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Identification of MEMS Geometric Uncertainties through Homogenization

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Abstract: Fabrication imperfections strongly influence the functioning of Micro-Electro-Mechanical Systems (MEMS) if not taken into account during the design process. They must be indeed identified or precisely predicted to guarantee a proper compensation during the calibration phase or directly in operation. In this work, we propose an efficient approach for the identification of geometric uncertainties of MEMS, exploiting the asymptotic homogenization technique. In particular, the proposed strategy is experimentally validated on a MEMS filter, a device constituted by a complex periodic geometry, which would require high computational costs if simulated through full-order models. The complex periodic structure is replaced by an equivalent homogeneous medium, allowing a fast optimization procedure to identify imperfections by comparing a simplified analytical model with the experimental data available for the MEMS filter. The actual over-etch, obtained after the release phase, and the electrode offset of a fabricated MEMS filter are effectively identified through the proposed strategy.

Keywords: MEMS; auxetic structure; geometric uncertainties; over-etch; asymptotic homogenization



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1. Introduction

The rapid spread of high-performance Micro-Electro-Mechanical Systems (MEMS) in consumer applications, in the automotive industry, in virtual/augmented reality and in the internet of things [1–3] requires improved design methods, new fabrication strategies and efficient control procedures.

In particular, a lot of work has been devoted to improving MEMS fabrication processes [4] and to controlling imperfections that are inevitably present [5–8]. Despite the increasing effort in such topics, MEMS fabrication process imperfections still represent unknowns for MEMS designers and a significant limitation for devices' performance and reliability. In [6,9], for example, the effect of process-induced uncertainties on the performance of MEMS devices is addressed, while in [10], it is demonstrated how fabrication imperfections impact energy loss through mechanical mode coupling. In [11], the imperfections coming from both the fabrication and assembly/packaging of Radio-Frequency MEMS devices are studied and categorized, while in [12], sensitivity analyses are proposed to study the effect of geometric and material uncertainties on the performance of MEMS resonators.

The direct experimental measurement of fabrication imperfections is very complex and time-consuming due to the scale of the devices. For this purpose, optical/scanning electron microscopy [13] or laser interferometry [14] is usually employed. However, these optical techniques cannot provide significant information on the fabrication imperfections of the final geometry of the device since MEMS are typically encapsulated and consequently not directly accessible through visual inspection. On the other hand, electrostatic measurements represent a very powerful experimental technique, but, while affected by uncertainties, they do not allow for the direct quantification of fabrication process imperfections.

Analytical or numerical models able to predict the correct functioning of MEMS devices also in the presence of fabrication process imperfections are fundamental for design, as evidenced in [15] for a tuning-fork gyroscope or in [16] for an electrostatically actuated microbeam resonator.

In [17], a theory to predict the strength of microstructures in the presence of defects is proposed and good agreement with experimental results is demonstrated; in [18], the effect of etch holes on ferromagnetic MEMS is modelled, while in [19], Monte Carlo analyses are combined with finite element method (FEM) simulations to investigate the effect of the grain morphology and orientation on the effective elastic properties of polysilicon structures usually employed in MEMS.

Inverse identification of the unknown mechanical, geometric and material parameters through the combination of experiments and finite element models seems a promising strategy, but its huge computational cost arising from the use of full-order FEM simulations makes it hardly applicable to the complex structures typical of MEMS devices. To overcome such limitations, parametric model order reduction techniques based on the combination of the proper orthogonal decomposition and the kriging metamodeling have been recently proposed in [20,21].

