/\text{mg}$ and (b) LET = $100 \text{ MeV} \cdot \text{cm}^2/\text{mg}$ (Ta = 250 K, Vds = 300 V).

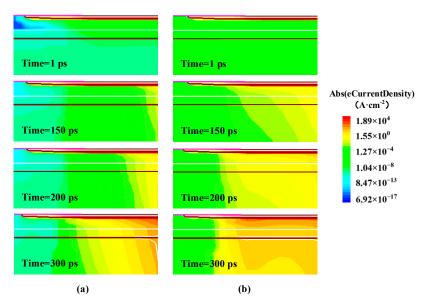


Figure 9. The electron current density distributions near the parasitic BJT after the ion strike at (a) Ta = 250 K and (b) Ta = 423 K under condition B.

Besides, we know that the lifetime of the carriers will increase with rising temperature, and it can be modeled by a power law [24,25]:

$$\tau(T) = \tau_0 \left(\frac{T}{300 \text{ K}}\right)^{\alpha} \tag{2}$$

Micromachines 2023, 14, 1028 8 of 10

where τ_0 is the carrier lifetime when the lattice temperature is 300 K, T is the lattice temperature and α is 2.1 [26]. Figure 10 shows a comparison of the minority carrier electron lifetime distribution in the base region of the BJT at Ta = 250 K and Ta = 423 K, from which we can observe the longer minority carrier electron lifetime at Ta = 423 K and it is consistent with the relationship between carrier lifetime and temperature as shown in (2). Thus, with the higher ambient temperature, more electrons injected from the emitter can be collected by the reverse-biased collector junction, resulting in a larger current gain.

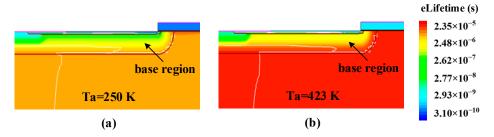


Figure 10. The minority carrier electron lifetime distribution in the base region under condition B after the ion strike 500 ps at (a) Ta = 250 K and (b) Ta = 423 K.

The current gain β of the parasitic BJT in the power MOSFET cannot be obtained directly because of the unmeasurable base current I_B . So, we calculate the common-base current gain α of the parasitic BJT firstly by extracting the ratio of the drain current (collector current) to the source current (emitter current) of the power MOSFET. Then we get the common-emitter current gain β of the BJT through the relationship between α and β as shown in (3). Table 3 gives the calculation results of the current gain of the parasitic BJT under condition B. As we can see, in the whole temperature range, the current gain varies from 19.44 to 123.24, increasing by a factor of five to six. The current gain of the BJT increases strongly with the increasing temperature, which is responsible for the large drain current and higher temperature rise at higher ambient temperature in Figure 7.

$$\beta = \frac{\alpha}{1 - \alpha} \tag{3}$$

Table 3. The current gain of the parasitic BJT of power MOSFET as a function of Ta under condition B after the ion strike 500 ps.

Та	$eta_{(BJT)}$
250 K	19.44
300 K	24.95
350 K	40.49
400 K	82.27
423 K	123.24

In conclusion, with the LET value increasing to 40 MeV·cm²/mg, the temperature dependence of SEB in power MOSFET is completely different from that of 10 MeV·cm²/mg. Due to the more deposited charges in the epitaxy region, it will be easier to maintain the peak electric field at the N-epi/N-buffer junction (N-buffer/N+sub interface), therefore, whether the parasitic BJT can provide continuous and enough electron current for the carrier multiplication process becomes extremely important. Our research shows that with the ambient temperature increasing, the difficulty to turn on the parasitic BJT decreases due to the lower built-in potential and the increasing base resistance, meanwhile, the current gain increases due to the increasing minority electron lifetime in the base region [27–30]. As a result, it is easier to form a local hot spot and induce the SEB failure at a higher temperature, thus power MOSFET is more sensitive to SEB at a higher temperature when the LET value is greater than 40 MeV·cm²/mg.

Micromachines 2023, 14, 1028 9 of 10

4. Conclusions

This paper presents the simulation-based comparison of the temperature dependence of single-event burnout (SEB) for power MOSFETs at different LETs. Our simulations show that at the lower LET (10 MeV·cm²/mg), with the ambient temperature (ranging from $-55\,^{\circ}\text{C}$ to $150\,^{\circ}\text{C}$) increasing, the peak impact ionization rate decreases, resulting in stronger robustness to SEB at a higher temperature. However, the temperature dependence of SEB exhibits different responses when the LET value increases to $40\,\text{MeV·cm²/mg}$. This is because the more deposited charges in the collector region make it easier to maintain the carrier multiplication thus the state of the parasitic BJT starts to play a dominant role in the regenerative feedback mechanism. When the LET is $100\,\text{MeV·cm²/mg}$, the time to turn on the parasitic BJT reduces from 225.8 ps to 99.2 ps, and the current gain increases from 19.44 to 123.24 due to the increase of the minority lifetime in the base region with the temperature increasing from 250 K to 423 K, which will be easier to form a local hot spot and induce the SEB failure at a higher temperature. Consequently, the power MOSFET exhibits stronger SEB robustness at a lower temperature at the higher LET.