An alternative, new and promising strategy is proposed and applied in this work to reduce the computational cost of the inverse identification of fabrication imperfections in geometrically complex MEMS. Our proposal makes use of asymptotic homogenization. The two-scale homogenization is a mathematical tool that provides explicit expressions of the effective properties of periodically heterogeneous materials, under the hypothesis of linearity and scale separation [22,23]. The method has been widely employed in solid mechanics to study the static and dynamic behavior of composite materials and metamaterials [24–27], but seldom applied to MEMS, to the best of authors' knowledge. In this work, we combine an analytical model with the asymptotic homogenization technique to identify two typical geometric unknowns of capacitive devices: the electrode offset and the over-etch. The electrode offset is here defined as the discrepancy between nominal and after-fabrication gaps between movable components, i.e., rotor, and fixed electrodes, i.e., stators. The overetch is instead defined as the deviation from the nominal geometric dimensions of the fabricated mechanical structure. Such discrepancy is an inevitable consequence of the acid attack exploited during the MEMS fabrication to selectively remove the oxide deposited below the polysilicon and to consequently obtain suspended mechanical structures able to move with respect to the substrate. The acid indeed penetrates in the polysilicon structure from the free edges and reduces their in-plane dimensions by a quantity that is referred to as over-etch. Its actual value is unknown and strongly affects the behavior of the device, especially when elements with small in-plane dimensions are present, e.g., suspension springs. In the present work, we implement a proper optimization procedure based on the minimization of the difference between modeling predictions (analytical model plus homogenization) and experimental data coming from electro-mechanical tests.

The paper is organized as follows. The asymptotic homogenization technique is introduced in Section 2 and applied to a MEMS electrically tunable mechanical filter in Section 3 to identify an equivalent continuum. In Section 4, the new identification procedure of over-etch and electrode offset is detailed and then applied to the MEMS filter previously studied; conclusions are finally summarized in Section 5.

2. Asymptotic Homogenization

Numerical analyses of periodic media often represent a huge computational burden, especially when the size ℓ of the microstructure is very small with respect to the macroscopic dimension L of the body. In such a case, i.e., when $\epsilon = \ell/L \ll 1$, the problem can be studied by employing two-scale asymptotic homogenization, which is a mathematical tool that provides the effective homogenized properties of a periodic medium.

We consider a single-phase linear-elastic body Ω , such as the one shown in Figure 1a, which is constructed by a periodic repetition of a unit cell Y^{ϵ} , shown in Figure 1b, in which

only a subset $Y_m^{\epsilon} \subset Y^{\epsilon}$ is filled by the material. The geometry of the unit cell is completely free at this stage; a specific example will be provided in the following section (Figure 2, right panel).

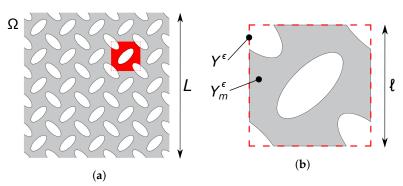


Figure 1. (a) Single-phase heterogeneous medium with periodic structure. (b) Close-up view of its unit cell.

In the small strains and displacements hypothesis, the static equilibrium of the periodic media is governed by the differential equations

$$\operatorname{div}(\mathbb{D}^{\epsilon}: \epsilon(\mathbf{u}^{\epsilon})) + \mathbf{F}^{\epsilon} = \mathbf{0}, \tag{1}$$

where \mathbf{F}^{ϵ} are the body forces, \mathbf{u}^{ϵ} is the displacement field, $\epsilon(\diamond)$ is the symmetric part of the gradient of (\diamond) and \mathbb{D}^{ϵ} is the fourth-order elastic stiffness tensor of the constituent material.

Using the standard argument of asymptotic homogenization [22,23], we denote by x the macroscopic variable and by $y = x/\epsilon$ the microscopic one, which lives in the re-scaled unit cell $Y = Y^{\epsilon}/\epsilon$. The assumed material periodicity gives

$$\mathbb{D}^{\epsilon}(\mathbf{x}) = \mathbb{D}\left(\frac{\mathbf{x}}{\epsilon}\right),\tag{2}$$

while body forces can be non-uniform and read

$$\mathbf{F}^{\epsilon}(\mathbf{x}) = \mathbf{F}\left(\mathbf{x}, \frac{\mathbf{x}}{\epsilon}\right). \tag{3}$$

The solution of (1) is searched by developing the displacement with respect to ϵ in the form

$$\mathbf{u}^{\epsilon}(\mathbf{x}) = \mathbf{u}^{0}\left(\mathbf{x}, \frac{\mathbf{x}}{\epsilon}\right) + \epsilon \mathbf{u}^{1}\left(\mathbf{x}, \frac{\mathbf{x}}{\epsilon}\right) + o(\epsilon),$$
 (4)

where all the fields on the right-hand side of Equation (4) are defined on $\Omega \times Y_m$ and are periodic with respect to \mathbf{y} . For any vector function of \mathbf{x} and \mathbf{y} , we will denote by a subscript \mathbf{x} and \mathbf{y} the partial derivative with respect to \mathbf{x} and \mathbf{y} ; for instance, $\varepsilon_{\mathbf{x}}$ and $\varepsilon_{\mathbf{y}}$ will denote the symmetric gradients of \mathbf{u} with respect to \mathbf{x} and \mathbf{y} , respectively.

It is possible to prove—see [23] for the full derivation—that the 0-th order displacement only depends on the macroscopic variable, i.e., $\mathbf{u}^0(\mathbf{x}, \mathbf{y}) = \mathbf{U}^0(\mathbf{x})$, and that the effective equilibrium equation is given by

$$\operatorname{div}_{\mathbf{x}}\left(\mathbb{D}^{0}: \varepsilon_{\mathbf{x}}(\mathbf{U}^{0})\right) + \mathbf{F}^{0} = \mathbf{0}. \tag{5}$$

In Equation (5), \mathbb{D}^0 is the homogenized stiffness tensor of the periodic media, whose components are evaluated as

$$D_{ijhk}^{0} = \frac{1}{|Y|} \int_{Y_m} \left(\mathbf{e}_i \odot \mathbf{e}_j + \varepsilon_{\mathbf{y}}(\boldsymbol{\chi}^{ij}) \right) : \mathbb{D} : \left(\mathbf{e}_h \odot \mathbf{e}_k + \varepsilon_{\mathbf{y}}(\boldsymbol{\chi}^{hk}) \right) d\mathbf{y}, \tag{6}$$

 \mathbf{e}_i being the unit vector in the *i*-th direction, \odot the symmetric tensorial product and $\chi^{ij}(\mathbf{y})$ the solution of the cell problem

$$\begin{cases}
\operatorname{div}_{\mathbf{y}}(\mathbb{D} : \varepsilon_{\mathbf{y}}(\chi^{ij})) = \mathbf{0} & \text{in } Y_{m} \\
\chi^{ij} \text{ periodic} & \text{on } \partial Y_{m} \cap \partial Y \\
[\mathbb{D} : (\varepsilon_{\mathbf{y}}(\chi^{ij}) + \mathbf{e}_{i} \odot \mathbf{e}_{j})] \cdot \mathbf{n} \text{ anti-periodic} & \text{on } \partial Y_{m} \cap \partial Y \\
[\mathbb{D} : (\varepsilon_{\mathbf{y}}(\chi^{ij}) + \mathbf{e}_{i} \odot \mathbf{e}_{j})] \cdot \mathbf{n} = \mathbf{0} & \text{on } \partial Y_{m} \setminus \partial Y
\end{cases} \tag{7}$$

The field $\chi^{ij}(\mathbf{y})$ represents the elastic displacement of the re-scaled unit cell, up to a rigid body motion, when it is subjected to periodic boundary conditions and the uniform eigenstrain $\mathbf{e}_i \odot \mathbf{e}_j$.

As can be seen from (6), the effective stiffness tensor possesses all the minor and major symmetries, i.e., $D^0_{ijhk} = D^0_{hkij} = D^0_{jihk} = D^0_{ijkh}$. However, it should be noticed that, in general, \mathbb{D}^0 is anisotropic even if the constituent material is isotropic.