Author Contributions: Data curation, Z.W.; funding acquisition, D.C.; methodology, C.W. and X.L.; project administration, Y.L.; writing—original draft, C.W., and X.L.; writing—review and editing, C.X. All authors have read and agreed to the published version of the manuscript.

Funding: This work was supported by the Equipment Pre-research Project of China under Grant 80924020307.

Data Availability Statement: Not applicable.

Conflicts of Interest: The authors declare no conflict of interest.

References

- 1. Waskiewicz, A.E.; Groninger, J.W.; Strahan, V.H.; Long, D.M. Burnout of Power MOS Transistors with Heavy Ions of Californium-252. *IEEE Trans. Nucl. Sci.* **1986**, 33, 1710–1713. [CrossRef]
- 2. Jia, Y.-P.; Peng, L.; Su, H.-Y.; Hu, D.-Q.; Wu, Y. Effect of grade doping buffer layer on SEE failure in VDMOSFET. In Proceedings of the 2016 IEEE 23rd International Symposium on the Physical and Failure Analysis of Integrated Circuits (IPFA), Singapore, 18–21 July 2016; pp. 276–279.
- 3. Johnson, G.H.; Hohl, J.H.; Schrimpf, D.R.; Galloway, F.K. Simulating single-event burnout of n-channel power MOSFET's. *IEEE Trans. Electron Devices* **1993**, *40*, 1001–1008. [CrossRef]
- 4. Liu, S.; Titus, J.L.; DiCienzo, C.; Cao, H.; Zafrani, M.; Boden, M.; Berberian, R. Recommended Test Conditions for SEB Evaluation of Planar Power DMOSFETs. *IEEE Trans. Nucl. Sci.* **2008**, *55*, 3122–3129. [CrossRef]
- 5. Titus, J.L. An Updated Perspective of Single Event Gate Rupture and Single Event Burnout in Power MOSFETs. *IEEE Trans. Nucl. Sci.* **2013**, *60*, 1912–1928. [CrossRef]
- 6. Hohl, J.H.; Johnson, G.H. Features of the triggering mechanism for single event burnout of power MOSFETs. *IEEE Trans. Nucl. Sci.* **1989**, *36*, 2260–2266. [CrossRef]
- 7. Zhou, X.-T.; Jia, Y.-P.; Hu, D.-Q.; Wu, Y. A Simulation-Based Comparison Between Si and SiC MOSFETs on Single-Event Burnout Susceptibility. *IEEE Trans. Electron Devices* **2019**, *66*, 2551–2556. [CrossRef]
- 8. Yu, C.-H.; Wang, Y.; Cao, F.; Huang, L.-L.; Wang, Y.-Y. Research of Single-Event Burnout in Power Planar VDMOSFETs by Localized Carrier Lifetime Control. *IEEE Trans. Electron Devices* **2015**, *62*, 143–148.
- 9. Yu, C.-H.; Wang, Y.; Liu, J.; Sun, L.-L. Research of Single-Event Burnout in Floating Field Ring Termination of Power MOSFETs. *IEEE Trans. Electron Devices* **2017**, *64*, 2906–2911. [CrossRef]
- 10. Johnson, G.H.; Schrimpf, R.D.; Galloway, K.F.; Koga, R. Temperature dependence of single-event burnout in n-channel power MOSFETs (for space application). *IEEE Trans. Nucl. Sci.* **1992**, *39*, 1605–1612. [CrossRef]
- 11. Nichols, D.K.; Coss, J.R.; McCarty, K.P. Single event gate rupture in commercial power MOSFETs. In Proceedings of the Second European Conference on Radiation and its Effects on Components and Systems (RADECS), Saint Malo, France, 13–16 September 1993; pp. 462–467.
- 12. Titus, J.L.; Wheatley, C.F. Experimental studies of single-event gate rupture and burnout in vertical power MOSFETs. *IEEE Trans. Nucl. Sci.* **1996**, *43*, 533–545. [CrossRef]
- 13. Nichols, D.K.; McCarty, K.P.; Coss, J.R. Observations of single event failure in power MOSFET's. In Proceedings of the 1994 IEEE Radiation Effects Data Workshop (REDW), Tucson, AZ, USA, 20–20 July 1994; pp. 41–54.
- 14. Huang, S.; Amaratunga, G.A.J.; Udrea, F. Analysis of SEB and SEGR in super-junction MOSFETs. *IEEE Trans. Nucl. Sci.* **2000**, 47, 2640–2647. [CrossRef]