Finally, in Equation (5), \mathbf{F}^0 are the homogenized body forces, which are given by

$$\mathbf{F}^{0}(\mathbf{x}) = \frac{1}{|Y|} \int_{Y_{m}} \mathbf{F}(\mathbf{x}, \mathbf{y}) \ d\mathbf{y}. \tag{8}$$

3. MEMS Filter

3.1. Mechanical Design

The innovative MEMS electrically tunable mechanical filter recently proposed in [28,29] is here employed to demonstrate the efficiency of the proposed fabrication imperfection identification procedure.

A scanning electron microscopy (SEM) image of the MEMS filter, fabricated through surface micromachining by the STMicroelctronics Thelma© (Thick Epitaxial Layer for Micro-Gyroscopes and Accelerometers) process [3], is shown in Figure 2. The device consists of a central structure Ω made by a 5 \times 5 periodic repetition of the auxetic unit cell reported in the right-hand close-up schematic view of Figure 2, and four external frames Tu, Td, S and D rigidly connected to the central structure through thick beams and suspended to the substrate through folded springs.

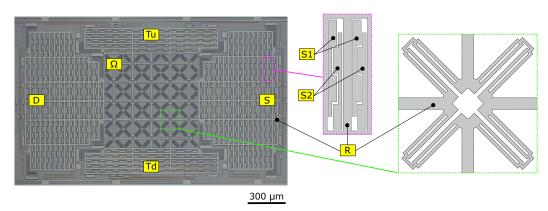


Figure 2. SEM immage of the MEMS structure (**left**) with a close-up view of the stators S1, S2 (**mid**) and the auxetic unit cell (**right**).

Electrodes for the electrostatic push–pull actuation and differential readout are located inside the four frames according to the scheme shown in Figure 2 (middle panel). The central part of the device represents an example of a two-dimensional periodic medium, as studied in Section 2; hence, homogenization will be applied to this part. The four external frames are also "quasi-periodic" (i.e., the constituent cells slightly vary in space), but they will not be homogenized as their geometry is simple and their deformation in operation is negligible. A coarse mesh can be then employed to discretize external frames without

losing accuracy and simultaneously not affecting the overall computational cost of the FEM simulation.

3.2. Homogenization of the Auxetic Core

The inner core of the MEMS structure is here analyzed through the asymptotic homogenization. In particular, after solving the cell problems (7), the effective properties of the media are computed from Equation (6). Figure 3 shows the solution of the three cell problems in which uniform eigenstrains are applied within the unit cell: (a) normal horizontal, (b) normal vertical and (c) angular eigenstrain.

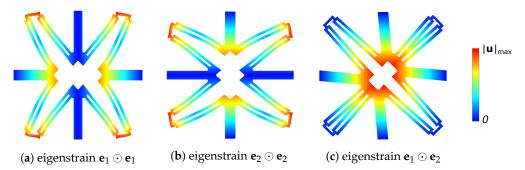


Figure 3. Contour of the displacement magnitude over the (magnified) deformed shape of the solution of cell problems (a) χ^{11} , (b) χ^{22} and (c) χ^{12} .

The in-plane anisotropy of the auxetic core can be visualized by computing the effective in-plane Young's modulus E^0 and Poisson's ratio v^0 as a function of the stretching direction $\mathbf{d} = \cos \vartheta \mathbf{e}_1 + \sin \vartheta \mathbf{e}_2$, being ϑ the angle with respect to the horizontal direction. Introducing the transversal direction $\mathbf{t} = \sin \vartheta \mathbf{e}_1 - \cos \vartheta \mathbf{e}_2$ —see [30] for further details—one has

$$E^{0}(\vartheta) = \frac{1}{\mathbf{d} \otimes \mathbf{d} : \mathbb{C}^{0} : \mathbf{d} \otimes \mathbf{d}} \quad \text{and} \quad \nu^{0}(\vartheta) = -\frac{\mathbf{d} \otimes \mathbf{d} : \mathbb{C}^{0} : \mathbf{t} \otimes \mathbf{t}}{\mathbf{d} \otimes \mathbf{d} : \mathbb{C}^{0} : \mathbf{d} \otimes \mathbf{d}'}$$
(9)

where \mathbb{C}^0 is the effective compliance of the periodic media, i.e., the inverse tensor of \mathbb{D}^0 .

The homogenized properties are shown in the polar plots of Figure 4 as a function of $\vartheta \in [0, \pi/2]$ for different values of the over-etch. As expected, for increasing values of the over-etch, the auxetic core of the MEMS filter becomes more and more compliant since E^0 monotonically decreases. When the direction of stretching is mainly horizontal or vertical, the inner core exhibits an auxetic behavior ($v^0 < 0$), which is quite independent of the over-etch.

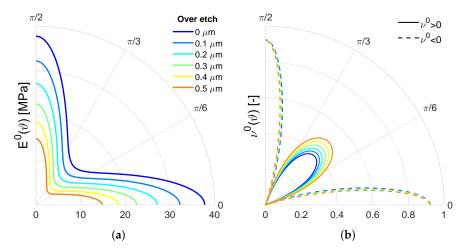


Figure 4. Polar plots of the homogenized Young's modulus (**a**) and Poisson's ratio (**b**) as a function of ϑ for different values of the over-etch. Negative values of v^0 are indicated with dashed lines.

3.3. Validation of Effective Properties

The asymptotic homogenization technique can be employed to reduce the computational burden of numerical analyses involving periodic linear-elastic media, without a significant loss in terms of accuracy.

To validate the homogenized properties of the auxetic core of the MEMS structure of Figure 2, we compare the result of finite element analyses (FEA) obtained in the case of the real geometry and those obtained by replacing the inner core with an equivalent homogeneous medium.

We simulate the case in which the left and the right frames of the MEMS structure are displaced outwards of \bar{U} in the horizontal direction. The FEA are carried out in the case of the MEMS nominal geometry with the commercial finite element software COMSOL Multiphysics.

As discussed in [31], a fine discretization is required to correctly reproduce the bending of the thin elements constituting the auxetic unit cells and the anchoring springs. The mesh employed for the real geometry is composed of 1,600,000 quadratic serendipity tetrahedral elements. When the inner core is replaced by the effective homogenized medium, a coarser mesh can be used and the number of elements is reduced to 600,000.

Figure 5 shows the contours of the horizontal (a,b) and vertical (c,d) displacement fields of the inner core of the MEMS structure for the real and the homogenized model. The fields are non-uniform due to the fact that only the three central cells are attached to the frames. Due to the auxetic behavior, the structure expands in the vertical direction. The homogenization approach allows a reduction in the computational time of 90% and a 4% error ($\bar{V}=0.669\bar{U}$ FEM with actual geometry and $\bar{V}=0.696\bar{U}$ FEM with homogenized medium) in terms of vertical displacement of the top/bottom frame.

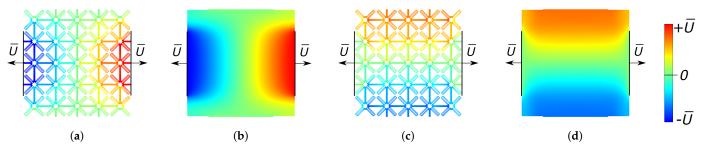


Figure 5. Contours of the horizontal (**a**,**b**) and vertical (**c**,**d**) displacement of the MEMS auxetic core (**a**,**c**) and the corresponding homogenized medium (**b**,**d**).

4. Identification of Geometric Uncertainties

4.1. Experimental Results

To experimentally validate the hybrid modeling reduction strategy, an electrical characterization in the static regime of the push–pull parallel-plate capacitors located in the external frames of the auxetic MEMS device is performed. Figure 6a shows the block diagram of the experimental setup implemented to perform the electrical characterization of the MEMS device.