Micromachines 2023, 14, 1028 10 of 10

15. Dachs, C.; Roubaud, F.; Palau, J.-M.; Bruguier, G. Simulation aided hardening of N-channel power MOSFETs to prevent single event burnout. *IEEE Trans. Nucl. Sci.* **1995**, 42, 1935–1939. [CrossRef]

- 16. Liu, S.; Boden, M.; Girdhar, D.; Dev, A.; Titus, J.L. Single-Event Burnout and Avalanche Characteristics of Power DMOSFETs. *IEEE Trans. Nucl. Sci.* **2006**, *53*, 3379–3385. [CrossRef]
- 17. Liu, S.; Titus, J.L.; Boden, M. Effect of Buffer Layer on Single-Event Burnout of Power DMOSFETs. *IEEE Trans. Nucl. Sci.* **2007**, *54*, 2554–2560. [CrossRef]
- 18. Luu, A.; Austin, P.; Miller, F. Sensitive Volume and Triggering Criteria of SEB in Classic Planar VDMOS. *IEEE Trans. Nucl. Sci.* **2010**, *57*, 1900–1907. [CrossRef]
- 19. Dachs, C.; Roubaud, F.; Palau, J.-M.; Bruguier, G.; Gasiot, J.; Tastet, P. Evidence of the ion's impact position effect on SEB in N-channel power MOSFETs. *IEEE Trans. Nucl. Sci.* **1994**, *41*, 2167–2171. [CrossRef]
- 20. Liao, X.-F.; Liu, Y.; Xu, C.-Q.; Li, J.; Yang, Y.-T. Single Event Burnout Hardening Technique for High-Voltage p-i-n Diodes with Field Limiting Rings Termination Structure. *IEEE Trans. Electron Devices* **2022**, *69*, *675*–*681*. [CrossRef]
- 21. Wrobel, T.F.; Coppage, F.N.; Hash, G.L.; Smith, A.J. Current Induced Avalanche in Epitaxial Structures. *IEEE Trans. Nucl. Sci.* 1985, 32, 3991–3995. [CrossRef]
- 22. Alexakis, P.; Alatise, O.; Hu, J.; Smith, A.J. Improved Electrothermal Ruggedness in SiC MOSFETs Compared with Silicon IGBTs. *IEEE Trans. Electron Devices* **2014**, *61*, 2278–2286. [CrossRef]
- 23. Sun, D.-P.; Law, M.-K.; Wang, B.; Mak, P.-I. Piecewise BJT process spread compensation exploiting base recombination current. In Proceedings of the 30th IEEE International Symposium on Circuits and Systems (ISCAS), Baltimore, MD, USA, 28–31 May 2017; pp. 1–4.
- 24. Tyagi, M.-S.; Overstraeten, R.-V. Minority carrier recombination in heavily doped silicon. *Solid-State Electron* **1983**, 26, 577–597. [CrossRef]
- Goebel, H.; Hoffmann, K. Full dynamic power diode model including temperature behavior for use in circuit simulators. In Proceedings of the 4th International Symposium on Power Semiconductor Devices and ICs (ISPSD), Tokyo, Japan, 19–21 May 1992; pp. 130–135.
- 26. Liao, X.-F.; Yang, Y.-T.; Liu, Y.; Xu, C.-Q. Simulation Aided Hardening of Power Diodes to Prevent Single Event Burnout. *IEEE Trans. Electron Devices* **2022**, *69*, 5088–5095. [CrossRef]
- 27. Kauffman, W.L.; Bergh, A.A. The temperature dependence of ideal gain in double diffused silicon transistors. *IEEE Trans. Electron Devices* **1968**, 15, 732–735. [CrossRef]
- 28. Dillard, W.C.; Jaeger, R.C. The temperature dependence of the amplification factor of bipolar-junction transistors. *IEEE Trans. Electron Devices* **1987**, 34, 139–142. [CrossRef]
- 29. Zhang, Y.-R.; Zhang, B.; Li, Z.-H.; Lai, C.-J.; Li, Z.-J. Analysis of thermal characteristics of current gain for BJT-BSIT compound device. In Proceedings of the 2008 International Conference on Communications, Circuits and Systems (ICCCAS), Fujian, China, 25–27 May 2008; pp. 1278–1281.
- 30. Liang, S.-W.; Wang, J.; Deng, L.-F.; Shi, Y.-Z. An Improved Proportional Base Driver for Minimizing Driver Power Consumption of SiC BJT Over Wide Current and Temperature Range. *IEEE J. Emerg. Sel. Top. Power Electron* **2019**, *7*, 1727–1735. [CrossRef]

Disclaimer/Publisher's Note: The statements, opinions and data contained in all publications are solely those of the individual author(s) and contributor(s) and not of MDPI and/or the editor(s). MDPI and/or the editor(s) disclaim responsibility for any injury to people or property resulting from any ideas, methods, instructions or products referred to in the content.