The capacitance between each distinctive stator terminal, S1, S2, D1, D2, Tu1, Tu2, Td1 and Td2, and the common rotor terminal R is measured by exploiting the impedance meter HP4194A by applying a proper sinusoidal probing signal $v_p(t) = A_p \sin(2\pi f_p t)$ and a tunable DC-bias voltage V_b . Specifically, Figure 6a reports the experimental configuration used to measure the capacitance C_{S1-R} , i.e., the capacitance between terminals S1 and R. In the reported configuration, the probing signal $v_p(t)$ and the DC-bias voltage V_b is applied between terminals S1 and R while keeping all other terminals grounded to avoid unwanted induced electrostatic forces. The bias voltage V_b is tuned within the range 0–12 V with a step size of 0.5 V to induce a tunable electrostatic force between the parallel plates S1-R. We suitably set the maximum value of V_b to prevent the pull-in effect and the root mean square value of the amplitude A_p to 50 mV to produce a negligible electrostatic

force compared to the one induced by V_b . The frequency f_p is chosen equal to 200 kHz to avoid mechanical resonances of the MEMS structure. The first translational/auxetic modes actuable through the electrodes located on the D (or S) frame are indeed in the frequency range 16 kHz-60 kHz.

A similar experimental configuration is implemented to measure all other capacitances, i.e., C_{S2-R} , C_{D1-R} , C_{D2-R} , C_{Tu1-R} , C_{Tu2-R} , C_{Td1-R} and C_{Td2-R} . The top view of the tailored printed circuit board (PCB) developed for the static characterization of the auxetic MEMS device bonded in the 68-pin plastic-leaded-chip-carrier (PLCC68) package is reported in Figure 6b.

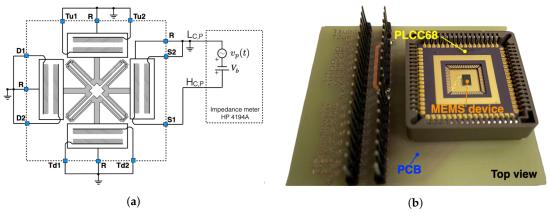
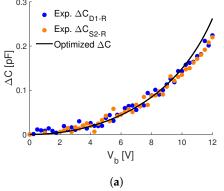


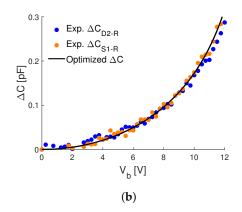
Figure 6. (a) Block diagram of the experimental setup employed for the static characterization of the auxetic MEMS device; (b) image of tailored printed circuit board (PCB) containing the MEMS device housed in a PLCC68 package.

The experimental capacitance–voltage curves between the rotor and each one of the stators of the MEMS structure are reported in Figure 7 with dots, having defined $\Delta C_{\alpha-R}^{exp}(V_b) = C_{\alpha-R}^{exp}(V_b) - C_{\alpha-R}^{exp}(0)$.

Due to the symmetries of the MEMS nominal geometry, in the absence of any non-symmetric imperfection, the experimental curves in Figure 7a,c should be identical to those of Figure 7b,d (respectively). As can be seen, this is not the case for Figure 7a,b, since, at a given bias voltage, the experimental values of ΔC_{D1-R}^{exp} and ΔC_{S2-R}^{exp} are lower with respect to ΔC_{D2-R}^{exp} and ΔC_{S1-R}^{exp} . This effect can be explained by an initial gap between the rotor and the stators D1, S2 that is larger than the one between the rotor and the stators D2, S1, i.e., arguing that the rotor is rigidly shifted in the horizontal direction with respect to the substrate. This shift can actually occur due to geometric imperfection and/or to the presence of pre-stresses in the suspension springs.







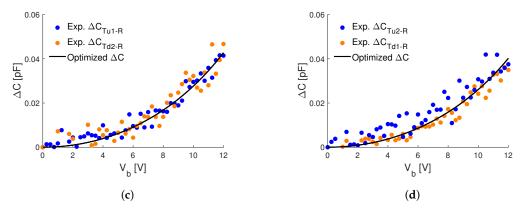


Figure 7. Capacitance–voltage curves, between the rotor and the different stators (a) Exp. ΔC_{D1-R} ; Exp. ΔC_{S2-R} , (b) Exp. ΔC_{D2-R} ; Exp. ΔC_{S1-R} , (c) Exp. ΔC_{Tu1-R} ; Exp. ΔC_{Td2-R} , (d) Exp. ΔC_{Tu2-R} ; Exp. ΔC_{Td1-R} , obtained experimentally (dotted), and the optimal analytical ones (black continuos).

4.2. Optimization Procedure

In this work, we focus on the identification of only two possible geometric uncertainties of a MEMS device, but the same approach can be applied also in the case of more fabrication imperfections. The first one is the over-etch oe, which significantly modifies the MEMS geometry and the stiffness of its components. The second parameter is a possible rigid horizontal shift X_S of the rotor R with respect to the substrate, which leads to asymmetric gaps between the stators of the left and right frames.

When measuring the capacitance–voltage relationship between the rotor R and a stator α , with $\alpha = D1, D2, S1, S2, Tu1, Tu2, Td1$ or Td2, the functioning of the MEMS structure can be described by the simplified scheme shown in Figure 8a. The left plate represents the stators α fixed on the substrate, while the right one represents the rotor R, which is anchored to the substrate by means of a spring.

The stiffness $k_{\alpha,eq}$ of the spring represents the linear-elastic equivalent stiffness of the MEMS frame where the stators α are located. These values can be evaluated by FEA, as a function of the over-etch oe, exploiting the homogenization of the inner auxetic core; see [31] for details. For example, the equivalent stiffness of the D frame is evaluated as the ratio of the applied uniform horizontal force on the frame and its (average) horizontal displacement, while the stiffness of the Tu frame is evaluated as the ratio between the uniform vertical force on the frame and the corresponding vertical displacement.

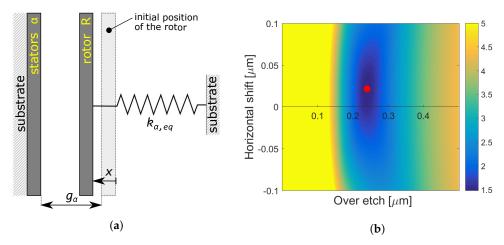


Figure 8. (a) Schematic representation of the analytical model; (b) contour of the error as a function of the over-etch and horizontal shift.

In the hypothesis that all capacitors in the MEMS device act as parallel plates, which is quite reasonable since the dimensions of the capacitors are far larger than their distance, the capacitance between α and R can be evaluated as

$$C_{\alpha-R}^{an} = \frac{\varepsilon_0 A_{\alpha}}{g_{\alpha} - x'} \tag{10}$$

where ε_0 is the vacuum permittivity, A_α is the total surface of the plates, x is the displacement of the rotor and g_α is the initial gap between the plates. This latter can be expressed as a function of the fabrication imperfections here under investigation as

$$g_{\alpha}(oe, X_S) = \begin{cases} g_{\alpha,0} + 2oe + X_S & \text{if } \alpha = D1, S2 \\ g_{\alpha,0} + 2oe - X_S & \text{if } \alpha = D2, S1 \\ g_{\alpha,0} + 2oe & \text{if } \alpha = Tu1, Tu2, Td1, Td2 \end{cases}$$
(11)

where $g_{\alpha,0}$ is the nominal initial gap between the plates. The attractive electrostatic force acting on the rotor can be evaluated as

$$F = \frac{1}{2}V^2 \frac{dC}{dx} \simeq \frac{1}{2}V_b^2 \frac{\varepsilon_0 A_\alpha}{(g_\alpha - x)^2},\tag{12}$$

where the simplification $V^2 \simeq V_b^2$ is justified by the small amplitude of the probing signal $v_p(t)$ employed in the experimental tests.

Enforcing the balance between the electrostatic force (12) and the restoring elastic force $k_{\alpha,eq}x$, one obtains

$$V_b^2 \left(\Delta C_{\alpha-R}^{an} + \frac{\varepsilon_0 A_\alpha}{g_\alpha} \right)^3 - 2\varepsilon_0 A_\alpha k_{\alpha,eq} g_\alpha \Delta C_{\alpha-R}^{an} = 0, \tag{13}$$

which defines implicitly the analytical capacitance–voltage relation between the stators α and the rotor R, as a function of oe and X_S , where $\Delta C_{\alpha-R}^{an} = C_{\alpha-R}^{an} - \varepsilon_0 A_\alpha / g_\alpha$.

Once the analytical and experimental capacitances are obtained from Equation (13) and measurements, respectively, the optimal over-etch oe and horizontal shift X_S are chosen as those values that minimize the relative error

$$\mathcal{E}(oe, X_S) = \sum_{\alpha} \frac{||\Delta C_{\alpha-R}^{exp}(V_b) - \Delta C_{\alpha-R}^{an}(V_b, oe, X_S)||_{\ell^2}}{||\Delta C_{\alpha-R}^{exp}(V_b)||_{\ell^2}},$$
(14)

where $||(\diamond)||_{\ell^2}$ is the ℓ^2 -norm of (\diamond) .

4.3. Results

Considering reasonable intervals of the uncertainty parameters, e.g., $oe \in [0,0.5]$ µm and $X_S \in [-0.1,+0.1]$ µm, one can evaluate the relative error (14) between the analytical and experimental capacitance curves. Note that the error should be evaluated in a limited voltage interval, e.g., between 0 V and 10 V, to avoid nonlinear phenomena that may occur at high voltage and that are not taken into account in the equivalent stiffness analytical model.

Figure 8b shows the contour of the relative error as a function of the over-etch and the horizontal shift. In particular, the yellow indicates the region where $\mathcal{E} \geq 5$, while the red marker indicates the minimum of the error, which occurs at $oe=0.24~\mu m$ and $X_S=0.021~\mu m$. The optimal over-etch is in good agreement with typical values available for the fabrication process employed for the MEMS device under study [32].

The analytical capacitance–voltage curves associated with the optimal values of overetch and horizontal shift are shown in Figure 7 with solid black lines and are in good agreement with the experimental curves.

5. Conclusions

In this work, we propose a time-saving approach for the identification of geometric uncertainties of MEMS devices having a periodic structure.

In particular, we consider a tunable MEMS mechanical filter, already proposed in a previous work by some co-authors, and study the linear-elastic effective properties of the periodic core by employing the two-scale asymptotic homogenization technique as a function of over-etch. In this way, the complex geometry of the periodic media has been replaced by an equivalent homogeneous one, allowing a strong reduction in computational time without a significant loss in terms of accuracy, even with only 5×5 cells.

We experimentally measure the capacitance–voltage relation between the rotor and the different stators of the MEMS in order to characterize the real functioning of the device. Then, we propose a simplified electro-mechanical model able to predict the MEMS behavior and able to account for the geometric uncertainties to be identified, i.e., over-etch and electrode offset in this work.

The optimal values of such geometric imperfections are identified as those that minimize the relative error between experimental data and analytical predictions. The identified parameters are in agreement with typical values available for the employed fabrication process and provide analytical capacitance—voltage curves that correctly reproduce the experimental ones.

This first contribution shows one of the potentialities of asymptotic homogenization in the identification of geometric imperfections in fabricated MEMS. A similar approach would be also very effective to develop efficient design tools for complex microsystems.

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Conflicts of Interest: The authors declare no conflicts of interest.

Abbreviations

The following abbreviations are used in this manuscript:

FEA Finite Element Analysis

MEMS Micro-Electro-Mechanical Systems

PCB Printed Circuit Board

PLCC68 68-Pin Plastic-Leaded-Chip-Carrier SEM Scanning Electron Microscopy

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